EE 4CL4 – Control System Design

Solutions to Homework Assignment #5

1. Determine the open-loop transfer function of the system generating the root locus plot shown in Figure 1. (25 pts)





The open-loop poles of this system are at s = -1, -2 and -3. There is one finite open-loop zero at s = -4 and two implicit zeros at ∞ . Thus the open-loop transfer function of this system is of the form:

$$H(s) = \lambda \frac{s+4}{(s+1)(s+2)(s+3)},$$

where λ is the gain-parameter varied to created the root-locus plot.

2. Determine the PID controller parameters (for the *standard form*) for a plant with the nominal model:

$$G_o(s) = \frac{-s+2}{(s+2)^2}$$

using the Ziegler-Nichols oscillation method.

The closed-closed characteristic polynomial for this nominal plant model in a one-d.o.f. unity-feedback loop with a proportional controller is:

$$1 + K_p G_o(s) = (s+2)^2 + K_p(-s+2) = 0.$$

At the point of critical stability, $K_p = K_c$ and $s = j\omega_c$, such that:

$$(s+2)^{2} + K_{p}(-s+2) = (j\omega_{c}+2)^{2} + K_{c}(-j\omega_{c}+2) = 0$$

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(25 pts)

$$\Rightarrow K_c = \frac{-(j\omega_c + 2)^2}{(-j\omega_c + 2)} = \frac{6\omega_c^2 - 8 + j(\omega_c^3 - 12\omega_c)}{\omega_c^2 + 4}.$$

The critical gain $K_c \in \Re$, so the complex term in the equation above must equal zero, which gives:

$$\omega_c^3 - 12\omega_c = 0 \Longrightarrow \omega_c = 2\sqrt{3} \Longrightarrow P_c = \frac{2\pi}{\omega_c} = \frac{\pi}{\sqrt{3}}.$$

Substituting the value for ω_c into the equation for K_c yields:

$$K_{c} = \frac{6\omega_{c}^{2} - 8}{\omega_{c}^{2} + 4} = \frac{6 \cdot 12 - 8}{12 + 4} = 4.$$

From Table 6.1 of Goodwin et al., the PID parameters are then:

- $K_p = 0.6K_c = 2.4$, $T_r = 0.5P_c = 0.9069$, and $T_d = P_c/8 = 0.2267$.
- 3. Use the pole placement method to synthesize a controller C(s) for the nominal plant model:

$$G_o(s) = \frac{1}{(s+2)^2},$$

that produces the nominal closed-loop characteristic polynomial $A_{cl}(s) = (s^2 + 4s + 9)(s + 8)$, using MATLAB to solve the matrix equations. (25 pts)

The eliminant matrix for a nominal plant model with a maximum degree of 2 is:

$$\mathbf{M}_{\mathbf{e}} = \begin{bmatrix} a_2 & 0 & b_2 & 0 \\ a_1 & a_2 & b_1 & b_2 \\ a_0 & a_1 & b_0 & b_1 \\ 0 & a_0 & 0 & b_0 \end{bmatrix},$$

and the controller to be synthesized is of the form:

$$C(s) = \frac{p_1 s + p_0}{l_1 s + l_0}.$$

The pole-assignment matrix equation for this system is then:

$$\begin{bmatrix} 1 & 0 & 0 & 0 \\ 4 & 1 & 0 & 0 \\ 4 & 4 & 1 & 0 \\ 0 & 4 & 0 & 1 \end{bmatrix} \begin{bmatrix} l_1 \\ l_0 \\ p_1 \\ p_0 \end{bmatrix} = \begin{bmatrix} 1 \\ 12 \\ 41 \\ 72 \end{bmatrix} \Rightarrow \begin{bmatrix} l_1 \\ l_0 \\ p_1 \\ p_0 \end{bmatrix} = \begin{bmatrix} 1 \\ 8 \\ 5 \\ 40 \end{bmatrix},$$

giving the controller $C(s) = \frac{5s + 40}{s + 8}$.

4. Find suitable PID controller parameters (for the *standard form*) for a plant with the nominal model:

$$G_o(s) = \frac{10}{(s+1)(s+10)},$$
(1)

using the reaction curve method with:

- a. the Ziegler-Nichols parameters, and
- b. the Cohen-Coon parameters.

(25 pts)

The process reaction curve for this plant can be obtained by calculating the unit step response y(t) of the plant in open loop:

$$Y(s) = G_o(s)U(s) = \frac{10}{s(s+1)(s+10)}$$

$$\Rightarrow y(t) = \mathcal{L}^{-1}[Y(s)] = \mathcal{L}^{-1}\left[\frac{10}{s(s+1)(s+10)}\right] = 1 - \frac{10}{9}e^{-t} + \frac{1}{9}e^{-10t}.$$

The slope of y(t) is then:

$$\dot{y}(t) = \frac{10}{9} e^{-t} - \frac{10}{9} e^{-10t}$$
,

and the maximal slope can be found at the time *t* when the derivate of the slope is zero:

$$\ddot{y}(t) = -\frac{10}{9}e^{-t} + \frac{100}{9}e^{-10t} = 0$$

$$\Rightarrow -e^{-t} + 10e^{-10t} = 0$$

$$10e^{-10t} = e^{-t}$$

$$10e^{-10t}e^{t} = 1$$

$$e^{-9t} = 1/10$$

$$-9t = \log_{e}(1/10)$$

$$t = -\frac{\log_{e}(1/10)}{9} \approx 0.2558$$

The maximal slope is then:

$$\dot{y}(2.558) = \frac{10}{9}e^{-0.2558} - \frac{10}{9}e^{-2.558} = 0.7743,$$

at the point t = 0.2558, y(0.2558) = 0.1483, giving the maximum slope tangent:

$$m.s.t. = 0.7743(t - 0.2558) + 0.1483$$
.

The *m.s.t.* is equal to $y_0 = 0$ at time $t_1 = 0.0643$ and is equal to $y_\infty = 1$ at $t_2 = 1.3558$. The unit step $(u_0 = 0; u_\infty = 1)$ was applied at time $t_0 = 0$, giving the parameter model:

$$K_0 = \frac{y_{\infty} - y_0}{u_{\infty} - u_0} = 1; \qquad \tau_0 = t_1 - t_0 = 0.0643; \qquad v_0 = t_2 - t_1 = 1.2915.$$

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a. From Table 6.2 of Goodwin et al., the Ziegler-Nichols PID parameters are then:

$$K_{p} = \frac{1.2\nu_{0}}{K_{0}\tau_{0}} = 24.1026,$$

$$T_{r} = 2\tau_{0} = 0.1286, \text{ and}$$

$$T_{d} = 0.5\tau_{0} = 0.0321.$$

b. From Table 6.3 of Goodwin et al., the *Cohen-Coon* PID parameters are:

$$K_{p} = \frac{v_{0}}{K_{0}\tau_{0}} \left[\frac{4}{3} + \frac{\tau_{0}}{4v_{0}} \right] = 27.0307,$$

$$T_{r} = \frac{\tau_{0} [32v_{0} + 6\tau_{0}]}{13v_{0} + 8\tau_{0}} = 0.1550, \text{ and}$$

$$T_{d} = \frac{4\tau_{0}v_{0}}{11v_{0} + 2\tau_{0}} = 0.0232.$$