

Multimedia Communications

Mathematical Preliminaries for Lossy compression



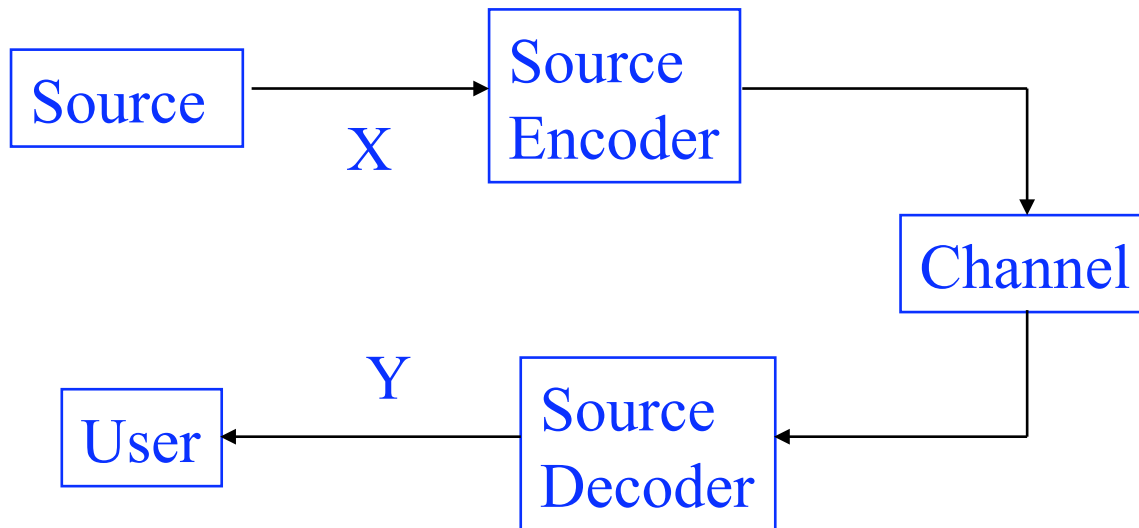
Lossy compression

- In a lossless compression, the reconstructed signal is identical to the original sequence
- Only a limited amount of compression can be obtained with lossless compression
- There is a floor (entropy of the source) below which we cannot drive the size of the compressed sequence
- In some applications consequences of loss of information may prohibit us from loss of information (bank records, medical images)

Lossy compression

- If the resources are limited and we do not require absolute integrity, we can improve the amount of compression by accepting certain degree of loss during compression
- Lossless compression: rate is the only performance measure
- In lossy compression rate by itself is not enough
- A measure of difference between original and reconstructed data
- Our goal: incur minimum amount of distortion while compressing to the lowest possible rate
- There is a tradeoff between minimizing rate and keeping distortion small

Lossy compression



- X can be modeled as a random variable
- Y is also a random variable

Lossy compression

- Exp: suppose source output consists of $\{0,1,2,\dots,15\}$. Source encoder quantizes the output into one of $\{0,1,2,\dots,7\}$. Source decoder inverse-quantizes the output of source encoder to $\{0,2,\dots,14\}$
- $X=\{0,1,2,\dots, 15\}$: 16 elements, $Y=\{0,2,4,\dots,14\}$: 8 elements
- Y is a random variable:
 - $P(Y=0)=P(Y=0|X=0)P(X=0)+P(Y=0|X=1)P(X=1)$
- Source encoder-decoder can be described by $P(Y|X)$

$$Q = \begin{bmatrix} p(y_1 | x_1) & p(y_1 | x_2) & p(y_1 | x_N) \\ p(y_2 | x_1) & p(y_2 | x_2) & \\ p(y_M | x_1) & p(y_M | x_2) & p(y_M | x_N) \end{bmatrix}$$

$$\sum_{i=1}^M p(y_i | x_j) = 1$$

Joint Entropy

- Let X and Y be two information sources with alphabets A_N and B_M and joint probability distribution $P : A_N \times B_M \rightarrow [0,1]$.

- The self-information of the event $(X=x, Y=y)$ is $i(x,y) = -\log P(x,y)$. The joint entropy of the two sources is $H(X,Y) = E[i(x,y)] = -\sum_{x \in A_N, y \in B_M} P(x,y) \log P(x,y)$

- $H(X,Y)$ is the average amount of information carried by a pair of values of the two information sources.
- $H(X,Y) \leq H(X) + H(Y)$ and the equality holds when the two sources are independent.

Conditional Entropy

- The conditional self-information of the event $(X=x|Y=y)$ is $i(x|y)=-\log P(x|y)$. The conditional entropy of the two sources is

$$H(X|Y) = E[i(x|y)] = -\sum_{x,y} P(x,y) \log P(x|y) = -\sum_{x,y} P(x,y) \log \frac{P(x,y)}{P(y)}$$

- $H(X|Y)$ is the average amount of information carried by X when the value of Y is known.
- $H(X|Y) + H(Y) = H(Y|X) + H(X) = H(X,Y)$
- $H(X|Y) \leq H(X)$, equality holds for independent sources.

Mutual Information

- Probability of source symbols $p(x)$: a priori probability
- After reconstructing y the probability of input symbol being x becomes $p(x|y)$: a posteriori probability
- Difference between information (uncertainty) before and after reconstruction of y measures the gain in information due to y :
$$i(x;y) = i(x) - i(x|y)$$
$$= -\log(p(x)) + \log(p(x|y)) = \log(p(x|y)/p(x)) = \log(p(x,y)/p(x)p(y))$$
- $i(x,y)$ is called mutual information of the x and y .
- Average mutual information of the two sources is the information that Y carries about X :

$$I(X;Y) = E[i(x; y)] = - \sum_{x,y} P(x, y) \log \frac{P(x, y)}{P(x)P(y)}$$

Mutual Information

- $I(X;Y) = I(Y;X)$
- $I(X;Y) \geq 0$, equality holds when the sources are independent.
- $I(X;Y) \leq H(X)$ and $I(X;Y) \leq H(Y)$
- $I(X;Y) = H(Y) - H(Y|X) = H(X) - H(X|Y) = H(X) + H(Y) - H(X,Y)$

Distortion

- Distortion criteria:
 1. Subjective criteria: measured by the effect the distortion has on the receiver. Not possible to model analytically.
 2. Objective mathematical measures: use mathematical formula to measure distortion. Analytical or numerical optimizations are possible.
 3. Objective measures in conjunction with models of the receiver (human auditory system, human visual system).
 - The process of human perception is very difficult to model and the models obtained are very complex

Mathematical Distortion Measures

- Average distortion

$$d_l(\underline{x}, \underline{y}) = \frac{1}{N} \sum_{i=1}^N |x_i - y_i|^l$$

- $l = 1 \rightarrow$ mean absolute error (MAE) $d_1(x, y) = |x - y|$
- $l = 2 \rightarrow$ mean square error (MSE)
- $l = \infty \rightarrow$ maximum error $d_\infty(\underline{x}, \underline{y}) = \max_i |x_i - y_i|$
- Sometimes the MSE is measured in dB:

$$\text{SNR(dB)} = 10 \log_{10} \frac{\sigma_x^2}{\sigma_d^2} \quad \text{PSNR(dB)} = 10 \log_{10} \frac{x_{\max}^2}{\sigma_d^2}$$

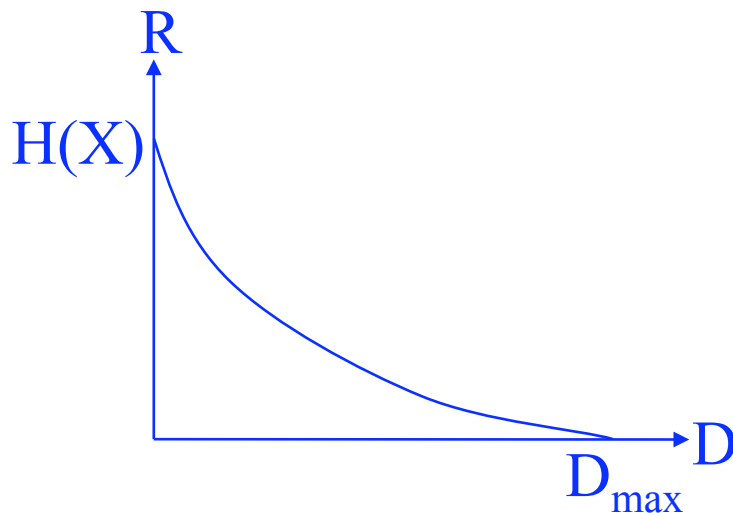
- In a probabilistic modeling framework:

$$D(X, Y) = \sum_{x, y} P(x, y) d(x, y) = \sum_{x, y} P(x) P(y | x) d(x, y)$$

Rate-Distortion Function: Definition

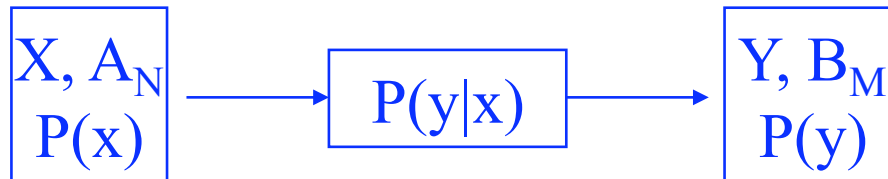
- Main trade-off: number of bits vs. signal distortion
- $R(D)$ specifies the lowest rate R at which the output of a source can be encoded while keeping the distortion $\leq D$.

- continuous
- monotonic decreasing
- convex
- $R(0) = H(X)$
- $R(D) = 0$ for $D \geq D_{\max}$



Rate-Distortion Function: Definition

- A coding scheme can be viewed as



where Y is the reconstructed signal, A_N and $P(x)$ describe the source and the conditional pdf $P(y|x)$ and B_M describe the coding scheme.

- We can express the rate-distortion function of a source in terms of these functions.

Rate-Distortion Function: Definition

- The average distortion of a coding scheme is:

$$D(X, Y) = \sum_{x, y} P(x, y) d(x, y) = \sum_{x, y} P(x) P(y | x) d(x, y)$$

- To find $R(D)$, we choose from the set of coding schemes that have a distortion smaller than D :

$$\Gamma = \{P(y | x); D(X, Y) \leq D\}$$

the one that achieves the smallest rate R .

- $R(D)$ is defined as

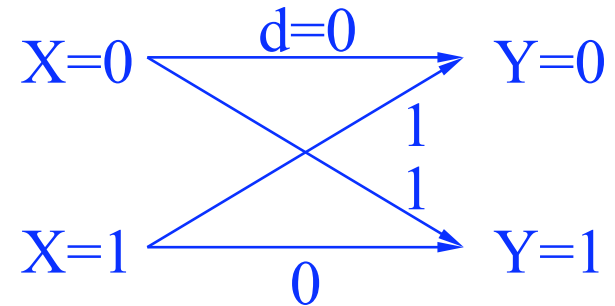
$$R(D) = \min_{p(y|x) \in \Gamma} I(X, Y) = - \sum_{x, y} P(x, y) \log \frac{P(x | y)}{P(x)}$$

Rate-Distortion Function: Computation

- How do we compute $R(D)$?
 - Analytically: find lower bound, show that it can be achieved.
 - Method of Lagrangian multipliers.
 - Computational approach: convex programming.
- In general, a difficult task.
- Can be found analytically only in a few simple cases.

Rate-Distortion Function: Example

- Consider a binary source with $P(0) = p$, $P(1) = 1-p$, $p < 1/2$.
- Let the distortion function be $d(x,y) = x \oplus y$ (xor)
- Find the rate distortion function for this source.
- We follow two approaches: Lagrangian multipliers and an analytical method



Rate-Distortion Function: Example

- Source encoder-decoder:

$$Q = \begin{bmatrix} p(y0|x0) & p(y0|x1) \\ p(y1|x0) & p(y1|x1) \end{bmatrix} = \begin{bmatrix} a & 1-b \\ 1-a & b \end{bmatrix}$$

$$I(X;Y) = - \sum_{x,y} P(y|x)P(x) \log \frac{P(y|x)P(x)}{P(x)P(y)}$$

$$D(X,Y) = \sum_{x,y} P(x)P(y|x)d(x,y)$$

$$R(D) = \min I(x,y)$$

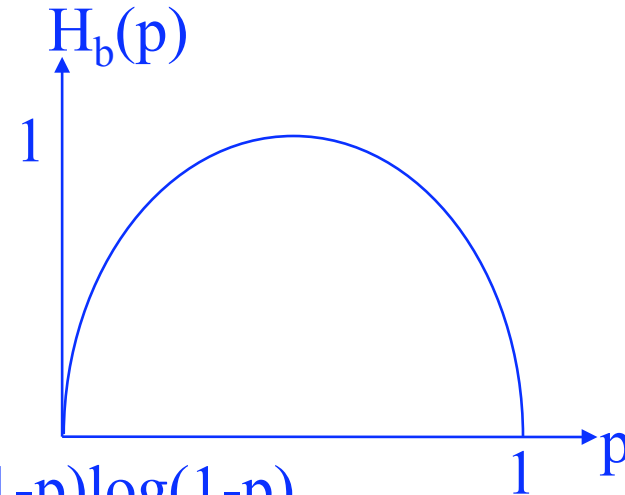
$$D(X,Y) < D$$

$$J(a,b,\lambda) = I(x,y) - \lambda(D(X,Y) - D)$$

$$R(D) = -p \log p - (1-p) \log(1-p) + D \log D + (1-D) \log(1-D)$$

Rate-Distortion Function: Example

- $P(x=0) = p, P(x=1) = 1-p, p < 1/2.$
- $d(x,y) = x \oplus y$ (xor)
- $R(0) = H(X) = -p \log p - (1-p) \log(1-p)$
- $H_b(p)$ is defined as $H_b(p) = -p \log p - (1-p) \log(1-p)$
- $H_b(p) = H_b(1-p)$
- $D(X,Y) = 1 \cdot P(x=0, y=1) + 1 \cdot P(x=1, y=0) = P(x \oplus y = 1)$
- D_{\max} is obtained when we send no bits, so we always decode $Y=1$ ($D = p$) or $Y=0$ ($D = 1-p$). The best of these two is $Y=1$, so $D_{\max} = p.$



Rate-Distortion Function: Example

- Then $\Gamma = \{P(y|x); D(X,Y) \leq D\} = \{P(y|x); P(x \oplus y = 1) \leq D\}$
- The mutual information is $I(X, Y) = H(X) - H(X|Y)$
- Since, when Y is known, we can obtain $X \oplus Y$ from X and vice versa, $H(X|Y) = H(X \oplus Y|Y)$ and then

$$I(X, Y) = H(X) - H(X \oplus Y | Y).$$

- A lower bound on $I(X, Y)$ is then

$$\begin{aligned} I(X, Y) &= H(X) - H(X|Y) \\ &= H(X) - H(X \oplus Y | Y) \\ &\geq H(X) - H(X \oplus Y) \\ &= H_b(p) - H_b(P(x \oplus y) = 1). \end{aligned}$$

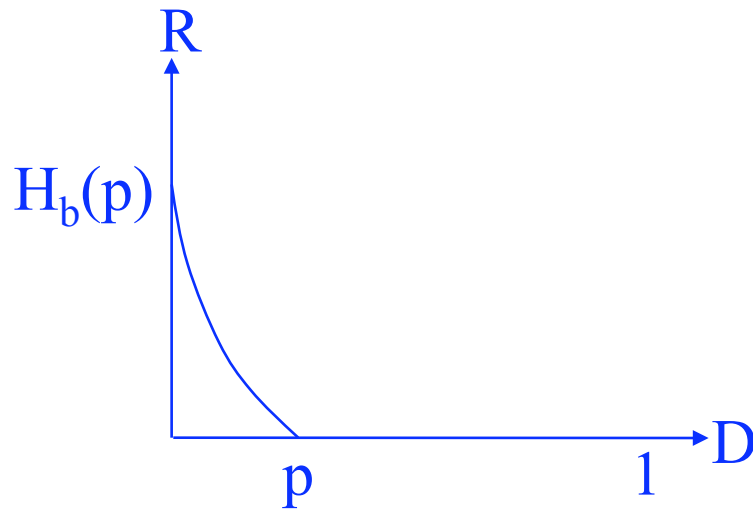
Rate-Distortion Function: Example

- But $P(x \oplus y = 1) \leq D$, so $I(X, Y) \geq H_b(p) - H_b(D)$.
- Let us now show that the lower bound is actually achievable:
Let $P(X=0|Y=1) = P(X=1|Y=0) = D$. Then
 - $P(X \oplus Y = 1) = P(X=1, Y=0) + P(X=0, Y=1)$
 $= P(Y=0)P(X=1|Y=0) + P(Y=1)P(X=0|Y=1)$
 $= P(Y=0) \cdot D + P(Y=1) \cdot D = D$
 - $P(X \oplus Y = 0) = 1 - D$
 - $I(X, Y) = H(X) - H(X \oplus Y) = H_b(p) - H_b(D)$, q.e.d.

Rate-Distortion Function: Example

- Putting it together,

$$R(D) = \begin{cases} H_b(p) - H_b(D), & D < p \\ 0, & \text{otherwise} \end{cases}$$



Continuous Amplitude Sources

- Continuous-amplitude memoryless sources are modeled as continuous RVs described by their pdf.
- Continuous RVs have infinite absolute entropy. We define the differential entropy as

$$h(X) = E[-\log p_X(x)] = - \int_{-\infty}^{+\infty} p_X(x) \log p_X(x) dx$$

- The entropy of a Gaussian random variable $h(X) = 1/2 \log(2\pi e\sigma^2)$
- For any source with variance σ^2 , $h(X) \leq h_G$.

Continuous Amplitude Sources

- The rate-distortion function of a continuous source is defined similarly with the discrete source:

$$I(X, Y) = -\int \int p(x, y) \log \frac{p(x | y)}{p(x)} dx dy$$

$$d(X, Y) = \int \int p(x) p(y | x) d(x, y) dx dy$$

$$R(D) = \inf_{p(y|x) \in \Gamma} I(X, Y)$$

$$\Gamma = \{p(y | x); d(X, Y) \leq D\}$$

Rate distortion

- The rate-distortion function for a zero mean Gaussian with distortion function $d(x,y) = (x-y)^2$ is

$$R(D) = \begin{cases} \frac{1}{2} \log \frac{\sigma^2}{D}, & D < \sigma^2 \\ 0, & \text{otherwise} \end{cases}$$

- Rate-distortion function for the Gaussian source is larger than the rate distortion function for any other source with a continuous distribution and the same variance
- The following are the lower bounds for a random variable X :
- squared error $R(D) = h(X) - \frac{1}{2} \log(2eD)$
- Absolute value error $R(D) = h(X) - \log(2eD)$

Noisy Source Coding Theorem

- Let $R(D)$ be the rate-distortion function of a stationary source. Then, for any $D > 0$ and $\epsilon > 0$:
 - POSITIVE THEOREM: a source code of block length n (sufficiently large) exists that encodes source vectors $\underline{X} = \{X_1, \dots, X_n\}$ at a rate $R < R(D) + \epsilon$ with $d(X, Y) \leq D$.
 - NEGATIVE THEOREM: a source code does not exist that encodes the source at a rate $R < R(D)$ with $d(X, Y) \leq D$.

Models for Continuous Sources

- **Uniform distribution:**
$$p_X(x) = \begin{cases} \frac{1}{b-a}, & x \in [a, b] \\ 0 & \text{otherwise} \end{cases}$$
 - The entropy $h(X) = \log(b-a)$.
- **Gaussian distribution:**
$$p_X(x) = \frac{1}{\sqrt{2\pi\sigma^2}} \exp\left[-\frac{(x-\mu)^2}{2\sigma^2}\right]$$
 - The entropy $h(X) = 1/2 \log(2\pi e\sigma^2) = h_G$
 - For any source with variance σ^2 , $h(X) \leq h_G$.

Models for Continuous Sources

- **Exponential distribution:**

- single-sided distribution

$$p_X(x) = \lambda \exp[-\lambda x] \quad x \geq 0$$

- **Laplacian distribution:**

- good for modeling peaked distributions

$$p_X(x) = \frac{1}{\sqrt{2\sigma^2}} \exp\left[-\frac{\sqrt{2}|x|}{\sigma}\right]$$

- **Gamma distribution:**

- even more peaked

$$p_X(x) = \frac{3^{1/4}}{\sqrt{8\pi\sigma|x|}} \exp\left[-\frac{\sqrt{3}|x|}{2\sigma}\right]$$

Models for Continuous Sources

- Generalized Gaussian distribution:

- $c = 2$ Gaussian
- $c = 1$ Laplacian
- $c = 0.7$ good model for subbands

$$p_X(x) = a \exp[-b(x - \mu)^c]$$

$$a = \frac{bc}{2\Gamma(1/c)}$$

$$b = \frac{1}{\sigma} \sqrt{\frac{\Gamma(3/c)}{\Gamma(1/c)}}$$

Models for Sources with Memory

- Linear system models:

$$x_n = \sum_{i=1}^N a_i x_{n-i} + \sum_{j=1}^M b_j \varepsilon_{n-j} + \varepsilon_n,$$

where ε_n is a Gaussian white noise sequence with variance σ_ε^2 .

- LTI system with N poles and M zeros.
- Auto-regressive moving average (ARMA) model.
- For $M = 0$ we obtain the “all-pole” or “AR” or “Markov” model. Good model for speech production.

Models for Sources with Memory

- AR(N) is an N-th order Markov model

$$x_n = \sum_{i=1}^N a_i x_{n-i} + \varepsilon_n$$

- For the first-order Markov source with an autocorrelation function of $\phi_k = |r|^k$ the rate-distortion function is

$$R(D) = \begin{cases} \frac{1}{2} \log \frac{1-r^2}{D}, & D \leq \frac{1-r}{1+r} \\ (\text{complex}), & \text{otherwise} \end{cases}$$