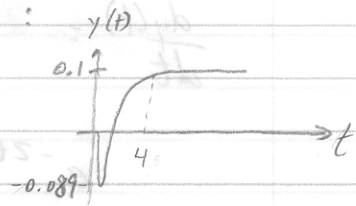


①

A system transfer function is given by:

$$H(s) = \frac{-s+1}{(s+2)(s+5)}$$



a) Will this system exhibit under or overshoot in response to a step? zero is slow and unstable \therefore NMP \rightarrow undershoot

b) Compute the magnitude and time instant of the under or overshoot, and the steady-state output.

$$\begin{aligned} Y(s) &= \frac{-s+1}{(s+2)(s+5)} \cdot \frac{1}{s} = \frac{A}{s} + \frac{B}{s+2} + \frac{C}{s+5} \\ &= \frac{A(s+2)(s+5) + Bs(s+5) + Cs(s+2)}{s(s+2)(s+5)} \\ &= \frac{As^2 + 7As + 10A + Bs^2 + 5Bs + Cs^2 + 2Cs}{s(s+2)(s+5)} \end{aligned}$$

$$A + B + C = 0$$

$$7A + 5B + 2C = -1$$

$$10A = 1$$

$$B + C = -\frac{1}{10}$$

$$5B + 2C = -\frac{17}{10}$$

$$A = \frac{1}{10}$$

$$B = -\frac{1}{10} - C$$

$$= -\frac{5}{10} = -\frac{1}{2}$$

$$-\frac{5}{10} - 5C + 2C = -\frac{17}{10}$$

$$-3C = -\frac{12}{10}$$

$$C = \frac{12}{30} = \frac{4}{10} = \frac{2}{5}$$

$$Y(s) = \frac{1/10}{s} - \frac{1/2}{s+2} + \frac{2/5}{s+5}$$

$$y(t) = 0.1 - 0.5e^{-2t} + 0.4e^{-5t}$$

$$\therefore y_{\infty} = 0.1$$

P.T.O. \rightarrow

Hilroy

$$\frac{dy(t)}{dt} = e^{-2t} - 2e^{-5t}$$

$$e^{-2t_u} = 2e^{-5t_u}$$

$$-2t_u = \ln 2 - 5t_u$$

$$3t_u = \ln 2$$

$$t_u = \frac{\ln 2}{3} = 0.2310$$

Assuming step occurs at $t=0$.

$$y(0.2310) = 0.1 - 0.5e^{-2(0.2310)} + 0.4e^{-5(0.2310)}$$

$$= -0.089$$

$$1 = A(1)$$

$$\frac{1}{10} = A$$

$$-1 = 25A + 5C$$

$$\frac{-1}{10} = 25 + 5C$$

$$0 = C + B + A$$

$$\frac{1}{10} = C + B$$

$$\frac{11}{10} = 25 + 5C$$

$$\frac{11}{10} - 25 = 5C$$

$$-\frac{1}{10} = C - B$$

$$\frac{1}{10} = B - C$$

$$\frac{11}{10} = 25 + 5C$$

$$\frac{11}{10} - 25 = 5C$$

$$\frac{11}{10} - \frac{250}{10} = 5C$$

$$\frac{-239}{10} = 5C$$

$$C = \frac{-239}{50}$$

$$\frac{1}{10} = C + B$$

$$\frac{1}{10} = \frac{-239}{50} + B$$

$$\frac{1}{10} + \frac{239}{50} = B$$

$$\frac{1}{10} + \frac{478}{100} = B$$

$$\frac{1}{10} + \frac{478}{100} = B$$

$$\frac{10}{100} + \frac{478}{100} = B$$

$$\frac{488}{100} = B$$

$$B = \frac{488}{100}$$

$$Y(s) = \frac{1}{2} \frac{1}{s+2} + \frac{1}{5} \frac{1}{s+5} - \frac{1}{2} \frac{1}{s} = (s)Y$$

$$y(t) = 0.1 - 0.2e^{-2t} + 0.1e^{-5t}$$

$$y(0) = 0.1$$

← 0.19

Handwritten note

②

Bode Plots → Nyquist Plot

Consider the following open-loop transfer function:

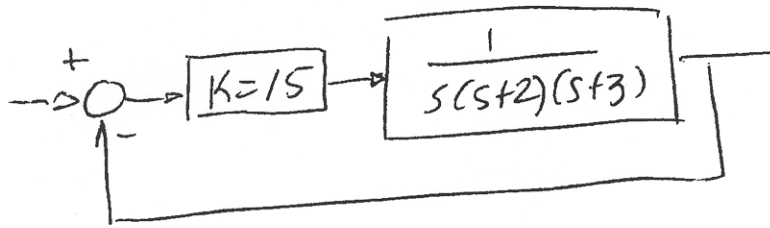
$$H(s) = \frac{15}{s(s+2)(s+3)}$$

To obtain the Bode plot, find the frequency response of $H(s)$: $s = j\omega$

| ω | $H(j\omega) = \frac{15}{j\omega(j\omega+2)(j\omega+3)}$ | |
|--------------|---|---|
| | MAG | PHASE |
| 0.001 | 2.5×10^3 | -90.05° |
| 0.01 | 250 | -90.48° |
| 0.1 | 25 | -94.7° |
| 0.2 | 12.4 | -99.5° |
| 0.5 | 4.8 | -113.5° |
| 1.0 | 2.1 | -135° |
| → 1.67 | 1.0 (0dB) | -159° → $180^\circ - 159^\circ = 21^\circ$ |
| 2.0 | 0.74 | -169° |
| → 2.45 (=√6) | 0.5 | -180° |
| 5.0 | 0.1 | -217° |
| 10.0 | 0.01 | -242° |
| 100 | 1.5×10^{-5} | -267.1° |
| 1000 | 1.5×10^{-8} | -269.7° |
| → ∞ | → 0 | → -270° |

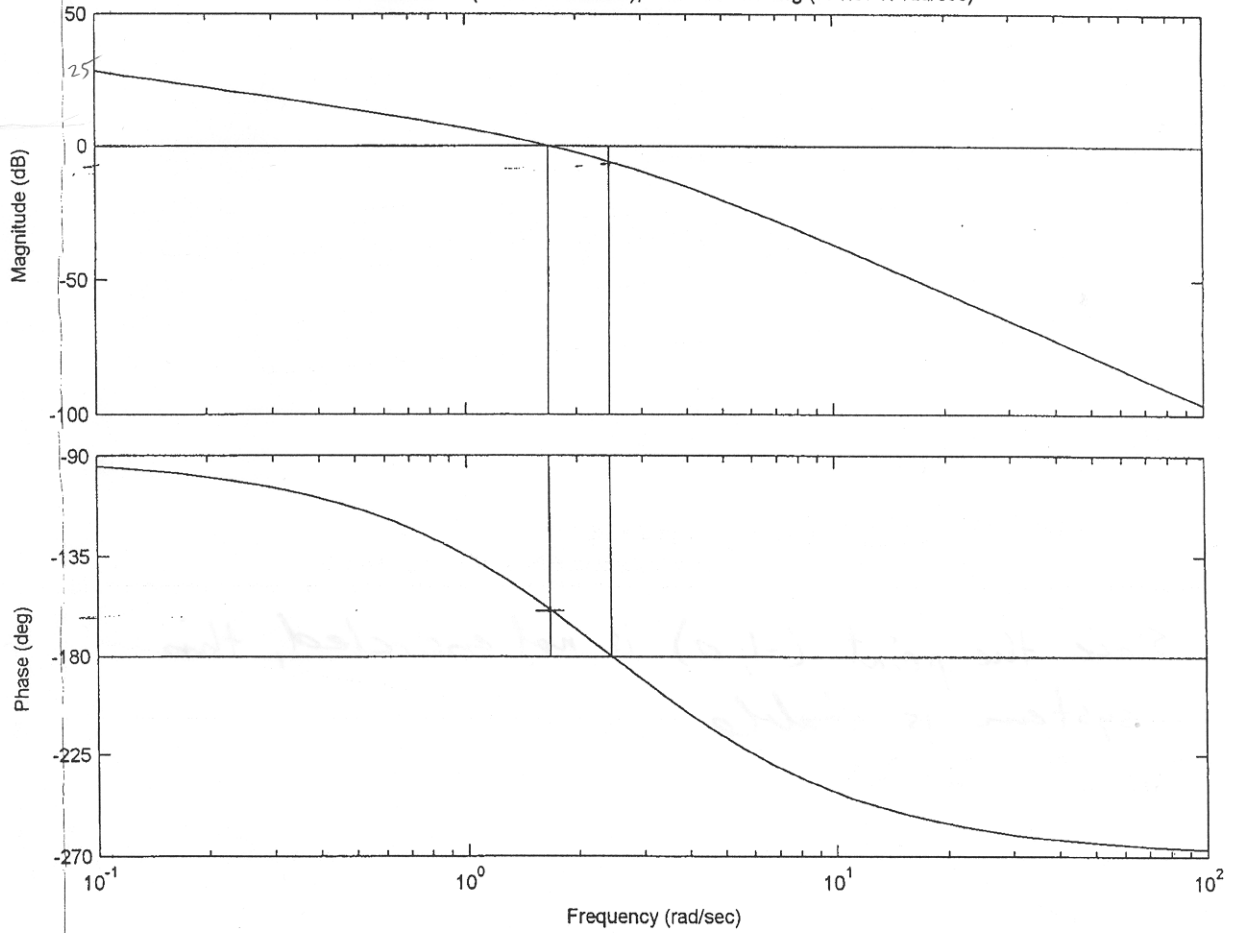
Which gives the Bode plot: (next page)

16057



Bode Diagram (using MARGIN command)

Gm = 6.0206 dB (at 2.4495 rad/sec), Pm = 20.907 deg (at 1.6741 rad/sec)

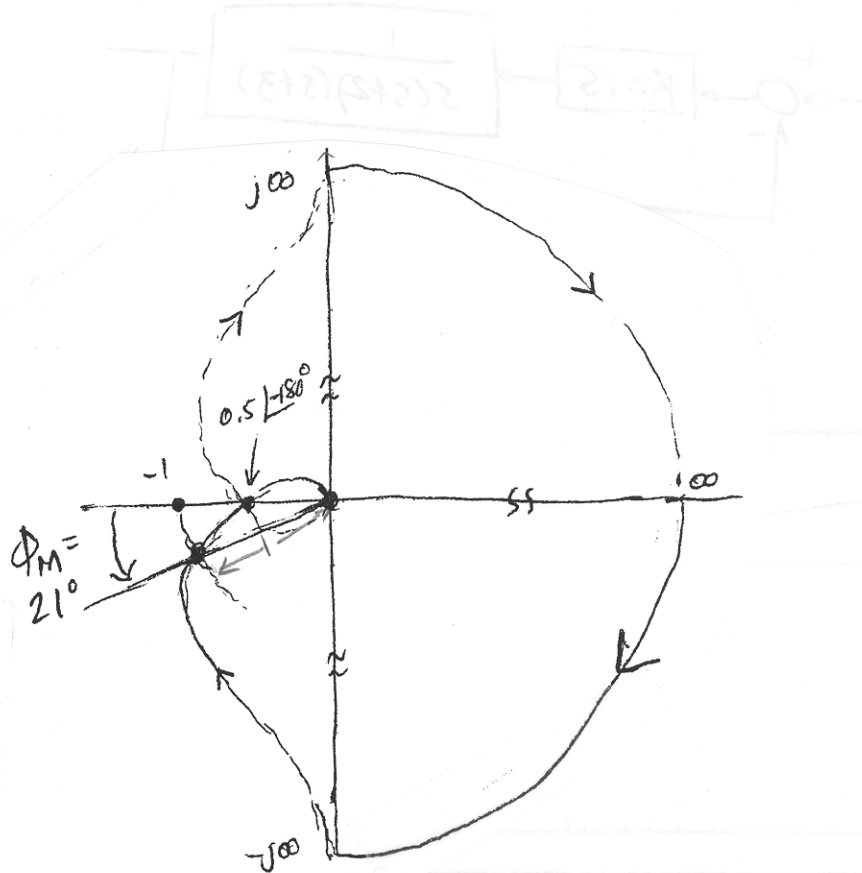


$$n = [0 \ 0 \ 0 \ 15]$$

$$d = [1 \ 5 \ 6 \ 15]$$

$$\text{margin}(n, d)$$

From the Bode plot, the Nyquist Plot can be drawn by combining the magnitudes and phases:



Since the point $(-1, 0)$ is not encircled, this system is stable.

$$n = [0 \ 0 \ 0 \ 1]$$

$$p = [1 \ 2 \ 1 \ 1]$$

$$h = [1 \ 2 \ 1 \ 1]$$

③ Example 5.2, p 128

Is this system stable?

$$G_0(s) = \frac{3}{(s+4)(-s+2)}$$

$$C(s) = \frac{-s+2}{s}$$

where $G_0(s) = \frac{B_0(s)}{A_0(s)}$

$$C(s) = \frac{P(s)}{L(s)}$$

for stability $A_0(s)L(s) + B_0(s)P(s) = 0$ has to have all roots in the left-half of the plane.

$G_0(s)$ has a positive root \rightarrow pole in positive plane
However the controller has a zero at the same location!

Nominal complementary sensitivity

$$T_0(s) = \frac{G_0(s)C(s)}{1+G_0(s)C(s)}$$

$$T_0(s) = \frac{3(-s+2)}{1+s(s+4)(-s+2)} = \frac{B_0(s)P(s)}{A_0(s)L(s) + B_0(s)P(s)}$$

Input-disturbance sensitivity:

$$S_{i0}(s) = \frac{G_0(s)}{1+G_0(s)C(s)} = \frac{B_0(s)L(s)}{A_0(s)L(s) + B_0(s)P(s)}$$

$$T_0(s) = \frac{3(-s+2)}{s(s+4)(-s+2) + 3(-s+2)} = \frac{3(-s+2)}{(-s+2)(s^2+4s+3)}$$

$$= \frac{3}{s^2+4s+3} = \frac{3}{(s+1)(s+3)} \text{ stable}$$

$$S_{i0}(s) = \frac{3s}{s(s+4)(-s+2) + 3(-s+2)} = \frac{3s}{(-s+2)(s^2+4s+3)}$$

so this is not stable!!

so Lemma 5.1 holds

$A_0(s)L(s) + B_0(s)P(s)$ should have roots in the negative half of the plane only.

(4)

Robust stability

Given:

$$G(s) = e^{-\tau s} F(s)$$

$$G_0(s) = \left(\frac{-\tau s + \tau k}{\tau s + \tau k} \right)^k F(s)$$

Let

$$C(s) = 1, \quad F(s) = \frac{1}{s(s+1)^2}, \quad \tau = \frac{1}{2}, \quad k = 1$$

Will this controller make the system robustly stable?

Need to satisfy $|T_0(j\omega)| |G_A(j\omega)| < 1$

$$G_A(s) = \frac{G(s) - G_0(s)}{G_0(s)} = e^{-\tau s} \left(\frac{-\tau s + \tau k}{\tau s + \tau k} \right)^k - 1$$

$$|G_A(j\omega)| = |e^{-j\tau\omega} - e^{-j2k\phi}|, \quad \text{where } \phi = \arctan\left(\frac{\omega\tau}{2k}\right)$$

$$T_0(s) = \frac{G_0(s) C(s)}{1 + G_0(s) C(s)} = \frac{B_0(s) P(s)}{A_0(s) L(s) + B_0(s) P(s)}$$

$$= \frac{-\frac{1}{2}s + 2}{(\frac{1}{2}s + 2)s(s+1)^2 + (\frac{1}{2}s + 2)} = \frac{-\frac{1}{2}s + 2}{\frac{1}{2}s^4 + 3s^3 + \frac{9}{2}s^2 + \frac{3}{2}s + 2}$$

Tutorial 4, Problem 4 (continued)

