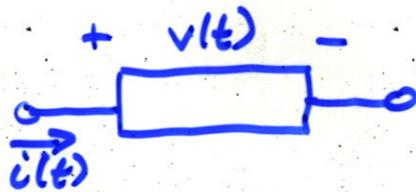


## AC Steady-State Power

- \* When dealing with power generation + distribution, we are often interested in steady-state performance of systems excited by sinusoidal signals
- \* One important quantity is the power generated or dissipated in a given element.



- \* The instantaneous power dissipated in this element is
- $$p(t) = v(t) i(t)$$
- in units of Watts
- \* However, this quantity can vary greatly in time
  - \* If the element is linear, then if the voltage is periodic with period  $T$ , ie if  $v(t+T)=v(t)$ , then the current is also periodic with period  $T$   
in that case,

$$p(t+T) = v(t+T) i(t+T) = p(t)$$

$\Rightarrow$  power is periodic with period  $T$

- \* In that case, we are often interested in the **average power**

$$P_{\text{av}} = \frac{1}{T} \int_{t_0}^{t_0+T} p(t) dt,$$

where  $t_0$  is arbitrary

- \* If the voltage is sinusoidal,

$$v(t) = V_m \cos(\omega t + \phi)$$

If the <sup>element</sup> circuit is linear, then in the steady state,

$$i(t) = I_m \cos(\omega t + \Theta)$$

- \* These signals are periodic with period  $T = \frac{2\pi}{\omega}$

- \* The instantaneous power is

$$p(t) = V_m I_m \cos(\omega t + \phi) \cos(\omega t + \Theta)$$

$$= \frac{V_m I_m}{2} [\cos(\phi - \Theta) + \cos(2\omega t + \phi + \Theta)]$$

- \* Choosing  $t_0 = 0$ ,

$$P_{\text{av}} = \frac{1}{T} \int_0^T p(t) dt$$

$$= \frac{V_m I_m \cos(\phi - \Theta)}{2T} \int_0^T dt$$

$$+ \frac{V_m I_m}{2T} \int_0^T \cos(2\omega t + \Theta + \phi) dt$$

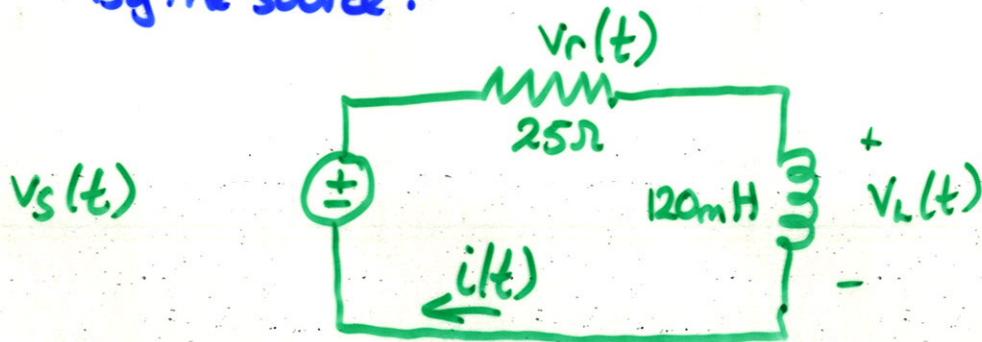
The second integral is the integral of a cosine over two periods, + hence is zero

Therefore

$$P_{av} = \frac{V_m I_m}{2} \cos(\phi - \theta)$$

### EXAMPLE

Find the average power delivered to the resistor and inductor. What is the average power generated by the source?



$$V_s(t) = 20 \cos(100t - 15^\circ) \text{ V}$$

Using phasor analysis, + voltage division,

$$V_r(t) = 18 \cos(100t - 41^\circ)$$

$$V_L(t) = 8.66 \cos(100t + 49^\circ)$$

$$i(t) = 721 \times 10^{-3} \cos(100t - 41^\circ)$$

## Resistor

$$P_{av,r} = \frac{18 \times 0.721}{2} \cos(-41^\circ - (-41^\circ)) \\ = 6.5W$$

Source,  $\rightarrow$  note the ~~current~~ passive sign convention means that power will be negative, because the source generates power

$$P_{s, \text{generated}} = \frac{20 \times 0.721}{2} \cos(-15^\circ - (-41^\circ)) = 6.5W$$

## Inductor -

$$P_{av,L} = \frac{8.66 \times 0.721}{2} \cos(49^\circ - (-41^\circ)) \\ = 0!$$

Why is this zero?

The voltage and current are  $90^\circ$  out of phase!  
The inductor absorbs/stores power on some part of the cycle, + provides power on other parts. These average out to zero.

## Effective Values of periodic waveforms.

Given a periodic current,  $i(t)$ , the average power delivered to a resistor  $R$  is

$$P_{av} = \frac{1}{T} \int_0^T i^2(t) R dt.$$

The effective value of  $i(t)$  is defined by

$$I_{eff} = \sqrt{\frac{P_{av}}{R}} \quad (*)$$

That is,

$$P_{av} = \frac{1}{T} \int_0^T i^2(t) R dt = I_{eff}^2 R.$$

Solving (\*) we have that

$$I_{eff} = \left( \frac{1}{T} \int_0^T i^2(t) dt \right)^{1/2}$$

for this reason  $I_{eff}$ , is often called the root mean square (RMS) value,  $I_{rms}$

Similarly, for a periodic voltage signal,

$$V_{rms} = \left( \frac{1}{T} \int_0^T v^2(t) dt \right)^{1/2}$$

For a sinusoidal signal  $s(t) = A \cos(\omega t + \phi)$

$$S_{\text{rms}} = \left( \frac{1}{T} \int_0^T A^2 \cos^2(\omega t + \phi) dt \right)^{\frac{1}{2}}$$
$$= \frac{A}{\sqrt{2}}$$

## Complex power

- \* It is often convenient to analyse the steady-state response of a linear circuit to a sinusoidal excitation in the "phasor domain"
- \* However to calculate power, we need to return to the time domain.
- \* An alternative approach is to define a notion of power in the phasor domain.
- \* Let there be a sinusoidal excitation of an element at one frequency,  $\omega_0$

Let  $\underline{I}(\omega) = I_m e^{j\Theta}$  and  $\underline{V}(\omega) = V_m e^{j\phi}$

be the current and voltage phasors, defined according to the passive sign convention

Then, the complex power delivered to the element is

$$S = \frac{\underline{V}(\omega) \underline{I}(\omega)^*}{2} = \frac{V_m e^{j\phi} \cdot I_m e^{-j\Theta}}{2}$$
$$= \frac{V_m I_m}{2} e^{j(\phi - \Theta)} = \frac{V_m I_m}{2} \angle \phi - \Theta$$

The asterisk denotes complex conjugation

The apparent power is

$$|S| = \frac{V_m I_m}{2}$$

- \* In rectangular coordinates,

$$S = \underbrace{\frac{V_m I_m}{2} \cos(\phi - \theta)}_{\text{This real part is the average power}} + j \underbrace{\frac{V_m I_m}{2} \sin(\phi - \theta)}_{\text{The imaginary part is called the reactive power}}$$

$$S = P + j Q.$$

Pav is measured in ~~Kilowatts~~ Watts (W)

S is measured in Volt-Amps (VA)

Q is measured in Volt-Amps-Reactive (VAR)

Given that the impedance of an element is

$$Z(\omega) = \frac{V(\omega)}{I(\omega)} = R + jX$$

The complex power can be written as  
(prove for homework)

$$S = \frac{I_m^2}{2} \operatorname{Re}\{Z\} + j \frac{I_m^2}{2} \operatorname{Im}\{Z\}$$

Hence  $P_{av} = \frac{I_m^2}{2} \operatorname{Re}\{Z\}$

Therefore, when the element is a resistor,

$$P_{av} = \frac{I_m^2 R}{2}$$

and when it is a capacitor or inductor

$$P_{av} = 0$$

See Table II-5-1 for a summary of power relations

Recall.

$$I_{eff} = \frac{I_m}{\sqrt{2}}$$

$$\Rightarrow P_{av} = I_{eff}^2 \cdot R$$

## Conservation of complex power

- \* The sum of the complex power absorbed by each element of a circuit is zero; ie complex power is conserved

$$\sum_{\text{all elements}} \frac{V_k I_k^*}{Z} = 0 \quad (*)$$

Note that this requires the passive sign convention to be adopted for each element. In that case, the real part of the complex power absorbed by a generating source will be negative

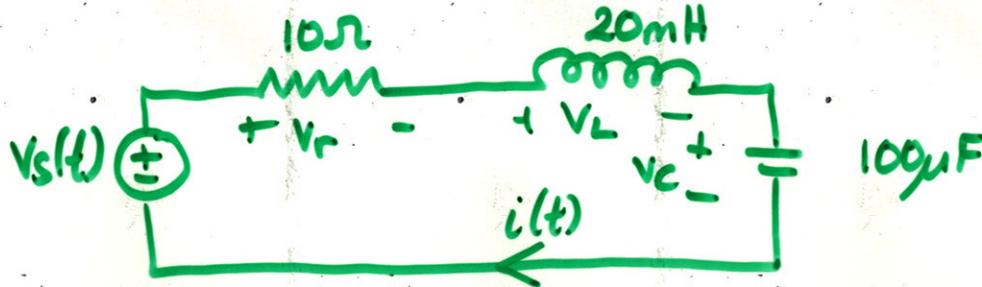
- \* By taking real + imaginary parts of (\*) average power + reactive power are conserved.  
ie

$$\sum_{\text{all elements}} P_{av, k} = 0$$

$$\sum_{\text{all elements}} Q_k = 0$$

## Example

Show that complex power is conserved in the following circuit when  $v_s(t) = 100 \cos 1000t$



Using phasor analysis

$$\underline{V}_s = 100$$

$$\underline{I}(\omega) = \frac{\underline{V}_s}{\underline{Z}(\omega)} = \frac{100}{10 + j\omega L - j\frac{1}{\omega C}}$$

$$= 7.07 e^{-j45^\circ}$$

Hence,  $\underline{V}_r = R \underline{I} = 70.7 e^{j45^\circ}$

$$\underline{V}_L = j\omega L \underline{I} = 141.4 e^{j45^\circ}$$

$$\underline{V}_C = -j\frac{1}{\omega C} \underline{I} = 70.7 e^{-j185^\circ}$$

\* For the source,  $V_s$  and  $I$  do not obey the passive sign convention, Hence

$$\underline{S}_s = \frac{V_s I^*}{2} = 353.5 e^{j45^\circ}$$

is the power generated by the source

\* For the other elements, we do satisfy the passive sign convention. Hence.

$$\underline{S}_r = 250$$

$$\underline{S}_L = 500 e^{j90^\circ}$$

$$\underline{S}_C = 250 e^{-j90^\circ}$$

\* If power is conserved, Then.

$$\underline{S}_s = \underline{S}_r + \underline{S}_L + \underline{S}_C$$

$$\begin{aligned}\text{Right hand side} &= 250 + j250 \\ &= 353.5 e^{j45^\circ}\end{aligned}$$

$\Rightarrow$  Hence power is conserved.