# Achieving the Rate-Distortion Bound with Low-Density Generator Matrix Codes

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Abstract—It is shown that binary low-density generator matrix codes can achieve the rate-distortion bound of discrete memoryless sources with general distortion measure via multilevel quantization. A practical encoding scheme based on the surveypropagation algorithm is proposed. The effectiveness of the proposed scheme is verified through simulation.

*Index Terms*—Rate-distortion bound, linear code, low-density generator matrix, message-passing algorithm.

## I. INTRODUCTION

CHIEVING the rate-distortion bound of discrete memoryless sources has been one of the grand challenges in information theory. Though Shannon [1] has shown that the bound can be achieved asymptotically by using random block codes, the exponential encoding complexity makes such codes impractical. Coincidentally, another grand challenge in information theory is to design channel codes for discrete memoryless channels that have reasonable decoding complexity. To meet this challenge, research in channel coding has been almost exclusively focused on linear codes whose linear structure can be utilized to achieve efficient encoding and decoding. The focus on linear codes is further justified by the well-known result that linear codes can achieve the capacity of symmetric discrete memoryless channels under maximum likelihood decoding [2], [3]. The focused effort proves to be successful. Recent breakthroughs have shown that linear codes, e.g., Turbo codes and low-density parity check (LDPC) codes, used in conjunction with various messagepassing decoding algorithms, can indeed approach the capacity of discrete memoryless channels.

Recall that in lossy source coding the main design problem is to reduce the exponential encoding complexity resulting from the search for a reconstruction sequence in an exponentially large sequence space. Similarly, the main design problem in channel coding is to reduce the exponential decoding

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complexity resulting from the search for a codeword in an exponentially large space. The similar nature of the two design problems implies that the techniques developed for channel coding might be useful for lossy source coding as well. Indeed, lossy source codes like tree codes[4] and trellis codes [5] all have their origins in channel coding. Note that these two types of codes are known to be capable of achieving the rate-distortion bound. However, while these codes get the strengths from their counterparts in channel coding, they also inherit their counterparts' drawbacks. For example, saturating the rate-distortion bound with trellis codes requires taking the constraint length to infinity, which incurs exponential complexity.

In view of the recent success of LDPC codes and messagepassing algorithms for channel coding, it is natural to ask whether similar approaches are also applicable to lossy source coding. Indeed, motivated by this line of thinking, it has been proposed to use low-density generator matrix (LDGM) codes (which can be viewed as the dual of LDPC codes) and various message-passing algorithms for lossy source coding. As a consequence, the following questions naturally arise.

- Q1: Can LDGM codes achieve the rate-distortion bound of discrete memoryless sources with general distortion measure?
- Q2: If the answer to Q1 is affirmative, can LDGM codes, used in conjunction with certain message-passing algorithms, approach the rate-distortion bound of discrete memoryless sources with general distortion measure?

For the binary erasure quantization problem, Q1 and Q2 have been successfully answered in [6]. For uniform binary sources with Hamming distortion measure, Q1 has been addressed in [7]; it is shown that LDGM codes<sup>1</sup> (under optimal encoding) can saturate the rate-distortion bound in this case. Moreover, for uniform binary sources with Hamming distortion measure, good empirical answers to Q2 have been given in [8]–[11], where LDGM codes are used in conjunction with variants of message-passing algorithms for efficient encoding. Beyond the aforementioned examples, however, the answers to Q1 and Q2 are still unknown.

The purpose of this paper is to settle Q1 in the affirmative for general discrete memoryless sources and distortion measure. Specifically, we show that binary LDGM codes can

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<sup>&</sup>lt;sup>1</sup>More precisely, a hybrid LDPC-LDGM based construction is used in [7]. Nevertheless, the proof technique developed in [7] works for LDGM codes as well.

achieve the rate-distortion bound via multilevel quantization. Moreover, our simulation results indicate that binary LDGM codes, used in conjunction with the survey-propagation algorithm [12]–[14], provide a promising answer to Q2.

Note that linear lossy source codes should not be confused with linear lossy source encoding. Indeed, although we shall use linear codebooks, the encoding operation (i.e., the mapping from the source sequence to the codeword index) is intrinsically nonlinear. In fact, it is known that linear encoding generally cannot achieve the rate-distortion bound [15], albeit it suffices for lossless data compression [16]. It should also be emphasized that research on practical lossy source coding is by no means restricted to the aforementioned ones. For example, there has been a significant amount of work devoted to designing lossy source coding schemes based on the existing lossless data compression algorithms [17]–[19]. Moreover, several promising new directions [20]-[23] have emerged during the writing of this paper. In particular, it is shown in [21] that nonlinear sparse-graph codes can achieve the rate-distortion bound of discrete memoryless sources with general distortion measure.

The remainder of this paper is organized as follows. In Section II, we discuss the fundamental limit of linear codes in channel coding and lossy source coding. In Section III, we show that binary LDGM codes can saturate the ratedistortion bound via multilevel quantization. A practical encoding scheme based on the survey-propagation algorithm is proposed in Section IV. The effectiveness of the proposed scheme is verified through simulation. Finally, we close with some concluding remarks in Section V.

## II. LINEAR CODES FOR CHANNEL CODING AND LOSSY SOURCE CODING

Consider the lossy source coding problem for an i.i.d. process  $\{X_i\}_{i=1}^{\infty}$  with marginal probability distribution  $P_X$  on  $\mathcal{X}$ . Let  $d(\cdot, \cdot) : \mathcal{X} \times \hat{\mathcal{X}} \to [0, d_{\max}]$  be a distortion measure, where  $\hat{\mathcal{X}}$  is the reconstruction alphabet. Throughout this paper,  $\mathcal{X}$  and  $\hat{\mathcal{X}}$  are assumed to be finite. The rate-distortion function R(D) for source X with distortion measure  $d(\cdot, \cdot)$  is given by [24]

$$R(D) = \min_{P_{\hat{X}|X}} I(X; \hat{X}), \tag{1}$$

where the minimization is over all test channels  $P_{\hat{X}|X}$ :  $\mathcal{X} \to \hat{\mathcal{X}}$  subject to the constraint  $\mathbb{E}[d(X, \hat{X})] \leq D$ . The ratedistortion bound is known to be achievable with random block codes. However, the encoding complexity of such codes grows exponentially with the block length since there is no structure to exploit. As a consequence, it is of considerable interest to see whether structured codes, such as linear codes, suffice to achieve the rate-distortion bound.

First we shall give a definition of linear codes. A binary linear code  $C_n$  of length n and rate  $\frac{k}{n}$  is a k-dimensional linear subspace of the vector space  $\mathbb{F}_2^n$ , where  $\mathbb{F}_2$  is the binary field with elements 0 and 1. Moreover, we can write  $C_n = \{c^n : c^n = u^k G, u^k \in \mathbb{F}_2^k\}$ , where G, referred to as the generator matrix of  $C_n$ , is a  $k \times n$  matrix with rows being the basis codewords of  $C_n$ . To see whether linear codes are capable of achieving the rate-distortion bound, it is instructive to examine a similar problem in channel coding. Let  $P_{V|U} : \mathcal{U} \to \mathcal{V}$  be a discrete memoryless channel. For illustrative purpose, we shall temporarily assume  $\mathcal{U} = \{0, 1\}$ . It is well known that the capacity (i.e., the maximum reliable communication rate) of channel  $P_{V|U}$  is given by

$$C = \max_{P_U} I(U; V), \tag{2}$$

where the maximization is over all probability distributions on  $\mathcal{U}$ . However, the channel capacity is not always directly achievable using binary linear codes. Specifically, it has been shown [3] that the maximum reliable communication rate for binary linear codes over channel  $P_{V|U}$  is given by

$$C_L \triangleq I(U;V)|_{P_U=P^*},\tag{3}$$

where  $P^*$  is the uniform distribution on  $\mathcal{U}$ . Since the role of reconstruction variable  $\hat{X}$  in lossy source coding is similar to that of channel input U in channel coding, while the role of generic source variable X is similar to that of channel output V, it is natural to restrict  $\hat{X}$  to be uniformly distributed if linear codes are used for lossy source coding. Specifically, for  $\hat{X} = \{0, 1\}$ , one might conjecture that the minimum achievable rate in lossy source coding using binary linear codes is given by

$$R_L(D) \triangleq \min_{P_{\hat{X}|X}} I(X; \hat{X}), \tag{4}$$

where the minimization is over all test channels  $P_{\hat{X}|X}$ :  $\mathcal{X} \to \hat{\mathcal{X}}$  subject to the constraints  $\mathbb{E}[d(X, \hat{X})] \leq D$  and  $P_{\hat{X}}(0) = P_{\hat{X}}(1) = \frac{1}{2}$ ; in particular, if there is no test channel  $P_{\hat{X}|X}: \mathcal{X} \to \hat{\mathcal{X}}$  satisfying all the constraints, then  $R_L(D)$ is set to be equal to infinity, which essentially asserts that no binary linear code can meet the distortion constraint D. This conjecture turns out to be false as shown by the following example. Let  $\mathcal{X} = \mathcal{X} = \{0, 1\}, P_X(0) = p$  with  $p \in (0, \frac{1}{2}),$ and  $d(\cdot, \cdot)$  be the Hamming distortion measure. It is easy to verify that  $R_L(D) = \infty$  when  $D < \frac{1}{2}(1-2p)$ . However, since the linear span of every binary block code is a binary linear code, it is always possible to find a binary linear code to meet the distortion constraint D (even when  $D < \frac{1}{2}(1-2p)$ ). Therefore, this example indicates that in lossy source coding one might not need to restrict the output of the test channel to be uniformly distributed when linear codes are used.

To better understand the subtle difference between channel coding and lossy source coding in terms of the fundamental limit of linear codes, we shall have a close look at the reasoning behind (3). First we need a few definitions. For any sequence  $a^n = (a_1, a_2, \dots, a_n)$  with  $a_i \in \mathcal{A}$ ,  $i = 1, 2, \dots, n$ , the empirical distribution  $P_{a^n}$  of  $a^n$  is defined as  $(P_{a^n}(a))_{a \in \mathcal{A}}$  with  $P_{a^n}(a) = \frac{1}{n} \sum_{i=1}^n \mathbb{I}(a_i = a)$ , where  $\mathbb{I}(\cdot)$  is the indicator function. We define the empirical distributions of all its codewords. In particular, for a binary linear code  $C_n$  of length n, the empirical distribution  $P_{c^n}$  of a codeword  $c^n \in C_n$  is given by  $\frac{1}{n}(n - w(c^n), w(c^n))$ , where  $w(\cdot)$  is the weight function that counts the number of 1's in a sequence; moreover, the empirical distribution of  $C_n$  is given by  $\frac{1}{|C_n|} \sum_{c^n \in C_n} P_{c^n}$ . It is easy to verify that the empirical distribution of an arbitrary

binary linear code is the uniform distribution on  $\{0, 1\}$  unless the code contains coordinates in which every codeword is zero (such coordinates can be deleted as they carry no information). Now one can readily prove (3) by invoking Fano's inequality, the data processing inequality, and a convexity argument (cf. [25, Problem 5, pp. 112-113]). However, in contrast to the fact that each codeword in a channel code is transmitted with the same probability, the codewords in a lossy source code may not be used with the same frequency<sup>2</sup>. As a consequence, the empirical distribution of a code does not play much of a role in lossy source coding (see [26] for a more in-depth discussion); indeed, this is the reason why (4) should not be viewed as the fundamental limit of binary linear lossy source codes. On the other hand, the rate-distortion function (1) does impose certain constraints on the empirical distributions of individual codewords of a good lossy source code. Let (R, D) be an arbitrary point on the rate-distortion curve. Let  $P_{\hat{X}^*|X}$  be the optimal test channel associated with this ratedistortion point and  $P_{\hat{X}^*}$  be the output distribution induced by  $P_X$  and  $P_{\hat{X}^*|X}$ . It can be shown using the result in [27] that a good lossy source code, in the sense that its rate-distortion performance is very close to the given (R, D)point, must contain an exponentially non-negligible fraction of codewords whose empirical distributions are close to  $P_{\hat{X}^*}$ . Note that  $P_{\hat{X}^*}$  is in general not a uniform distribution except for certain special cases such as the uniform binary source with Hamming distortion measure. However, it is well known that the codeword empirical distributions of many commonly used binary random linear code ensembles concentrate (on the exponential scale) around the uniform distribution (see, e.g., [28]). Actually it can be shown under mild conditions that a binary linear code must have an exponentially non-negligible fraction of codewords with approximately the uniform empirical distribution. Although this does not exclude the possibility that the code also contains an exponentially non-negligible fraction of codewords with empirical distributions close to a non-uniform  $P_{\hat{X}^*}$ , it does indicate that such linear codes are difficult, if not impossible, to find<sup>3</sup>. As a consequence, if  $P_{\hat{x}*}$  associated with the given (R, D) on the rate-distortion curve is not a uniform distribution, then it is very hard to use linear codes to directly achieve this rate-distortion point due to the mismatch between the output distribution of the optimal test channel and the dominant codeword empirical distribution of linear codes<sup>4</sup>. To illustrate this point, we shall use the binary lossy source coding problem with Hamming distortion measure as an example. It can be seen from Fig. 1 that the performance of the scheme proposed in [9] (which is based on linear codes) degrades significantly when applied to a non-uniform source although it works well for the uniform



Fig. 1. Performance degradation for non-uniform sources.

source. This is because when the source is not uniform, the corresponding  $P_{\hat{X}*}$  is also not uniform. Therefore, the performance degradation is not solely due to the limitation of the encoding algorithm; it is also a consequence of the intrinsic property of linear codes.

### **III. MULTILEVEL QUANTIZATION**

We shall show that (4) is actually an upperbound on the minimum achievable rate of binary linear lossy source codes. However, in order to prove the sufficiency of binary linear codes for lossy source coding, one has to deal with the following scenarios:

- 1) The output distribution  $P_{\hat{X}*}$  induced by  $P_X$  and the optimal test channel  $P_{\hat{X}*|X}$  is non-uniform;
- 2) The reconstruction alphabet is non-binary.

It turns out that both scenarios can be handled by leveraging a method called multilevel quantization. The multilevel quantization method has a natural counterpart in channel coding called multilevel coding, which was first proposed by Gallager [30]. The idea can be illustrated by a simple example. Suppose the optimal input distribution  $P_U$  for a binary-input memoryless channel  $P_{V|U}$  of capacity C is given

 $<sup>^{2}</sup>$ The frequency that a codeword gets used depends on the encoding function and the source distribution.

<sup>&</sup>lt;sup>3</sup>Since it is easy to generate linear codes for which the empirical distributions of most of the codewords concentrate around the uniform distribution, one can construct linear codes which contain an exponentially non-negligible fraction of codewords whose empirical distributions are close to a certain nonuniform distribution by adding coordinates in which every codeword is zero. However, the resulting linear codes are not very useful due to the inefficient use of the available freedom.

<sup>&</sup>lt;sup>4</sup>A related discussion of mismatched codebooks in lossy source coding can be found in [29].

by  $(P_U(0), P_U(1)) = (\frac{1}{4}, \frac{3}{4})$ . Now construct a random variable  $\tilde{U}$  that is uniformly distributed over  $\mathbb{F}_2^2$ . Define a deterministic mapping  $U = f(\tilde{U})$  such that U = 0 if  $\tilde{U} = (0,0)$  and U = 1 otherwise. Clearly, the induced distribution of U is exactly the capacity-achieving input distribution, and we have  $I(\tilde{U}; V) = I(U; V) = C$ . Here the essential idea is that, by incorporating a deterministic mapping, one can convert a channel with non-uniform capacity-achieving input distribution to a channel of the same capacity whose optimal input distribution is uniform. In general, we can approximate an arbitrary input distribution by a uniform distribution over sufficiently large alphabet.

Now we proceed to adopt this idea in lossy source coding. Specifically, we shall approximate the output distribution  $P_{\hat{X}^*}$  by a uniform distribution over  $\mathbb{F}_2^m$  through a deterministic mapping. The rationale underlying such an approximation is the following. Loosely speaking, it can be assumed that the value of a binary linear code at any m coordinates is approximately uniformly distributed over  $\mathbb{F}_2^m$  (which will be justified later). Therefore, by mapping every m coordinates into a reconstruction symbol through a certain function, one can convert a binary linear code into a lossy source code whose dominant codeword empirical distribution matches  $P_{\hat{X}^*}$ . It is clear that in order to do so, the length of the binary linear code should be m times that of the length of the source sequence.

In the rest of this section we shall provide a rigorous treatment of this multilevel quantization method. For each positive integer m, we define

$$R_{L,m}(D) = \min_{P_{\tilde{X}|X}} I(X; \tilde{X})$$

where the minimization is over all  $P_{\tilde{X}|X} : \mathcal{X} \to \mathbb{F}_2^m$  such that the output distribution  $P_{\tilde{X}}$  induced by  $P_X$  and  $P_{\tilde{X}|X}$  is uniform over  $\mathbb{F}_2^m$ , and there exists a function  $f : \mathbb{F}_2^m \to \hat{\mathcal{X}}$  satisfying the distortion constraint  $\mathbb{E}[d(X, f(\tilde{X}))] \leq D$ . We let  $R_{L,m}(D) = \infty$  if there does not exist any  $P_{\tilde{X}|X}$  satisfying the constraints. Equivalently, we can define

$$R_{L,m}(D) = \min_{P_{\hat{X}|X}} I(X; \hat{X}),$$

where the minimization is over all  $P_{\hat{X}|X} : \mathcal{X} \to \hat{\mathcal{X}}$  such that the output distribution  $P_{\hat{X}}$  induced by  $P_X$  and  $P_{\hat{X}|X}$  is in  $\mathcal{P}_m$ , and the distortion constraint  $\mathbb{E}[d(X, \hat{X})] \leq D$  is satisfied. Here  $\mathcal{P}_m$  is the set of probability distributions  $P_{\hat{X}}$  over  $\hat{\mathcal{X}}$  satisfying  $2^m P_{\hat{X}}(\hat{x}) \in \mathbb{Z}$  for all  $\hat{x} \in \hat{\mathcal{X}}$ .

Theorem 1:  $\lim_{m\to\infty} R_{L,m}(D) = R(D)$  for  $D > \sum_{x\in\mathcal{X}} P_X(x) \min_{\hat{x}\in\hat{\mathcal{X}}} d(x,\hat{x}).$ 

*Proof:* See Appendix A.

Now we proceed to show that  $R_{L,m}(D)$  is achievable using binary LDGM codes. A binary LDGM code is a binary linear code associated with a sparse generator matrix (i.e, a generator matrix in which the number of 1's is significantly less than the number of 0's). We shall construct a binary random LDGM code ensemble by specifying a random generator matrix **G**. Let **G** be a  $k \times mn$  matrix with entries selected independently from  $\mathbb{F}_2$  using a Bernoulli distribution with parameter  $p_n$  (i.e.,  $Ber(p_n)$ ). We shall choose  $p_n$  such that  $p_n \to 0$  and  $np_n \to \infty$  as  $n \to \infty$ . Note that for such  $p_n$ , the realization of **G** is a sparse generator matrix with high probability as  $n \to \infty$ . Define  $Y_i(u^k) = u^k \mathbf{g}_i, u^k \in \mathbb{F}_2^k$ , where  $\mathbf{g}_i$  is the *i*-th column of  $\mathbf{G}$ ,  $i = 1, 2, \cdots, mn$ . Partition  $\{1, 2, \cdots, mn\}$  into *n* subsets  $\mathcal{N}_1, \mathcal{N}_2, \cdots, \mathcal{N}_n$ , each of size *m*. We assume the elements in  $\mathcal{N}_i, i = 1, 2, \cdots, n$ , are ordered. For each  $\mathcal{N}_i$  (say,  $\mathcal{N}_i = (n_1, \cdots, n_m)$ ), we define  $\tilde{X}_i(u^k) = (Y_{n_1}(u^k), \cdots, Y_{n_m}(u^k))$ . For any deterministic mapping  $f : \mathbb{F}_2^m \to \hat{\mathcal{X}}$ , let  $\mathcal{C}(\mathbf{G}, f) = \{\hat{X}^n(u^k) : u^k \in \mathbb{F}_2^k\}$ , where  $\hat{X}^n(u^k) = (f(\tilde{X}_1(u^k)), f(\tilde{X}_2(u^k)), \cdots, f(\tilde{X}_n(u^k)))$ . Define

$$R(n,k) = \frac{k}{n},$$
  
$$D(n,k,m,f) = \mathbb{E}[d_n(X^n, \mathcal{C}(\mathbf{G}, f))],$$

where

$$d_n(x^n, \mathcal{C}(\mathbf{G}, f)) = \min_{u^k \in \mathbb{F}_2^k} \frac{1}{n} \sum_{i=1}^n d(x_i, \hat{X}_i(u^k)), \quad x^n \in \mathcal{X}^n.$$

Note that R(n,k) is not the rate of the linear code generated by **G**; instead, it should be interpreted as the rate<sup>5</sup> of the lossy source code  $C(\mathbf{G}, f)$  induced by **G** and  $f(\cdot)$ .

Now we are ready to state the main result of this section. Theorem 2: Let  $P_{\tilde{X}|X}: \mathcal{X} \to \mathbb{F}_2^m$  be a transition probability distribution such that the output distribution  $P_{\tilde{X}}$  induced by  $P_X$  and  $P_{\tilde{X}|X}$  is uniform over  $\mathbb{F}_2^m$ . Let  $f(\cdot)$  be a deterministic mapping from  $\mathbb{F}_2^m$  to  $\hat{\mathcal{X}}$ . For any  $\epsilon > 0$ , we have

$$\begin{split} R(n,k) &\leq I(X;X) + \epsilon, \\ D(n,k,m,f) &\leq \mathbb{E}[d(X,f(\tilde{X}))] + \epsilon \end{split}$$

for some sufficiently large n and k.

Remark: In view of the definition of  $R_{L,m}(D)$  and Theorem 1, one can readily show that binary LDGM codes can achieve the rate-distortion bound via multilevel quantization by optimizing  $(P_{\tilde{X}|X}, f(\cdot))$  and letting  $m \to \infty$ .

*Proof:* See Appendix B.

It has been shown recently [31] that for uniform binary sources with Hamming distortion measure, a necessary condition for achieving the rate-distortion bound using LDGM codes is that the generator matrix contains rows with the number of 1's growing unboundedly as  $n \to \infty$  (see [32] for a weak version of this result). Note that in our construction, the condition  $np_n \to \infty$  can be interpreted as the requirement that the number of 1's in each row of the generator matrix grows unboundedly with n; therefore, our result indicates that the aforementioned necessary condition is essentially sufficient.

#### **IV. A PRACTICAL ENCODING SCHEME**

In this section, we shall propose a low-complexity encoding scheme based on the survey-propagation algorithm.

Let G be a  $k \times mn$  low-density generator matrix. It is well known that one can represent G by a factor graph with k variable nodes and mn check nodes (see Fig. 2). Label these variable nodes and check nodes by  $\{V_1, V_2, \dots, V_k\}$ and  $\{C_1, C_2, \dots, C_{mn}\}$ , respectively. Let there be an edge connecting  $V_i$  and  $C_j$  if and only if the (i, j)-entry of G is one. Let  $\mathcal{A}_k$  be the set of indices of check nodes connected to

<sup>&</sup>lt;sup>5</sup>Strictly speaking, R(n, k) is an upper bound on the rate of  $\mathcal{C}(\mathbf{G}, f)$  since it is possible that  $\hat{X}^n(u^k) = \hat{X}^n(v^k)$  for some  $u^k \neq v^k$ .



Fig. 2. Factor graph of an LDGM code.

 $V_i$  and  $\mathcal{B}_i$  be the set of indices of variable nodes connected to  $C_i$ . Each variable node  $V_i$  is associated with its corresponding information bit  $u_i$ . The value of  $C_j$  can be computed using the values of variable nodes in  $\mathcal{B}_i$  through modulo-2 addition.

We further introduce n source nodes and n network nodes, denoted by  $\{S_1, S_2, \dots, S_n\}$  and  $\{N_1, N_2, \dots, N_n\}$ , respectively. Each source node  $S_l$  is connected with its corresponding network node  $N_l$ ; moreover, we let  $S_l$  and  $N_l$  be associated with source symbol  $x_l$  and its corresponding reconstruction symbol  $\hat{x}_l$ , respectively.

We partition  $\{C_1, C_2, \cdots, C_{mn}\}$  into *n* disjoint subsets, each containing m check nodes. Each subset is connected to a unique network node. Specifically, we denote the index of the network node connected with  $C_i$  as  $\mu(j)$  and denote the set of indices of check nodes connected to  $N_l$  as  $\mathcal{N}_l$ . Label each element in  $\mathcal{N}_l$  with a unique number in  $\{1, \dots, m\}$ . Note that such a labeling induces a mapping  $\nu(\cdot)$  from  $\{1, 2, \cdots, mn\}$ to  $\{1, \dots, m\}$ . The value of  $N_l$  (i.e.,  $\hat{x}_l$ ) is determined by the check nodes in  $\mathcal{N}_l$  through a mapping  $f: \mathbb{F}_2^m \to \hat{\mathcal{X}}$ .

Our goal is to find an efficient way to determine the values of the variable nodes so that the distortion between the resulting reconstruction sequence  $\hat{x}^n$  and the given source sequence  $x^n$  is made reasonably small.

To this end, we propose a message-passing algorithm tailored to the aforementioned factor graph. Specifically, messages are passed among different types of nodes with each message representing a probability distribution. They are specified as follows.

- From  $V_i$  to  $C_j$ ,  $j \in \mathcal{A}_i$ :  $M^0_{V_i \to C_j}$ ,  $M^1_{V_i \to C_j}$ , and  $M^*_{V_i \to C_j}$ . From  $C_j$  to  $V_i$ ,  $i \in \mathcal{B}_j$ :  $M^0_{C_j \to V_i}$ ,  $M^1_{C_j \to V_i}$ , and  $M^*_{C_j \to V_i}$ . From  $C_j$  to  $N_l$ ,  $l = \mu(j)$ :  $M^0_{C_j \to N_l}$ ,  $M^1_{C_j \to N_l}$ , and  $M^*_{C_i \to N_l}$ .
- From  $S_l$  to  $N_l$ :  $M_{S_l \to N_l}^x$ ,  $x \in \mathcal{X}$ . From  $N_l$  to  $C_j$ ,  $j \in \mathcal{N}_l$ :  $M_{N_l \to C_j}^0$  and  $M_{N_l \to C_j}^1$ .

Remark: The introduction of the free state \* is one of the main features of the survey-propagation algorithm.

Here is a description of the message-passing routine (also see Fig. 3).

- 1) Each check node sends messages to the adjacent variable nodes (cf. (6)) and network node (cf. (7) and (8)). The initial message is set to be (0.5, 0.5, 0).
- 2) Each variable node computes the new messages based on those from the adjacent check nodes, then sends the messages back to the adjacent check nodes (cf. (5)). Similarly, each network node computes the new messages based on those from the adjacent check nodes

and source node (cf. (9)), then sends the messages back to the adjacent check nodes (cf. (10)).

- 3) Each check node computes the new messages based on those from the adjacent variable nodes and network node. If the messages converge at all check nodes or the maximum number of iterations (e.g., 100) is reached, then go to 4), otherwise go to 1).
- 4) Each variable node (say,  $V_i$ ) computes  $(M_{V_i}^0, M_{V_i}^1, M_{V_i}^*)$ , where

$$M_{V_{i}}^{0} = \prod_{j \in \mathcal{A}_{i}} (1 - M_{C_{j} \to V_{i}}^{1}) - \prod_{j \in \mathcal{A}_{i}} (1 - M_{C_{j} \to V_{i}}^{1} - M_{C_{j} \to V_{i}}^{0}), M_{V_{i}}^{1} = \prod_{j \in \mathcal{A}_{i}} (1 - M_{C_{j} \to V_{i}}^{0}) - \prod_{j \in \mathcal{A}_{i}} (1 - M_{C_{j} \to V_{i}}^{1} - M_{C_{j} \to V_{i}}^{0}), M_{V_{i}}^{*} = \prod_{j \in \mathcal{A}_{i}} (1 - M_{C_{j} \to V_{i}}^{1} - M_{C_{j} \to V_{i}}^{0}).$$

Find those variable nodes whose bias value (i.e.,  $|M_{V_k}^0 M_{V_{i}}^{1}$  ) is greater than a certain threshold and fix those variables accordingly. If there is no variable node whose bias value is greater than the threshold, then we fix the value of the variable node with the maximum bias value.

- 5) Remove those fixed variable nodes from the factor graph. If all the variable nodes are fixed, then go to 6), otherwise go to 1).
- 6) Calculate the value of each check node through modulo-2 addition of the values of its adjacent variable nodes. Determine the value of each network node using the values of its adjacent check nodes through the deterministic mapping  $f(\cdot)$ .

Remark: All the messages need to be normalized.

Now we proceed to apply the proposed message-passing algorithm to various sources and distortion measures to verify its effectiveness.

## A. Non-Uniform Binary Source With Hamming Distortion Measure

Let  $\mathcal{X} = \hat{\mathcal{X}} = \{0, 1\}$  and  $d(\cdot, \cdot)$  be the Hamming distortion measure, i.e.,  $d(x, \hat{x}) = 0$  if  $x = \hat{x}$  and  $d(x, \hat{x}) = 1$  otherwise. Consider a binary source X with  $P_X(0) = 0.25$ . Let  $P_{\hat{X}^*}$  be the output distribution induced by  $P_X$  and the optimal test channel  $P_{\hat{X}^*|X}$ . Note that  $P_{\hat{X}^*}$  is rate-dependent. Let m =4 and  $\tilde{X}$  be uniformly distributed over  $\mathbb{F}_2^4$ . A deterministic mapping  $f : \mathbb{F}_2^4 \to \mathcal{X}$  is chosen so that  $P_{f(\tilde{X})}$  is a good approximation of  $P_{\hat{X}^*}$ . Table I lists the optimal  $P_{f(\tilde{X})}$  for different rates. Note that although the optimal  $P_{f(\tilde{X})}$  for a given rate is in general unique, the deterministic mapping  $f(\cdot)$ that induces this distribution is clearly not unique.

The message-passing equations are given in Fig. 3. In (9), we choose  $\alpha_{x_l,x'} = 1$  if  $x' = x_l$  and  $\alpha_{x_l,x'} = \alpha$  otherwise. This choice is tailored to the Hamming distortion measure.

Variable node to check node

$$M^{0}_{V_{i} \to C_{j}} = \prod_{q \in \mathcal{A}_{i} \setminus \{j\}} (1 - M^{1}_{C_{q} \to V_{i}}) - \prod_{q \in \mathcal{A}_{i} \setminus \{j\}} (1 - M^{1}_{C_{q} \to V_{i}} - M^{0}_{C_{q} \to V_{i}})$$

$$M^{1}_{V_{i} \to C_{j}} = \prod_{q \in \mathcal{A}_{i} \setminus \{j\}} (1 - M^{0}_{C_{q} \to V_{i}}) - \prod_{q \in \mathcal{A}_{i} \setminus \{j\}} (1 - M^{1}_{C_{q} \to V_{i}} - M^{0}_{C_{q} \to V_{i}})$$

$$M^{*}_{V_{i} \to C_{j}} = \prod_{q \in \mathcal{A}_{i} \setminus \{j\}} (1 - M^{1}_{C_{q} \to V_{i}} - M^{0}_{C_{q} \to V_{i}})$$
(5)

Check node to variable node

$$\begin{split} M^{0}_{C_{j} \to V_{i}} &= \frac{1}{2} (M^{0}_{N_{\mu}(j) \to C_{j}} + M^{1}_{N_{\mu}(j) \to C_{j}}) \prod_{q \in \mathcal{B}_{j} \setminus \{i\}} (M^{0}_{V_{q} \to C_{j}} + M^{1}_{V_{q} \to C_{j}}) \\ &+ \frac{1}{2} (M^{0}_{N_{\mu}(j) \to C_{j}} - M^{1}_{N_{\mu}(j) \to C_{j}}) \prod_{q \in \mathcal{B}_{j} \setminus \{i\}} (M^{0}_{V_{q} \to C_{j}} - M^{1}_{V_{q} \to C_{j}}) \\ M^{1}_{C_{j} \to V_{i}} &= \frac{1}{2} (M^{0}_{N_{\mu}(j) \to C_{j}} + M^{1}_{N_{\mu}(j) \to C_{j}}) \prod_{q \in \mathcal{B}_{j} \setminus \{i\}} (M^{0}_{V_{q} \to C_{j}} + M^{1}_{V_{q} \to C_{j}}) \\ &- \frac{1}{2} (M^{0}_{N_{\mu}(j) \to C_{j}} - M^{1}_{N_{\mu}(j) \to C_{j}}) \prod_{q \in \mathcal{B}_{j} \setminus \{i\}} (M^{0}_{V_{q} \to C_{j}} - M^{1}_{V_{q} \to C_{j}}) \\ M^{*}_{C_{j} \to V_{i}} &= 1 - M^{0}_{C_{j} \to V_{i}} - M^{1}_{C_{j} \to V_{i}} \end{split}$$

$$(6)$$

Check node to network node

$$\begin{split} M^{0}_{C_{j} \to N_{l}} &= \frac{1}{2} \prod_{q \in \mathcal{B}_{j}} (M^{0}_{V_{q} \to C_{j}} + M^{1}_{V_{q} \to C_{j}}) + \frac{1}{2} \prod_{q \in \mathcal{B}_{j}} (M^{0}_{V_{q} \to C_{j}} - M^{1}_{V_{q} \to C_{j}}) \\ M^{1}_{C_{j} \to N_{l}} &= \frac{1}{2} \prod_{q \in \mathcal{B}_{j}} (M^{0}_{V_{q} \to C_{j}} + M^{1}_{V_{q} \to C_{j}}) - \frac{1}{2} \prod_{q \in \mathcal{B}_{j}} (M^{0}_{V_{q} \to C_{j}} - M^{1}_{V_{q} \to C_{j}}) \\ M^{*}_{C_{j} \to N_{l}} &= 1 - M^{0}_{C_{j} \to N_{l}} - M^{1}_{C_{j} \to N_{l}} \end{split}$$

$$(7)$$

\*If all  $V_q \in \mathcal{B}_j$  are fixed, then

$$M_{C_j \to N_l}^0 = \frac{(1 - c_j) \exp(\delta) + c_j \exp(-\delta)}{\exp(\delta) + \exp(-\delta) + W_{sou}}$$

$$M_{C_j \to N_l}^1 = \frac{c_j \exp(\delta) + (1 - c_j) \exp(-\delta)}{\exp(\delta) + \exp(-\delta) + W_{sou}}$$

$$M_{C_j \to N_l}^* = \frac{W_{sou}}{\exp(\delta) + \exp(-\delta) + W_{sou}}$$
(8)

Source node to network node

$$M_{S_l \to N_l}^x = \frac{\alpha_{x_l, x_l} \exp(\gamma)}{\alpha_{x_l, x_l} \exp(\gamma) + \sum_{x' \in \mathcal{X}: x' \neq x_l} \alpha_{x_l, x'} \exp(-\gamma)} \qquad x = x_l$$

$$M_{S_l \to N_l}^x = \frac{\alpha_{x_l, x} \exp(-\gamma)}{\alpha_{x_l, x_l} \exp(\gamma) + \sum_{x' \in \mathcal{X}: x' \neq x_l} \alpha_{x_l, x'} \exp(-\gamma)} \qquad x \neq x_l$$
(9)

Network node to check node

$$M_{N_{l} \to C_{j}}^{0} = \sum_{(k_{1}, \cdots, k_{m}): k_{\nu(j)} = 0} (M_{S_{l} \to N_{l}}^{r} \prod_{q \in \mathcal{N}_{l} \setminus \{j\}} M_{C_{q} \to N_{l}}^{k_{\nu(q)}}) \quad r = f(k_{1}, \cdots, k_{m})$$

$$M_{N_{l} \to C_{j}}^{1} = \sum_{(k_{1}, \cdots, k_{m}): k_{\nu(j)} = 1} (M_{S_{l} \to N_{l}}^{r} \prod_{q \in \mathcal{N}_{l} \setminus \{j\}} M_{C_{q} \to N_{l}}^{k_{\nu(q)}}) \quad r = f(k_{1}, \cdots, k_{m})$$
(10)

Fig. 3. Message-passing equations. Here  $c_j$  is the modulo-2 addition of the values of the variable nodes in  $\mathcal{B}_j$  while  $\delta$ ,  $W_{sou}$ ,  $\alpha_{x_l,x'}$ , and  $\gamma$  are the parameters that can be adjusted.

B. Uniform Ternary Source With Hamming Distortion Measure

Let  $\mathcal{X} = \hat{\mathcal{X}} = \{0, 1, 2\}$  and  $d(\cdot, \cdot)$  be the Hamming distortion measure. Consider a ternary source X with  $P_X(0) = P_X(1) = P_X(2) = \frac{1}{3}$ . Let m = 4. It is easy to verify that we have  $P_{\hat{X}^*}(0) = P_{\hat{X}^*}(1) = P_{\hat{X}^*}(2) = \frac{1}{3}$  for all rates. Correspondingly, we choose  $f : \mathbb{F}_2^4 \to \mathcal{X}$  such that  $(P_{f(\tilde{X})}(0), P_{f(\tilde{X})}(1), P_{f(\tilde{X})}(2)) = (\frac{5}{16}, \frac{5}{16}, \frac{6}{16})$  for all rates.

## C. Uniform Ternary Source With $\ell_1$ Distortion Measure

Let  $\mathcal{X} = \hat{\mathcal{X}} = \{0, 1, 2\}$  and  $d(\cdot, \cdot)$  be the  $\ell_1$  distortion measure, i.e.,  $d(x, \hat{x}) = |x - \hat{x}|$ . Consider a ternary source X with  $P_X(0) = P_X(1) = P_X(2) = \frac{1}{3}$ . Let m = 4. Table II lists the optimal  $P_{f(\tilde{X})}$  for different rates.

The message-passing equations are given in Fig. 3. In (9), we choose  $\alpha_{x_l,x'} = 1$  if  $x' = x_l$ ,  $\alpha_{x_l,x'} = \alpha$  if  $|x_l - x'| = 1$ , and  $\alpha_{x_l,x'} = \beta$  if  $|x_l - x'| = 2$ . This choice is tailored to the  $\ell_1$  distortion measure. Since smaller distortion is preferable, we always set  $\alpha$  to be larger than  $\beta$ .

TABLE I THE OPTIMAL  $P_{f(\tilde{X})}$  for different rates.

R	$(P_{f(\tilde{X})}(0), P_{f(\tilde{X})}(1))$
0.479 - 1.000	$(\frac{4}{16}, \frac{12}{16})$
0.228 - 0.478	$(\frac{3}{16}, \frac{13}{16})$
0.096 - 0.227	$(\frac{2}{16}, \frac{14}{16})$
0.000 - 0.095	$(\frac{1}{16}, \frac{15}{16})$

TABLE II The optimal  $P_{f(\tilde{X})}$  for different rates.

R	$(P_{f(\tilde{X})}(0), P_{f(\tilde{X})}(1), P_{f(\tilde{X})}(2))$
0.43 - 1.58	$(\frac{5}{16}, \frac{6}{16}, \frac{5}{16})$
0.22 - 0.43	$(\frac{4}{16}, \frac{8}{16}, \frac{4}{16})$
0.13 - 0.22	$(\frac{3}{16}, \frac{10}{16}, \frac{3}{16})$
0 - 0.13	$(\frac{2}{16}, \frac{12}{16}, \frac{2}{16})$

TABLE III The optimal  $P_{f(\tilde{X})}$  for different rates.

#### D. Simulation Results

Now we examine the performance of the proposed scheme for the aforementioned three cases. The algorithm is implemented using C. The degree distributions of the LDGM codes are obtained from [33], which are optimized for the AWGN channel.

The simulation results are shown in Fig. 4, Fig. 5, and Fig. 6, respectively. Source sequences of length 1000 and 10000 are tested. Since m = 4, the corresponding LDGM codes are of length 4000 and 40000, respectively. The damping method [11] is used in the message-passing algorithm if the messages do not converge after 30 iterations. The decimation threshold is set to be 0.9 and the maximum number of iterations is set to be 100. Each plotted point is obtained by averaging over 1000 source sequences. It can be seen that the resulting distortions are close to the theoretical lower bound in all these three cases.

Note that although the degrees of variable nodes need to grow unboundedly in n to achieve the rate-distortion bound, simulation results indicate that relatively small degrees suffice for practical purposes. Moreover, it is worth mentioning that for a fixed n, the performance of the proposed message-passing algorithm may not improve with increasing degrees, and may even degrade.

#### V. CONCLUSION

We have shown that binary LDGM codes can achieve the rate-distortion bound of discrete memoryless sources with general distortion measure via multilevel quantization. Moreover, a practical encoding scheme based on the survey-propagation algorithm has been proposed. The simulation results validate the effectiveness of the proposed scheme. It is of considerable interest to generalize the results in the present work to the



Fig. 4. Non-uniform binary source with Hamming distortion measure.



Fig. 5. Uniform ternary source with Hamming distortion measure.

multi-user scenarios such as distributed source coding and multiple description coding.

## APPENDIX A Proof of Theorem 1

It is clear that  $R_{L,m}(D)$  is lower-bounded by R(D) for all m. Therefore, it suffices to show that  $\limsup_{m\to\infty} R_{L,m}(D) \leq R(D)$ .

Let  $P_{\hat{X}^*|X}$  be the optimal test channel associated with the rate-distortion pair (R(D), D) and  $P_{\hat{X}^*}$  be the output distribution induced by  $P_X$  and  $P_{\hat{X}^*|X}$ . We shall construct a random variable  $\tilde{X}$  over  $\mathbb{F}_2^m$  such that  $X - \hat{X}^* - \tilde{X}$  form a Markov chain. Let  $k_{\hat{x}}$  be the greatest integer satisfying  $\frac{k_{\hat{x}}}{2^m} \leq P_{\hat{X}^*}(\hat{x}), \hat{x} \in \hat{\mathcal{X}}$ . Let  $\mathcal{K}_{\hat{x}}, \hat{x} \in \hat{\mathcal{X}}$ , be disjoint subsets of  $\mathbb{F}_2^m$  with  $|\mathcal{K}_{\hat{x}}| = k_{\hat{x}}$ . Let  $\mathcal{K}^* = \mathbb{F}_2^m \setminus \cup_{\hat{x} \in \hat{\mathcal{X}}} \mathcal{K}_{\hat{x}}$ . Now construct the transition probability distribution  $P_{\tilde{X}|\hat{X}^*}$  as follows:

$$P_{\tilde{X}|\hat{X}^{*}}(\tilde{x}|\hat{x}) = \begin{cases} \frac{1}{2^{m}P_{\tilde{X}^{*}}(\hat{x})}, & \tilde{x} \in \mathcal{K}_{\hat{x}} \\ \frac{1}{|\mathcal{K}^{*}|} \left(1 - \frac{k_{\hat{x}}}{2^{m}P_{\tilde{X}^{*}}(\hat{x})}\right), & \tilde{x} \in \mathcal{K}^{*} \\ 0, & \text{otherwise} \end{cases}$$

j

1650



Fig. 6. Uniform ternary source with  $\ell_1$  distortion measure.

Without loss of generality, here we only define  $P_{\tilde{X}|\hat{X}^*}(\cdot|\hat{x})$  for  $\hat{x} \in \hat{\mathcal{X}}_+$ , where  $\hat{\mathcal{X}}_+ = \{\hat{x} \in \hat{\mathcal{X}} : P_{\hat{X}^*}(\hat{x}) > 0\}$ . It can be verified that

$$P_{\tilde{X}}(\tilde{x}) = \sum_{\hat{x} \in \hat{\mathcal{X}}_+} P_{\hat{X}^*}(\hat{x}) P_{\tilde{X}|\hat{X}^*}(\tilde{x}|\hat{x}) = \frac{1}{2^m}$$

if  $\tilde{x} \in \mathcal{K}_{\hat{x}}$  for some  $\hat{x} \in \hat{\mathcal{X}}$ , and

$$\begin{split} P_{\tilde{X}}(\tilde{x}) &= \sum_{\hat{x} \in \hat{\mathcal{X}}_{+}} P_{\hat{X}^{*}}(\hat{x}) P_{\tilde{X}|\hat{X}^{*}}(\tilde{x}|\hat{x}) \\ &= \frac{1}{|\mathcal{K}^{*}|} \sum_{\hat{x} \in \hat{\mathcal{X}}_{+}} \left( P_{\hat{X}^{*}}(\hat{x}) - \frac{k_{\hat{x}}}{2^{m}} \right) = \frac{1}{2^{m}} \end{split}$$

if  $\tilde{x} \in \mathcal{K}^*$ . Therefore, the constructed  $\tilde{X}$  is uniformly distributed over  $\mathbb{F}_2^m$ .

Define a deterministic mapping  $f : \mathbb{F}_2^m \to \hat{\mathcal{X}}$  such that  $f(\tilde{x}) = \hat{x}$  if  $\tilde{x} \in \mathcal{K}_{\hat{x}}$  for some  $\hat{x} \in \hat{\mathcal{X}}$ , and  $f(\tilde{x})$  is set to be an arbitrary reconstruction symbol if  $\tilde{x} \in \mathcal{K}^*$ . By the data processing inequality, we have

$$R(D) = I(X; \hat{X}^*) \ge I(X; \tilde{X}) \ge I(X; f(\tilde{X})).$$
(11)

Moreover, it is easy to verify that  $\lim_{m\to\infty} P_{X,f(\tilde{X})}(x,\hat{x}) = P_{X,\hat{X}^*}(x,\hat{x})$  for all  $x \in \mathcal{X}$  and  $\hat{x} \in \hat{\mathcal{X}}$ . Therefore, we have

$$\lim_{m \to \infty} \mathbb{E}[d(X, f(\tilde{X}))] = \mathbb{E}[d(X, \hat{X}^*)] \le D.$$
(12)

By (11), (12), and the continuity of the rate-distortion function, it can be readily shown that  $\limsup_{m\to\infty} R_{L,m}(D) \leq R(D)$ . The proof is complete.

## APPENDIX B PROOF OF THEOREM 2

Let  $\mathcal{T}_{P_X}^{\delta,n}$  denote the  $\delta$ -strongly typical set of sequences  $x^n$  with respect to the probability distribution  $P_X$ . Similarly, let  $\mathcal{T}_{P_{X,\tilde{X}}}^{\delta,n}$  denote the  $\delta$ -strongly typical set of sequences  $(x^n, \tilde{x}^n)$  with respect to  $P_{X,\tilde{X}}$ , where  $P_{X,\tilde{X}}$  is the probability distribution induced by  $P_X$  and  $P_{\tilde{X}|X}$ . Note that if  $(x^n, \tilde{x}^n) \in \mathcal{T}_{P_{X,\tilde{X}}}^{\delta,n}$ , then  $x^n \in \mathcal{T}_{P_X}^{\delta,n}$ . The definition of strongly typical set and its properties can be found in [24].

For each  $x^n \in \mathcal{X}^n$ , define

$$\begin{split} \phi_n(x^n, \mathbf{G}) &= \sum_{u^k \in \mathbb{F}_2^k} \mathbb{I}((x^n, \tilde{X}^n(u^k)) \in \mathcal{T}_{P_{X, \tilde{X}}}^{\delta, n}), \\ \psi_n(x^n, \mathbf{G}) &= \sum_{u^k \in \mathbb{F}_2^k} \mathbb{I}\Big(\frac{1}{n} \sum_{i=1}^n d(x_i, \hat{X}_i(u^k)) \le \mathbb{E}[d(X, f(\tilde{X}))] + \frac{\epsilon}{2}\Big). \end{split}$$

Let  $\delta$  be small enough so that  $\frac{1}{n} \sum_{i=1}^{n} d(x_i, f(\tilde{x}_i)) \leq \mathbb{E}[d(X, f(\tilde{X}))] + \frac{\epsilon}{2}$  whenever  $(x^n, \tilde{x}^n) \in \mathcal{T}_{P_{X, \tilde{X}}}^{\delta, n}$ . As a consequence, we have

$$\phi_n(x^n, \mathbf{G}) \le \psi_n(x^n, \mathbf{G}) \tag{13}$$

for all  $x^n \in \mathcal{X}^n$ .

Let  $\{w_n\}$  be a sequence such that  $w_n p_n \to \infty$  and  $\frac{w_n}{n} \to 0$ as  $n \to \infty$ . Since  $Y_i(u^k) = u^k \mathbf{g}_i$ ,  $i = 1, 2, \cdots, mn$ , it is easy to see that  $Y_1(u^k), Y_2(u^k), \cdots, Y_{mn}(u^k)$  are independent for any fixed  $u^k \in \mathbb{F}_2^k$ . Moreover, in view of the fact that

$$\Pr\{u^k \mathbf{g}_i = 0\} = \frac{1 + (1 - 2p_n)^{w(u^k)}}{2},$$

we have

$$\Pr\{Y_i(u^k) = 0\} \to \frac{1}{2} \tag{14}$$

as  $n \to \infty$ , where the convergence in (14) is uniform for all  $i \in \{1, 2, \dots, mn\}$  and  $u^k \in \mathbb{F}_2^k$  with  $w(u^k) \ge w_n$ . Therefore, it can be shown using [24, Lemma 13.6.2, p. 359] and a continuity argument that for any  $\delta' > 0$ , one can choose a sufficiently small  $\delta$  such that

$$2^{-n(I(X;f(\tilde{X}))-\delta')} \ge \mathbb{E}[\mathbb{I}((x^n, \tilde{X}^n(u^k)) \in \mathcal{T}_{P_{X,\tilde{X}}}^{\delta,n})]$$
$$\ge 2^{-n(I(X;f(\tilde{X}))+\delta')}$$
(15)

for all  $x^n \in \mathcal{T}_{P_X}^{\delta,n}$  and  $u^k \in \mathbb{F}_2^k$  with  $w(u^k) \ge w_n$  when n is sufficiently large. Note that

$$\sum_{u^k \in \mathbb{F}_2^k} \mathbb{E}[\mathbb{I}((x^n, \tilde{X}^n(u^k)) \in \mathcal{T}_{P_{X,\tilde{X}}}^{\delta,n})]$$
  
= 
$$\sum_{w(u^k) \ge w_n} \mathbb{E}[\mathbb{I}((x^n, \tilde{X}^n(u^k)) \in \mathcal{T}_{P_{X,\tilde{X}}}^{\delta,n})]$$
  
+ 
$$\sum_{w(u^k) < w_n} \mathbb{E}[\mathbb{I}((x^n, \tilde{X}^n(u^k)) \in \mathcal{T}_{P_{X,\tilde{X}}}^{\delta,n})].$$

Let  $R \triangleq R(n,k) = \frac{k}{n}$ . In view of (15) and the fact that  $2^{nR} \ge |\{u^k \in \mathbb{F}_2^k : w(u^k) \ge w_n\}| \ge 2^{n(R-\delta')}$  and  $|\{u^k \in \mathbb{F}_2^k : w(u^k) < w_n\}| \le 2^{n\delta'}$  for sufficiently large n, we have

$$2^{n(R-I(X;f(\tilde{X}))+\delta')} + 2^{n\delta'}$$

$$\geq \sum_{u^k \in \mathbb{F}_2^k} \mathbb{E}[\mathbb{I}((x^n, \tilde{X}^n(u^k)) \in \mathcal{T}_{P_{X,\tilde{X}}}^{\delta,n})]$$

$$\geq 2^{n(R-I(X;f(\tilde{X}))-2\delta')}$$
(16)

for all  $x^n \in \mathcal{T}_{P_X}^{\delta,n}$  when n is sufficiently large.

Now we shall derive a lower bound on  $\mathbb{E}[\phi_n(X^n, \mathbf{G})]$ . Note that

$$\mathbb{E}[\phi_n(X^n,\mathbf{G})]$$

$$\begin{split} &= \sum_{x^n \in \mathcal{T}_{P_X}^{\delta,n}} \Pr\{X^n = x^n\} \sum_{u^k \in \mathbb{F}_2^k} \mathbb{E}[\mathbb{I}((x^n, \tilde{X}^n(u^k)) \in \mathcal{T}_{P_{X, \tilde{X}}}^{\delta,n})] \\ &+ \sum_{x^n \notin \mathcal{T}_{P_X}^{\delta,n}} \Pr\{X^n = x^n\} \sum_{u^k \in \mathbb{F}_2^k} \mathbb{E}[\mathbb{I}((x^n, \tilde{X}^n(u^k)) \in \mathcal{T}_{P_{X, \tilde{X}}}^{\delta,n})] \\ &= \sum_{x^n \in \mathcal{T}_{P_X}^{\delta,n}} \Pr\{X^n = x^n\} \sum_{u^k \in \mathbb{F}_2^k} \mathbb{E}[\mathbb{I}((x^n, \tilde{X}^n(u^k)) \in \mathcal{T}_{P_{X, \tilde{X}}}^{\delta,n})], \end{split}$$

where the last equality is because  $\mathbb{I}((x^n, \tilde{x}^n) \in \mathcal{T}_{P_X, \tilde{x}}^{\delta, n}) = 0$ for all  $\tilde{x}^n$  if  $x^n \notin \mathcal{T}_{P_X}^{\delta, n}$ . In view of (16) and the fact that  $\Pr\{X^n \in \mathcal{T}_{P_X}^{\delta, n}\} \ge 1 - \delta'$  for all sufficiently large n, we have

$$\mathbb{E}[\phi_n(X^n, \mathbf{G})] \ge (1 - \delta')2^{n(R - I(X; f(X)) - 2\delta')}$$
(17)

when n is sufficiently large.

Now we proceed to derive an upper bound on  $\mathbb{E}[(\phi_n(X^n,\mathbf{G}))^2]$ . Note that

$$\begin{split} \mathbb{E}[(\phi_{n}(X^{n},\mathbf{G}))^{2}] \\ &= \mathbb{E}\Big[\sum_{u^{k},v^{k}\in\mathbb{F}_{2}^{k}}\mathbb{I}((X^{n},\tilde{X}^{n}(u^{k}))\in\mathcal{T}_{P_{X,\bar{X}}}^{\delta,n})\mathbb{I}((X^{n},\tilde{X}^{n}(v^{k})) \\ &\quad \in\mathcal{T}_{P_{X},\bar{X}}^{\delta,n})\Big] \\ &= \sum_{x^{n}\in\mathcal{T}_{P_{X}}^{\delta,n}}\Pr\{X^{n} = x^{n}\}\mathbb{E}\Big[\sum_{u^{k},v^{k}\in\mathbb{F}_{2}^{k}}\mathbb{I}((x^{n},\tilde{X}^{n}(u^{k})) \\ &\quad \in\mathcal{T}_{P_{X,\bar{X}}}^{\delta,n})\mathbb{I}((x^{n},\tilde{X}^{n}(v^{k}))\in\mathcal{T}_{P_{X,\bar{X}}}^{\delta,n})\Big] \\ &= \sum_{x^{n}\in\mathcal{T}_{P_{X}}^{\delta,n}}\Pr\{X^{n} = x^{n}\}\mathbb{E}\Big[\sum_{w(u^{k}\oplus v^{k})\geq 2w_{n}}\mathbb{I}((x^{n},\tilde{X}^{n}(u^{k})) \\ &\quad \in\mathcal{T}_{P_{X,\bar{X}}}^{\delta,n})\mathbb{I}((x^{n},\tilde{X}^{n}(v^{k}))\in\mathcal{T}_{P_{X,\bar{X}}}^{\delta,n}) \\ &\quad +\sum_{w(u^{k}\oplus v^{k})<2w_{n}}\mathbb{I}((x^{n},\tilde{X}^{n}(u^{k}))\in\mathcal{T}_{P_{X,\bar{X}}}^{\delta,n})\mathbb{I}((x^{n},\tilde{X}^{n}(v^{k})) \\ &\quad \in\mathcal{T}_{P_{X},\bar{X}}^{\delta,n})\Big], \end{split}$$

$$(18)$$

where  $\oplus$  is the modulo-2 addition. Define  $(u \wedge v)^k = ((u \wedge v)_1, (u \wedge v)_2, \cdots, (u \wedge v)_k)$  with  $(u \wedge v)_i = 1$  if  $u_i = v_i = 1$ and  $(u \wedge v)_i = 0$  otherwise. Define  $(u \setminus v)^k = u^k \oplus (u \wedge v)^k$ and  $(v \setminus u)^k = v^k \oplus (u \wedge v)^k$ . Note that  $w(u^k \oplus v^k) \ge 2w_n$ implies  $w((u \setminus v)^k) \ge w_n$  or  $w((v \setminus u)^k) \ge w_n$ . Without loss of generality, we shall assume  $w((v \setminus u)^k) \ge w_n$ , which further implies  $w(v^k) \ge w_n$ . In view of (14), we have

$$\begin{aligned} &\Pr\{\tilde{X}^n(v^k) = \tilde{x}^n\} \geq 2^{-mn} 2^{-\frac{n\delta'}{2}}, \\ &\Pr\{\tilde{X}^n((v \backslash u)^k) = \tilde{x}^n\} \leq 2^{-mn} 2^{\frac{n\delta'}{2}} \end{aligned}$$

for all  $\tilde{x}^n$  when n is sufficiently large. Therefore, if  $w(u^k \oplus v^k) \ge 2w_n$ , then

$$\leq 2^{-mn} 2^{\frac{n\delta'}{2}} \sum_{\tilde{c}^n} \Pr\{\tilde{X}^n((u \setminus v)^k) = \tilde{a}^n \oplus \tilde{c}^n, \\ \tilde{X}^n((u \wedge v)^k) = \tilde{c}^n\}$$
$$= 2^{-mn} 2^{\frac{n\delta'}{2}} \Pr\{\tilde{X}^n(u^k) = \tilde{a}^n\}$$
$$\leq 2^{n\delta'} \Pr\{\tilde{X}^n(u^k) = \tilde{a}^n\} \Pr\{\tilde{X}^n(v^k) = \tilde{b}^n\}$$
(19)

uniformly for all  $\tilde{a}^n$  and  $\tilde{b}^n$  when *n* is sufficiently large. Continuing from (18),

$$\mathbb{E}[(\phi_n(X^n, \mathbf{G}))^2] \leq \sum_{x^n \in \mathcal{T}_{P_X}^{\delta, n}} \Pr\{X^n = x^n\} \Big[ \sum_{w(u^k \oplus v^k) \ge 2w_n} 2^{n\delta'} \\
\times \mathbb{E}[\mathbb{I}((x^n, \tilde{X}^n(u^k)) \in \mathcal{T}_{P_{X, \tilde{X}}}^{\delta, n})] \mathbb{E}[\mathbb{I}((x^n, \tilde{X}^n(v^k)) \in \mathcal{T}_{P_{X, \tilde{X}}}^{\delta, n})] \\
+ \sum_{w(u^k \oplus v^k) < 2w_n} \mathbb{E}[\mathbb{I}((x^n, \tilde{X}^n(u^k)) \in \mathcal{T}_{P_{X, \tilde{X}}}^{\delta, n})] \\
\times \mathbb{I}((x^n, \tilde{X}^n(v^k)) \in \mathcal{T}_{P_{X, \tilde{X}}}^{\delta, n})] \Big]$$
(20)

$$\leq \sum_{x^{n} \in \mathcal{T}_{P_{X}}^{\delta, n}} \Pr\{X^{n} = x^{n}\} \Big[ \sum_{u^{k}, v^{k} \in \mathbb{F}_{2}^{k}} 2^{n\delta'} \mathbb{E}[\mathbb{I}((x^{n}, \tilde{X}^{n}(u^{k})) \in \mathcal{T}_{P_{X,\tilde{X}}}^{\delta, n})] \mathbb{E}[\mathbb{I}((x^{n}, \tilde{X}^{n}(v^{k})) \in \mathcal{T}_{P_{X,\tilde{X}}}^{\delta, n})] \\ + \sum_{w(u^{k} \oplus v^{k}) < 2w_{n}} \mathbb{E}[\mathbb{I}((x^{n}, \tilde{X}^{n}(u^{k})) \in \mathcal{T}_{P_{X,\tilde{X}}}^{\delta, n})]\Big] \\ \leq \sum_{x^{n} \in \mathcal{T}_{P_{X}}^{\delta, n}} \Pr\{X^{n} = x^{n}\} \Big[ \sum_{u^{k}, v^{k} \in \mathbb{F}_{2}^{k}} 2^{n\delta'} \mathbb{E}[\mathbb{I}((x^{n}, \tilde{X}^{n}(u^{k})) \in \mathcal{T}_{P_{X,\tilde{X}}}^{\delta, n})]\Big] \\ = \sum_{u^{k} \in \mathbb{F}_{2}^{k}} \Pr\{X^{n} = x^{n}\} \Big[ 2^{n\delta'} \Big( \sum_{u^{k} \in \mathbb{F}_{2}^{k}} \mathbb{E}[\mathbb{I}((x^{n}, \tilde{X}^{n}(u^{k})) \in \mathcal{T}_{P_{X,\tilde{X}}}^{\delta, n})]\Big]$$
(21)

$$\left\{ \mathcal{T}_{P_{X,\tilde{X}}}^{\delta,n} \right\} \right\}^{2} + 2^{n\delta'} \sum_{u^{k} \in \mathbb{F}_{2}^{k}} \mathbb{E}[\mathbb{I}((x^{n}, \tilde{X}^{n}(u^{k})) \in \mathcal{T}_{P_{X,\tilde{X}}}^{\delta,n})]$$

$$\leq 2^{n\delta'} (2^{n(R-I(X;f(\tilde{X}))+\delta')} + 2^{n\delta'})^{2} + 2^{n\delta'} (2^{n(R-I(X;f(\tilde{X}))+\delta')} + 2^{n\delta'}),$$
(22)

where (20) and (22) follow from (19) and (16), respectively, while (21) is due to the fact that for any  $u^k \in \mathbb{F}_2^k$ ,  $|\{v^k \in \mathbb{F}_2^k : w(u^k \oplus v^k) < 2w_n\}| \le 2^{n\delta'}$  when *n* is sufficiently large.

By [7, Lemma 2],

$$\Pr\{\phi_n(X^n, \mathbf{G}) > 0\} \ge \frac{(\mathbb{E}[\phi_n(X^n, \mathbf{G})])^2}{\mathbb{E}[(\phi_n(X^n, \mathbf{G}))^2]}.$$

Set  $R = I(X; f(\tilde{X})) + \epsilon$ . In view of (13), (17), and (22), we have

$$\liminf_{n \to \infty} \frac{1}{n} \log \Pr\{\psi_n(X^n, \mathbf{G}) > 0\}$$
  

$$\geq \liminf_{n \to \infty} \frac{1}{n} \log \Pr\{\phi_n(X^n, \mathbf{G}) > 0\}$$
  

$$\geq \liminf_{n \to \infty} \frac{1}{n} \log \frac{(\mathbb{E}[\phi_n(X^n, \mathbf{G})])^2}{\mathbb{E}[(\phi_n(X^n, \mathbf{G}))^2]}$$
  

$$\geq -7\delta'.$$

Since  $\delta' > 0$  is arbitrary, it follows that

$$\liminf_{n \to \infty} \frac{1}{n} \log \Pr\{\psi_n(X^n, \mathbf{G}) > 0\} \ge 0.$$

Now invoking [7, Lemma 1], one can readily show that

$$D(n, k, m, f) \leq \mathbb{E}[d(X, f(\tilde{X}))] + \frac{\epsilon}{2} + \frac{\epsilon}{2}$$
$$= \mathbb{E}[d(X, f(\tilde{X}))] + \epsilon$$

for some sufficiently large n. The proof is complete.

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