

Chapter 9 - Solved Problems

Solved Problem 9.1. Consider an internally stable feedback loop with

$$S_o(s) = \frac{s(s+2)}{s^2 + 4s + 2} \quad (1)$$

Determine whether Lemma 9.1 or Lemma 9.2 of the book applies to this system.

Solutions to Solved Problem 9.1

Solved Problem 9.2. Consider a linear plant having a nominal model having relative degree 2 and no poles in the open RHP.

Compute a lower bound for the sensitivity peak using Lemma 9.1 from the book and assuming that the following specifications are met:

- $|S_o(j\omega)| < \epsilon$ for $0 \leq \omega \leq \omega_\ell$
- $|T_o(j\omega)| < \frac{\epsilon\omega_h^2}{\omega^2}$ for $\omega \geq \omega_h > \omega_\ell$

where $\epsilon = 0.1$, $\omega_\ell = 2$ [rad/s] and $\omega_h = 3$ [rad/s].

Solutions to Solved Problem 9.2

Solved Problem 9.3. Consider a plant having a linear nominal model of relative degree 2 with no poles in the open RHP. Assume that the nominal model has a zero at $s = c = 0.5$.

Further assume that the following design specifications are met:

$$S.1 \quad |S_o(j\omega)| < \epsilon \frac{\omega^2}{\omega_\ell^2} \text{ for } 0 \leq \omega \leq \omega_\ell$$

$$S.2 \quad |T_o(j\omega)| < \frac{\epsilon\omega_h^2}{\omega^2} \text{ for } \omega \geq \omega_h > \omega_\ell$$

where $\epsilon = 0.1$, $\omega_\ell = 2$ [rad/s] and $\omega_h = 3$ [rad/s].

Find a lower bound for the complementary sensitivity peak.

Solutions to Solved Problem 9.3

Solved Problem 9.4. Consider a stable plant with NMP zeros located at $s = 1 \pm j\alpha$. Further assume that we require $|S_o(j\omega)| \leq \epsilon < 1$ for all $\omega \in [0, \omega_\ell]$. Using Lemma 9.5 from the book investigate the effect of the imaginary part, α of the NMP zeros on the lower bound for the sensitivity peak.

Solutions to Solved Problem 9.4

Solved Problem 9.5. Consider a linear system having the nominal model

$$G_o(s) = \frac{4(-s+4)}{(-s+2)(s+5)(s+2)} \quad (2)$$

Determine a lower limit for the sensitivity peak, assuming that we require that

$$|S_o(j\omega)| \leq 0.1 \quad \omega \in [0, 1] \quad (3)$$

$$|T_o(j\omega)| \leq 0.1 \quad \omega \in [8, \infty] \quad (4)$$

Solutions to Solved Problem 9.5

Chapter 9 - Solutions to Solved Problems

Solution 9.1. Since the feedback loop is internally stable, no unstable pole-zero cancellation can occur. Thus, all open loop poles in the RHP will appear in the sensitivity numerator. Since no zero of $S_o(s)$ is located in the open RHP, then Lemma 9.1 applies.

We can calculate the open loop transfer function $H_{ol}(s)$ as

$$H_{ol}(s) = S_o(s)^{-1} - 1 = \frac{2(s+1)}{s(s+2)} \quad (5)$$

Note that $H_{ol}(s)$ has relative degree, n_r , equal to 1 and that

$$\kappa = \lim_{s \rightarrow \infty} H_{ol}(s) = 2 \quad (6)$$

We can thus apply equation (9.2.3) from the book to obtain

$$\int_0^{\infty} \ln |S_o(j\omega)| d\omega = -\kappa \frac{\pi}{2} = -\pi \quad (7)$$

Solution 9.2. We use (7), splitting the integration interval as follows: $[0; \infty] = [0; \omega_\ell] \cup (\omega_\ell; \omega_h] \cup (\omega_h, \infty)$. Then

$$0 = \int_0^{\infty} \ln |S_o(j\omega)| d\omega = \int_0^{\omega_\ell} \ln |S_o(j\omega)| d\omega + \int_{\omega_\ell}^{\omega_h} \ln |S_o(j\omega)| d\omega + \int_{\omega_h}^{\infty} \ln |S_o(j\omega)| d\omega \quad (8)$$

$$< \int_0^{\omega_\ell} \ln \epsilon d\omega + \int_{\omega_\ell}^{\omega_h} \ln |S_{max}| d\omega + \int_{\omega_h}^{\infty} \ln \left| 1 + \frac{\epsilon \omega_h^2}{\omega^2} \right| d\omega \quad (9)$$

The last integral on the right is¹

$$\int_{\omega_h}^{\infty} \ln \left| 1 + \frac{\epsilon \omega_h^2}{\omega^2} \right| d\omega = \int_{\omega_h}^{\infty} \ln \left(1 + \frac{\epsilon \omega_h^2}{\omega^2} \right) d\omega = 2\omega_h \sqrt{\epsilon} \arctan \sqrt{\epsilon} - \omega_h \ln(1 + \epsilon) \triangleq \omega_h f(\epsilon) \leq \omega_h \epsilon \quad (10)$$

From the above equations we conclude that

$$\ln |S_{max}| > \frac{\omega_\ell |\ln \epsilon| - \omega_h \epsilon}{\omega_h - \omega_\ell} \approx 4.4 \quad (11)$$

Note that the lower limit of the sensitivity peak increases as $\epsilon \rightarrow 0$ and/or $\omega_\ell \rightarrow \omega_h$

Solution 9.3. To compute the lower bound for the complementary sensitivity peak, we can use Lemma 9.4 of the book (with $\tau = 0$) splitting the integration interval, as in Solved Problem 9.2.

Note that specification [S.1] implies that the open loop transfer function $H_{ol}(s) = G_o(s)C(s)$ has one or more poles at the origin. Thus equation (9.3.2) of the book is satisfied. Actually, S.1 would seem to imply that when k_v is computed from (9.3.4) the result is $k_v = \infty$.

¹To compute this integral use either a mathematical symbolic package such as MAPLE or MATHEMATICA, or the MATLAB symbolic toolbox, which is a MAPLE subset.

Thus

$$0 = \int_0^\infty \frac{\ln |T_o(j\omega)|}{\omega^2} d\omega = \int_0^{\omega_\ell} \frac{\ln |T_o(j\omega)|}{\omega^2} d\omega + \int_{\omega_\ell}^{\omega_h} \frac{\ln |T_o(j\omega)|}{\omega^2} d\omega + \int_{\omega_h}^\infty \frac{\ln |T_o(j\omega)|}{\omega^2} d\omega \quad (12)$$

$$< \underbrace{\int_0^{\omega_\ell} \frac{1}{\omega^2} \ln \left(1 + \epsilon \frac{\omega^2}{\omega_\ell^2} \right) d\omega}_{I_1} + \ln |T_{max}| \underbrace{\int_{\omega_\ell}^{\omega_h} \frac{1}{\omega^2} d\omega}_{I_2} + \underbrace{\int_{\omega_h}^\infty \frac{1}{\omega^2} \ln \left(\frac{\epsilon \omega_h^2}{\omega^2} \right) d\omega}_{I_3} \quad (13)$$

These integrals can be computed using, for instance, MAPLE. This yields

$$I_1 = \frac{-\ln(1 + \epsilon) + 2\sqrt{\epsilon} \arctan \sqrt{\epsilon}}{\omega_\ell} \quad (14)$$

$$I_2 = \frac{\omega_h - \omega_\ell}{\omega_h \omega_\ell} \quad (15)$$

$$I_3 = \frac{\ln(\epsilon) - 2}{\omega_h} \quad (16)$$

We can now use equation (9.3.11) of the book with $\tau = 0$, $M = 1$ and $k_v = \infty$. Hence

$$\ln |T_{max}| > \frac{\pi \omega_h \omega_\ell}{c(\omega_h - \omega_\ell)} + \frac{(\ln(1 + \epsilon) - 2\sqrt{\epsilon} \arctan \sqrt{\epsilon}) \omega_h}{\omega_h - \omega_\ell} + \frac{(2 + |\ln(\epsilon)|) \omega_\ell}{\omega_h - \omega_\ell} \approx 46.4 \quad (17)$$

It is interesting to note the factors which will cause this bound to grow. They are

- $\frac{\omega_\ell}{\omega_h}$ tends to one, i.e. the design specification demands a very sharp transition from the pass band to the stop band in $T_o(j\omega)$.
- The parameter ϵ tends to zero. This means that the design specification requires an excessively flat frequency response.
- The ratio $\frac{c}{\omega_\ell}$ tends to zero. This is in agreement with time domain analysis previously given in Chapter 8 of the book.

It is also interesting to note that the most significant contribution to this lower limit is the NMP zero. It accounts for roughly 82% of the limit.

Solution 9.4. We apply Lemma 9.5 from the book, with $M = 2$, $c_1 = 1 + j\alpha$ and $c_2 = 1 - j\alpha$, and the right hand side in equation (9.4.2) of the book equal to 0 (since there are no unstable poles). Thus, for c_1

$$\int_{-\infty}^\infty \ln |S_o(j\omega)| \frac{1}{1 + (\alpha - \omega)^2} d\omega = \int_{-\infty}^{-\omega_\ell} \ln |S_o(j\omega)| \frac{1}{1 + (\alpha - \omega)^2} d\omega + \int_{-\omega_\ell}^{\omega_\ell} \ln |S_o(j\omega)| \frac{1}{1 + (\alpha - \omega)^2} d\omega + \int_{\omega_\ell}^\infty \ln |S_o(j\omega)| \frac{1}{1 + (\alpha - \omega)^2} d\omega \quad (18)$$

We can now substitute $\ln |S_o(j\omega)|$ by its upper bound on every interval

$$\max |S_o(j\omega)| = \begin{cases} S_{max} & \omega \in [-\infty, -\omega_\ell] \\ \epsilon & \omega \in [-\omega_\ell, \omega_\ell] \\ S_{max} & \omega \in [\omega_\ell, \infty] \end{cases} \quad (19)$$

Then

$$\begin{aligned} \int_{-\infty}^{\infty} \ln |S_o(j\omega)| \frac{1}{1 + (\alpha - \omega)^2} d\omega &< \ln |S_{max}| \underbrace{\int_{-\infty}^{-\omega_\ell} \frac{1}{1 + (\alpha - \omega)^2} d\omega}_{I_1} + \\ &+ \ln(\epsilon) \underbrace{\int_{-\omega_\ell}^{\omega_\ell} \frac{1}{1 + (\alpha - \omega)^2} d\omega}_{I_2} + \ln |S_{max}| \underbrace{\int_{\omega_\ell}^{\infty} \frac{1}{1 + (\alpha - \omega)^2} d\omega}_{I_3} \end{aligned} \quad (20)$$

We can now compute the integrals I_1 , I_2 and I_3 using MAPLE. This yields

$$I_1 = \frac{\pi}{2} - \arctan(\alpha + \omega_\ell) \quad (21)$$

$$I_2 = \arctan(-\alpha + \omega_\ell) + \arctan(\alpha + \omega_\ell) \quad (22)$$

$$I_3 = \frac{\pi}{2} - \arctan(-\alpha + \omega_\ell) \quad (23)$$

Combining the above expressions we have that

$$\ln |S_{max}| > |\ln(\epsilon)| \frac{1}{f(\alpha) - 1} \quad (24)$$

where

$$f(\alpha) \triangleq \frac{\pi}{\arctan(-\alpha + \omega_\ell) + \arctan(\alpha + \omega_\ell)} > 1 \quad \text{for all finite } \omega_\ell \text{ and all finite } \alpha \quad (25)$$

In conclusion, the following observations are seen to apply:

- The function $f(\alpha)$ is an even function of α . Thus the above result is also valid for the NMP zero c_2 .
- The maximum lower bound occurs when $\alpha = 0$. This can be proved by differentiating $f(\alpha)$ with respect to α
- When $\alpha \gg \omega_\ell$, we have that the lower bound reaches a minimum, since then $f(\alpha)$ goes to infinity.

Solution 9.5. We use the Poisson formula as in Lemma 9.5 to compute a lower bound for S_{max} .

We first identify the following parameters and expressions:

$$\epsilon = 0.1; \quad \omega_\ell = 1; \quad \omega_h = 8 \quad (26)$$

$$B_p(s) = \frac{s-2}{s+2}; \quad B_z(s) = \frac{s-2}{s+2} \quad (27)$$

Then, using equation (9.4.20) of the book

$$\ln S_{max} > \frac{1}{\Omega(4, 8) - \Omega(4, 1)} [|\pi \ln |B_p(4)|| + |(\ln 0.1)\Omega(4, 1)| - (\pi - \Omega(4, 8)) \ln(1.1)] \quad (28)$$

The lower limit for S_{max} can then be computed using the MATLAB routine **smax** (provided on the CD-ROM in the book). This gives $S_{max} > 24.8473$.
