

PHASOR ANALYSIS

Phasor representation:

$$A \cos(\omega t + \theta) = \operatorname{Re}\{A e^{j(\omega t + \theta)}\} = \operatorname{Re}\{A e^{j\theta} e^{j\omega t}\}$$

$A e^{j\theta}$ is the phasor representation of $A \cos(\omega t + \theta)$

we must keep ω around as "side information"

Now: show how phasor representation can be used to simplify circuit analysis when:

1. Circuit has sinusoidal sources, all of the same frequency
2. We are only interested in the steady-state response.

Can handle sinusoidal sources of different frequencies by superposition (in the time domain).

PHASOR RELATIONSHIPS

KCL:



At all points in time: $i_1(t) = i_2(t) + i_3(t)$

Therefore, if these are sinusoidal signals

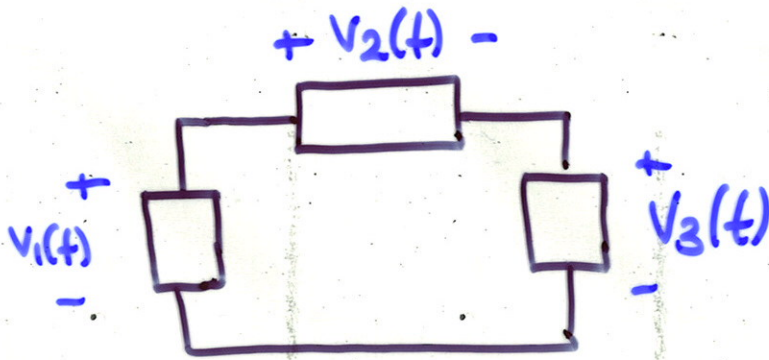
$$A \cos(\omega t + \theta) = B \cos(\omega t + \phi) + C \cos(\omega t + \psi)$$

$$\Rightarrow \operatorname{Re}\{Ae^{j\theta} \cdot e^{j\omega t}\} = \operatorname{Re}\{Be^{j\phi} \cdot e^{j\omega t}\} + \operatorname{Re}\{Ce^{j\psi} \cdot e^{j\omega t}\}$$

$$\Rightarrow Ae^{j\theta} = Be^{j\phi} + Ce^{j\psi}$$

i.e., $\mathbf{I}_1 = \mathbf{I}_2 + \mathbf{I}_3$

KVL:



At all times $-v_1(t) + v_2(t) + v_3(t) = 0$

For sinusoidal signals

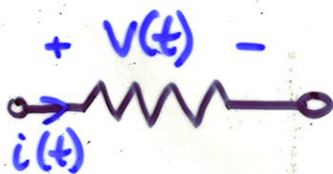
$$-A \cos(\omega t + \theta) + B \cos(\omega t + \phi) + C \cos(\omega t + \psi) = 0$$

$$\Rightarrow -\operatorname{Re}\{A e^{j\theta} \cdot e^{j\omega t}\} + \operatorname{Re}\{B e^{j\phi} \cdot e^{j\omega t}\} + \operatorname{Re}\{C e^{j\psi} \cdot e^{j\omega t}\} = 0$$

$$\Rightarrow -A e^{j\theta} + B e^{j\phi} + C e^{j\psi} = 0$$

i.e., $-V_1 + V_2 + V_3 = 0$

RESISTORS



$$v(t) = Ri(t)$$

$$\text{if } i(t) = A \cos(\omega t + \theta)$$

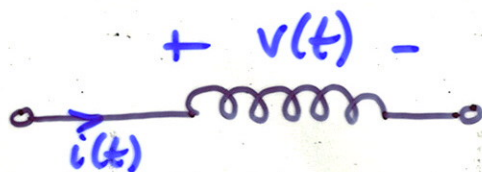
$$v(t) = RA \cos(\omega t + \theta)$$

That is, in phasor domain

$$\mathbf{V} = RAe^{j\theta}$$

$$= R\mathbf{I}$$

INDUCTORS



$$v(t) = L \frac{di(t)}{dt}$$

if $i(t) = A \cos(\omega t + \theta)$

$$v(t) = -\omega L A \sin(\omega t + \theta)$$

$$= \omega L A \cos(\omega t + \theta + \pi/2)$$

Voltage leads current by 90°

Phasor domain:

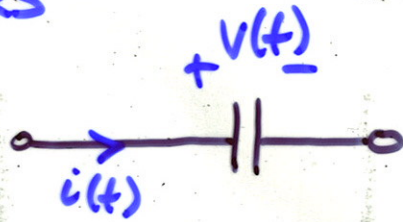
$$V = \omega L A e^{j(\theta + \pi/2)}$$

$$= e^{j\pi/2} \omega L A e^{j\theta}$$

$$= e^{j\pi/2} \omega L I$$

$$= j\omega L I$$

CAPACITORS



$$v(t) = \frac{1}{C} \int_{-\infty}^t i(x) dx$$

if $i(t) = A \cos(\omega t + \theta)$

$$v(t) = \frac{1}{\omega_0 C} A \sin(\omega t + \theta)$$

$$= \frac{1}{\omega_0 C} A \cos(\omega t + \theta - \pi/2)$$

voltage lags current by 90°

Phasor domain:

$$V = \frac{1}{\omega_0 C} A e^{j(\theta - \pi/2)}$$

$$= \frac{e^{-j\pi/2}}{\omega_0 C} A e^{j\theta}$$

$$= \frac{1}{j\omega_0 C} \mathbf{I}$$

IMPEDANCE

For resistors, inductors and capacitors, we have seen

$$V = Z I$$

Like Ohm's Law, but in the phasor domain

Z called impedance

In general, $Z = R + jX$

R : Resistance
 X : Reactance

We also define admittance

$$Y = \frac{I}{V}$$
$$= G + jB$$

G : conductance
 B : susceptance

$$Y = \frac{1}{Z} = \frac{1}{R + jX}$$

$$= \frac{1}{R + jX} \cdot \frac{R - jX}{R - jX} = \frac{R}{R^2 + X^2} - j \frac{X}{R^2 + X^2}$$

Note that $G \neq \frac{1}{R}$, $B \neq \frac{1}{X}$
except in special cases

What have we achieved?

- KCL, KVL work in the phasor domain
- For resistors, inductors and capacitors,

$$V = ZI$$

- Therefore

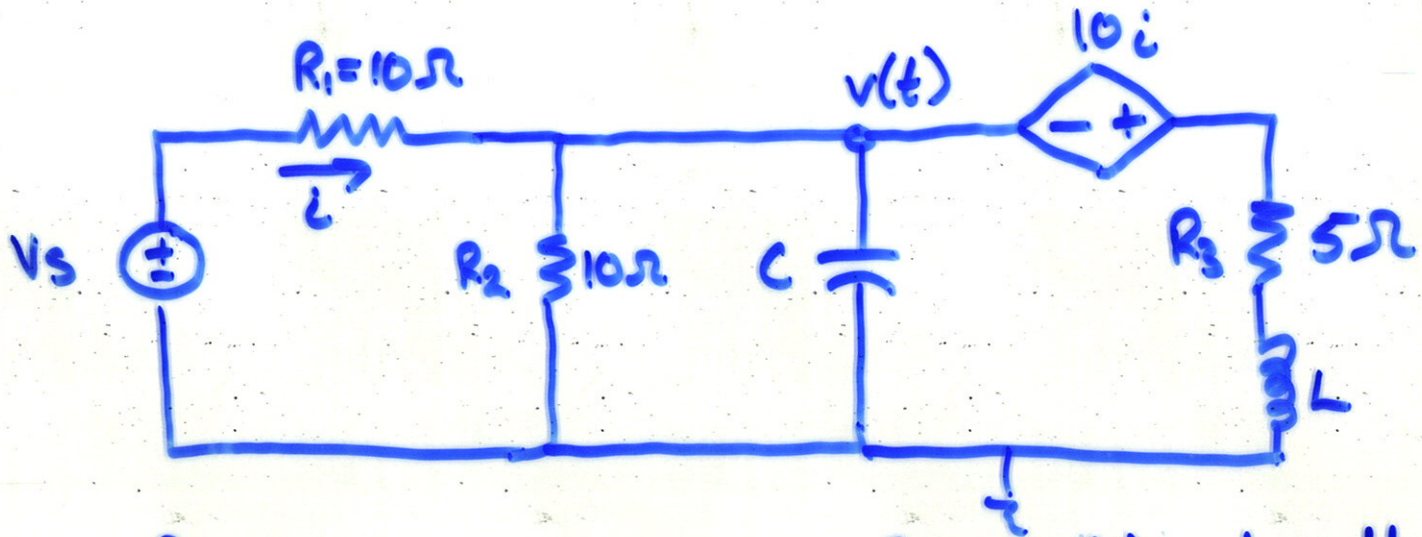
— Node, Mesh, Thevenin, Norton, Superposition
apply directly

- Impedances in series add.
- Admittances in parallel add

- As a result:

— for sinusoidal sources of one frequency,
finding the steady-state solution is only as hard as
circuits with DC sources and resistors
but algebra is complex rather than real

* Here we'll just do an example:



If $L = 0.5\text{ H}$, $C = 10\text{ mF}$, find $v(t)$ when the circuit has reached the steady-state, if $V_s(t) = 10 \cos(10t)$ Volts.

Step 1 - Identify supernode for dependent source

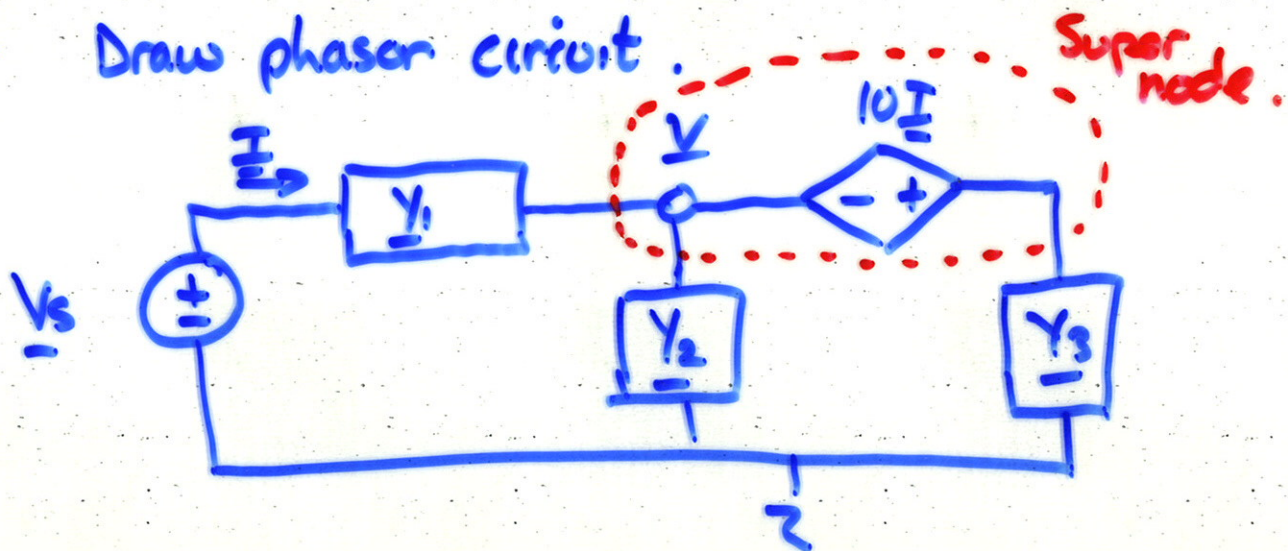
Step 2 - observe that $i(t) = \frac{V_s(t) - V(t)}{10}$

Step 3 - Calculate impedances

$$Z_L = j\omega h = j5$$

$$Z_C = \frac{1}{j\omega c} = -j10$$

Step 4 - Draw phasor circuit.



where $Y_1 = \frac{1}{R_1} = 0.1$

$$Y_2 = \frac{1}{R_2} + \frac{1}{Z_C} = 0.1(1+j)$$

$$Y_3 = \frac{1}{R_3 + Z_L} = 0.1(1-j)$$

KCL at supernode

$$Y_1(V - V_s) + Y_2V + Y_3(V + 10I) = 0$$

Dependent source $I = Y_1(V_s - V)$

Re-arrange

$$V = \frac{(Y_1 - 10Y_1Y_3)}{Y_1 + Y_2 + Y_3 - 10Y_1Y_3} \cdot V_s$$

Since $V_s = 10e^{j0}$

$$\Rightarrow V = \frac{10j}{2+j}$$

$$= \frac{10}{\sqrt{5}} e^{j63.4^\circ}$$

$$\Rightarrow v(t) = \frac{10}{\sqrt{5}} \cos(10t + 63.4^\circ)$$