

# ELEC ENG 4CL4 – Control System Design

## Solutions to Homework Assignment #1

1. Consider a system that obeys the differential equation:

$$\frac{d^2x}{dt^2} + 2\frac{dx}{dt} + \cos x = 0.$$

a. Linearize this equation around the operating point  $x = \pi/4$ .

b. Derive a state-space representation of the linear equation found in part a. (25 pts)

a. The nonlinear term  $\cos x$  is linearized via the Taylor series approximation:

$$\begin{aligned} f(x) = \cos x &\approx \cos\left(\frac{\pi}{4}\right) + \left.\frac{df}{dx}\right|_{x=\frac{\pi}{4}}\left(x - \frac{\pi}{4}\right) = \cos\left(\frac{\pi}{4}\right) - \sin\left(\frac{\pi}{4}\right)\left(x - \frac{\pi}{4}\right) \\ &= \frac{1}{\sqrt{2}} + \frac{\pi}{4\sqrt{2}} - \frac{x}{\sqrt{2}}, \end{aligned}$$

giving the linear equation:

$$\frac{d^2x}{dt^2} + 2\frac{dx}{dt} - \frac{x}{\sqrt{2}} = -\frac{1}{\sqrt{2}} - \frac{\pi}{4\sqrt{2}}.$$

Note that an equivalent model in terms of  $\Delta x = x - \frac{\pi}{4}$  is:

$$\frac{d^2\Delta x}{dt^2} + 2\frac{d\Delta x}{dt} - \frac{\Delta x}{\sqrt{2}} = -\frac{1}{\sqrt{2}}.$$

b. To obtain a state-space representation of the linearized model, we define two state variables

$x_1 = x$  and  $x_2 = \frac{dx}{dt}$ , the input variable  $u = \frac{1}{\sqrt{2}} + \frac{\pi}{4\sqrt{2}}$ , and the output variable  $y = x$ , giving:

$$\begin{aligned} \frac{dx_1}{dt} &= x_2, \\ \frac{dx_2}{dt} &= \frac{x_1}{\sqrt{2}} - 2x_2 - u, \\ y &= x_1. \end{aligned}$$

Note that an equivalent model in terms of  $\Delta x = x - \frac{\pi}{4}$  and  $\Delta u = \frac{1}{\sqrt{2}}$  is:

$$\begin{aligned} \frac{d\Delta x_1}{dt} &= \Delta x_2, \\ \frac{d\Delta x_2}{dt} &= \frac{\Delta x_1}{\sqrt{2}} - 2\Delta x_2 - \Delta u, \\ \Delta y &= \Delta x_1. \end{aligned}$$

2. A two-phase (i.e., two-input) permanent magnet stepper motor can be described by the following set of differential equations:

$$\frac{d^2\theta}{dt^2} = -K_2 i_a \sin(K_3\theta) + K_2 i_b \cos(K_3\theta) - K_1 \frac{d\theta}{dt},$$

$$\frac{di_a}{dt} = -K_5 i_a + K_4 \frac{d\theta}{dt} \sin(K_3\theta) + K_6 v_a,$$

$$\frac{di_b}{dt} = -K_5 i_b - K_4 \frac{d\theta}{dt} \cos(K_3\theta) + K_6 v_b,$$

where  $\theta$  is angular displacement of the rotor,  $i_a$  and  $i_b$  are the currents in the two phases,  $v_a$  and  $v_b$  are the voltages applied the two phases (i.e., the inputs), and  $K_1, \dots, K_6$  are constants.

- a. Derive a state-space representation of this system.  
 b. Linearize the state-space model found in part a around the operating point  $\theta = \text{constant}$ .  
 (25 pts)

- a. We assign four state variables  $x_1 = \theta$ ,  $x_2 = \frac{d\theta}{dt}$ ,  $x_3 = i_a$  and  $x_4 = i_b$ , two input variables  $u_1 = v_a$  and  $u_2 = v_b$ , and one output variable  $y = \theta$ , giving the state-space representation:

$$\frac{dx_1}{dt} = f_1(x_1, x_2, x_3, x_4, u_1, u_2) = x_2$$

$$\frac{dx_2}{dt} = f_2(x_1, x_2, x_3, x_4, u_1, u_2) = -K_2 x_3 \sin(K_3 x_1) + K_2 x_4 \cos(K_3 x_1) - K_1 x_2,$$

$$\frac{dx_3}{dt} = f_3(x_1, x_2, x_3, x_4, u_1, u_2) = -K_5 x_3 + K_4 x_2 \sin(K_3 x_1) + K_6 u_1,$$

$$\frac{dx_4}{dt} = f_4(x_1, x_2, x_3, x_4, u_1, u_2) = -K_5 x_4 - K_4 x_2 \cos(K_3 x_1) + K_6 u_2,$$

$$y = g(x_1, x_2, x_3, x_4, u_1, u_2) = x_1.$$

- b. To simplify the linearization of this state-space model, we reformulate it in terms of the distances from the operating point  $(x_{1Q}, x_{2Q}, x_{3Q}, x_{4Q}, u_{1Q}, u_{2Q}, y_Q)$ :

$$\Delta x_1 = x_1 - x_{1Q},$$

$$\Delta x_2 = x_2 - x_{2Q},$$

$$\Delta x_3 = x_3 - x_{3Q},$$

$$\Delta x_4 = x_4 - x_{4Q},$$

$$\Delta u_1 = u_1 - u_{1Q},$$

$$\Delta u_2 = u_2 - u_{2Q},$$

$$\Delta y = y - y_Q.$$

Defining the vectors:

$$\Delta \mathbf{x} \triangleq [\Delta x_1 \quad \Delta x_2 \quad \Delta x_3 \quad \Delta x_4]^T, \quad \Delta \dot{\mathbf{x}} \triangleq [\Delta \dot{x}_1 \quad \Delta \dot{x}_2 \quad \Delta \dot{x}_3 \quad \Delta \dot{x}_4]^T \quad \text{and} \quad \Delta \mathbf{u} \triangleq [\Delta u_1 \quad \Delta u_2]^T,$$

The linearized state-space model is obtained from the Taylor series approximation in the matrix formulation:

$$\begin{aligned} \Delta \dot{\mathbf{x}} &= \begin{bmatrix} \left. \frac{\partial f_1}{\partial x_1} \right|_{\substack{x=x_Q \\ u=u_Q}} & \left. \frac{\partial f_1}{\partial x_2} \right|_{\substack{x=x_Q \\ u=u_Q}} & \left. \frac{\partial f_1}{\partial x_3} \right|_{\substack{x=x_Q \\ u=u_Q}} & \left. \frac{\partial f_1}{\partial x_4} \right|_{\substack{x=x_Q \\ u=u_Q}} \\ \left. \frac{\partial f_2}{\partial x_1} \right|_{\substack{x=x_Q \\ u=u_Q}} & \left. \frac{\partial f_2}{\partial x_2} \right|_{\substack{x=x_Q \\ u=u_Q}} & \left. \frac{\partial f_2}{\partial x_3} \right|_{\substack{x=x_Q \\ u=u_Q}} & \left. \frac{\partial f_2}{\partial x_4} \right|_{\substack{x=x_Q \\ u=u_Q}} \\ \left. \frac{\partial f_3}{\partial x_1} \right|_{\substack{x=x_Q \\ u=u_Q}} & \left. \frac{\partial f_3}{\partial x_2} \right|_{\substack{x=x_Q \\ u=u_Q}} & \left. \frac{\partial f_3}{\partial x_3} \right|_{\substack{x=x_Q \\ u=u_Q}} & \left. \frac{\partial f_3}{\partial x_4} \right|_{\substack{x=x_Q \\ u=u_Q}} \\ \left. \frac{\partial f_4}{\partial x_1} \right|_{\substack{x=x_Q \\ u=u_Q}} & \left. \frac{\partial f_4}{\partial x_2} \right|_{\substack{x=x_Q \\ u=u_Q}} & \left. \frac{\partial f_4}{\partial x_3} \right|_{\substack{x=x_Q \\ u=u_Q}} & \left. \frac{\partial f_4}{\partial x_4} \right|_{\substack{x=x_Q \\ u=u_Q}} \end{bmatrix} \Delta \mathbf{x} + \begin{bmatrix} \left. \frac{\partial f_1}{\partial u_1} \right|_{\substack{x=x_Q \\ u=u_Q}} & \left. \frac{\partial f_1}{\partial u_2} \right|_{\substack{x=x_Q \\ u=u_Q}} \\ \left. \frac{\partial f_2}{\partial u_1} \right|_{\substack{x=x_Q \\ u=u_Q}} & \left. \frac{\partial f_2}{\partial u_2} \right|_{\substack{x=x_Q \\ u=u_Q}} \\ \left. \frac{\partial f_3}{\partial u_1} \right|_{\substack{x=x_Q \\ u=u_Q}} & \left. \frac{\partial f_3}{\partial u_2} \right|_{\substack{x=x_Q \\ u=u_Q}} \\ \left. \frac{\partial f_4}{\partial u_1} \right|_{\substack{x=x_Q \\ u=u_Q}} & \left. \frac{\partial f_4}{\partial u_2} \right|_{\substack{x=x_Q \\ u=u_Q}} \end{bmatrix} \Delta \mathbf{u} \\ &= \begin{bmatrix} 0 & 1 & 0 & 0 \\ -K_2 K_3 x_{3Q} \cos(K_3 x_{1Q}) - K_2 K_3 x_{4Q} \sin(K_3 x_{1Q}) & -K_1 & -K_2 \sin(K_3 x_{1Q}) & K_2 \cos(K_3 x_{1Q}) \\ K_3 K_4 x_{2Q} \cos(K_3 x_{1Q}) & K_4 \sin(K_3 x_{1Q}) & -K_5 & 0 \\ K_3 K_4 x_{2Q} \sin(K_3 x_{1Q}) & -K_4 \cos(K_3 x_{1Q}) & 0 & -K_5 \end{bmatrix} \Delta \mathbf{x} \\ &\quad + \begin{bmatrix} 0 & 0 \\ 0 & 0 \\ K_6 & 0 \\ 0 & K_6 \end{bmatrix} \Delta \mathbf{u}, \\ \Delta y &= \begin{bmatrix} \left. \frac{\partial g}{\partial x_1} \right|_{\substack{x=x_Q \\ u=u_Q}} & \left. \frac{\partial g}{\partial x_2} \right|_{\substack{x=x_Q \\ u=u_Q}} & \left. \frac{\partial g}{\partial x_3} \right|_{\substack{x=x_Q \\ u=u_Q}} & \left. \frac{\partial g}{\partial x_4} \right|_{\substack{x=x_Q \\ u=u_Q}} \end{bmatrix} \Delta \mathbf{x} + \begin{bmatrix} \left. \frac{\partial g}{\partial u_1} \right|_{\substack{x=x_Q \\ u=u_Q}} & \left. \frac{\partial g}{\partial u_2} \right|_{\substack{x=x_Q \\ u=u_Q}} \end{bmatrix} \Delta \mathbf{u} \\ &= [1 \quad 0 \quad 0 \quad 0] \Delta \mathbf{x} + [0 \quad 0] \Delta \mathbf{u}. \end{aligned}$$

3. Given the following differential equation, solve for  $y(t)$  using the Laplace transform if all initial conditions are zero:

$$\frac{d^2y}{dt^2} + 12\frac{dy}{dt} + 32y = 32\mu(t),$$

where  $\mu(t)$  is the unit step function.

(25 pts)

The Laplace transform of this equation is:

$$\begin{aligned} s^2Y(s) + 12sY(s) + 32Y(s) &= 32\frac{1}{s} \\ \Rightarrow Y(s)\{s^2 + 12s + 32\} &= \frac{32}{s}. \end{aligned}$$

Solving for  $Y(s)$  gives:

$$\begin{aligned} Y(s) &= \frac{32}{s(s^2 + 12s + 32)} \\ &= \frac{32}{s(s+4)(s+8)} \\ &= \frac{1}{s} - \frac{2}{s+4} + \frac{1}{s+8}. \end{aligned}$$

Taking the inverse Laplace transform of  $Y(s)$  we obtain:

$$y(t) = 1 - 2e^{-4t} + e^{-8t}, \quad t \geq 0.$$

**4. A system has the transfer function:**

$$H(s) = \frac{10}{(s+7)(s+8)}.$$

- a. **Compute the system's response to the Dirac delta function (unit impulse)  $\delta_D(t)$ .**
- b. **Compute the system's response to the unit step function  $\mu(t)$ . (25 pts)**
- a. Because the Laplace transform of  $\delta_D(t)$  is 1, the impulse response of a system is the inverse Laplace transform of its transfer function:

$$\begin{aligned} h(t) &= \mathcal{L}^{-1}\{H(s)\} = \mathcal{L}^{-1}\left\{\frac{10}{(s+7)(s+8)}\right\} = \mathcal{L}^{-1}\left\{\frac{10}{s+7} - \frac{10}{s+8}\right\} \\ &= 10e^{-7t} - 10e^{-8t}, \quad t \geq 0. \end{aligned}$$

- b. The step response of this system is:

$$\begin{aligned} y(t) &= \mathcal{L}^{-1}\{H(s)U(s)\} = \mathcal{L}^{-1}\left\{\frac{10}{(s+7)(s+8)} \frac{1}{s}\right\} = \mathcal{L}^{-1}\left\{\frac{10}{56} - \frac{80}{56} \frac{1}{s+7} + \frac{5}{4} \frac{1}{s+8}\right\} \\ &= \frac{10}{56} - \frac{80}{56} e^{-7t} + \frac{5}{4} e^{-8t}, \quad t \geq 0. \end{aligned}$$