

COMP ENG 4TL4:

# Digital Signal Processing

Notes for Lecture #23

Friday, October 31, 2003

Modern window types (have been derived based on optimality criteria):

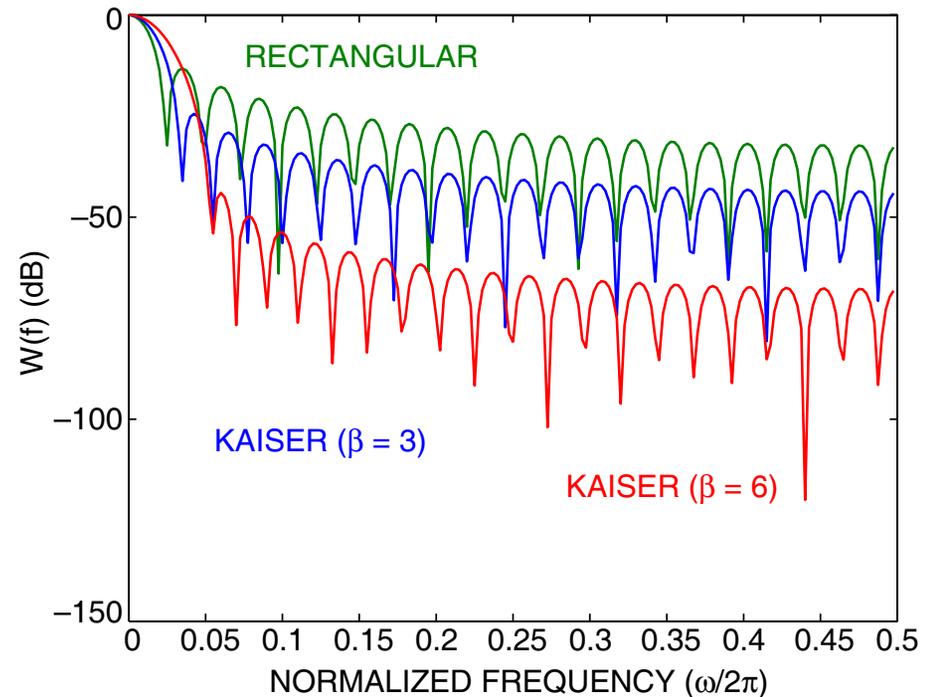
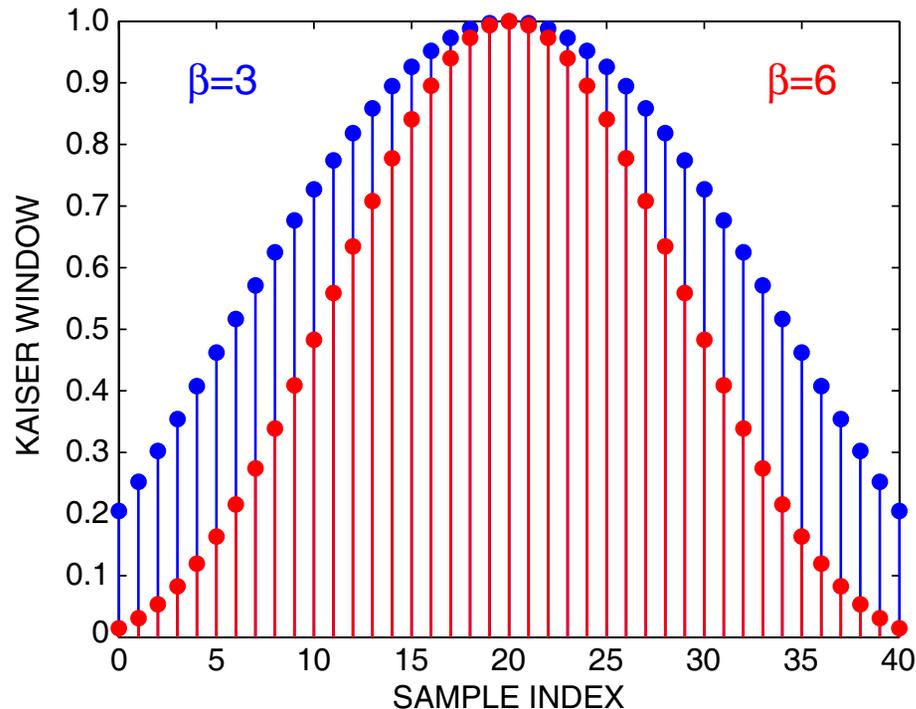
Kaiser window and Dolph-Chebyshev window:

*minimize the width of the mainlobe* under the constraints that:

1. *the window length be fixed* and
2. *the sidelobe levels not exceed a given value.*

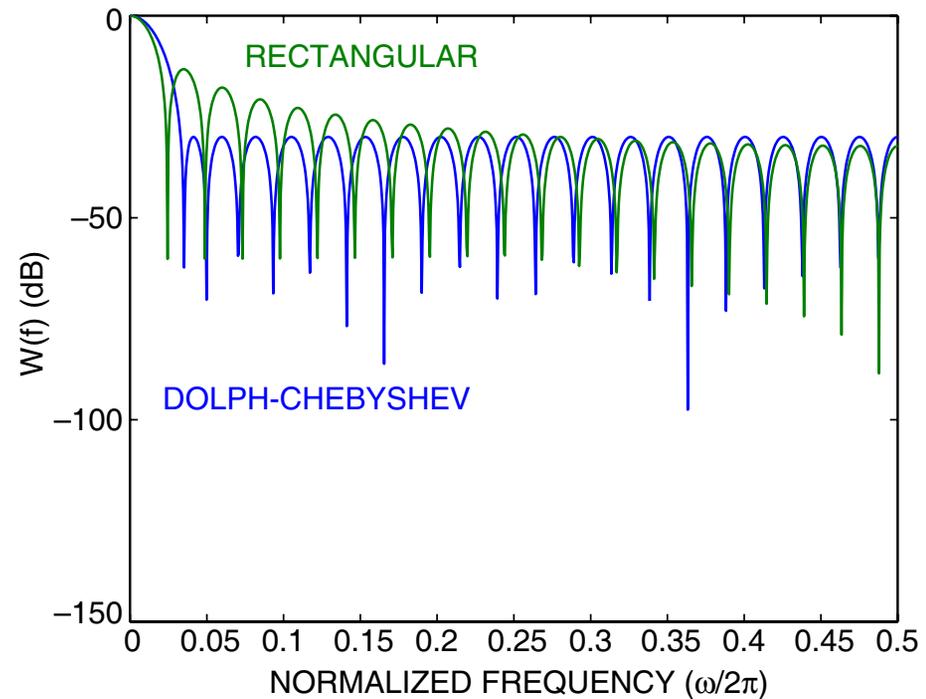
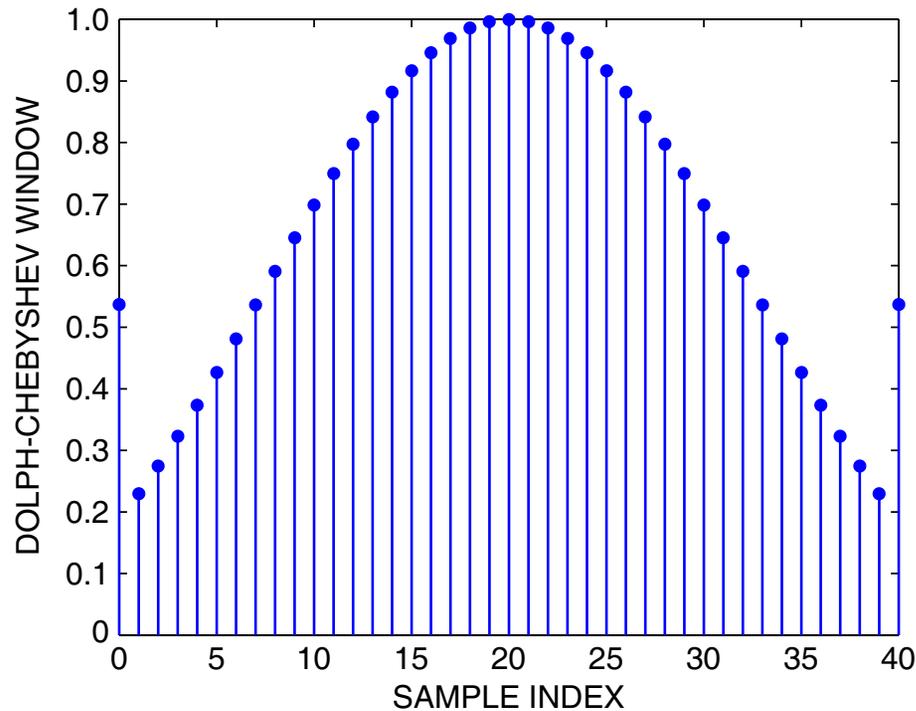
These windows provide more flexibility than the classical windows because a desired tradeoff between mainlobe width and sidelobe levels can be achieved!

# Kaiser window compared to the rectangular window in the frequency domain: ( $M = 40$ )



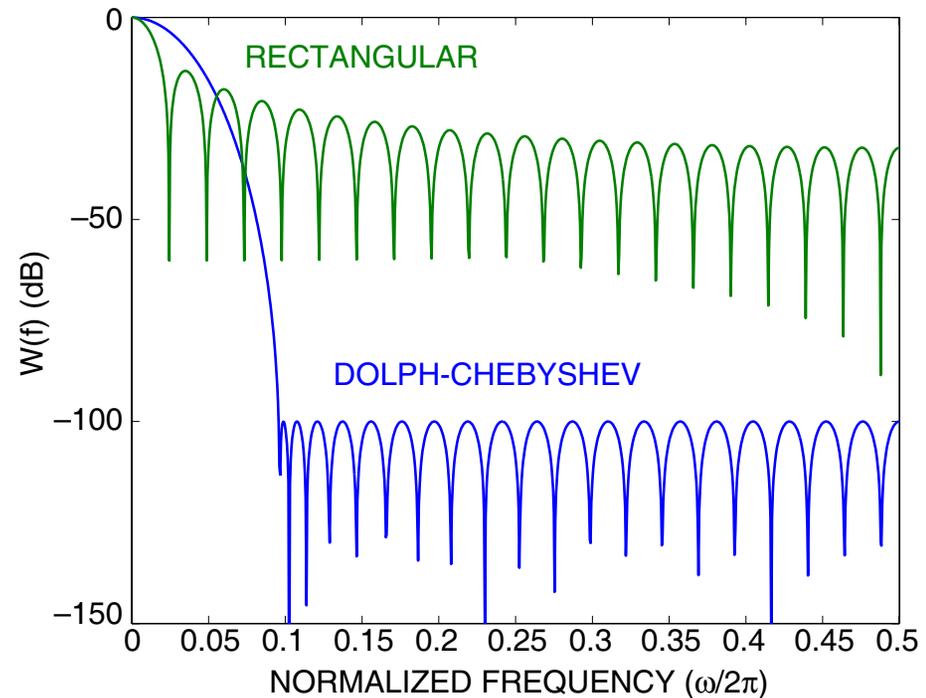
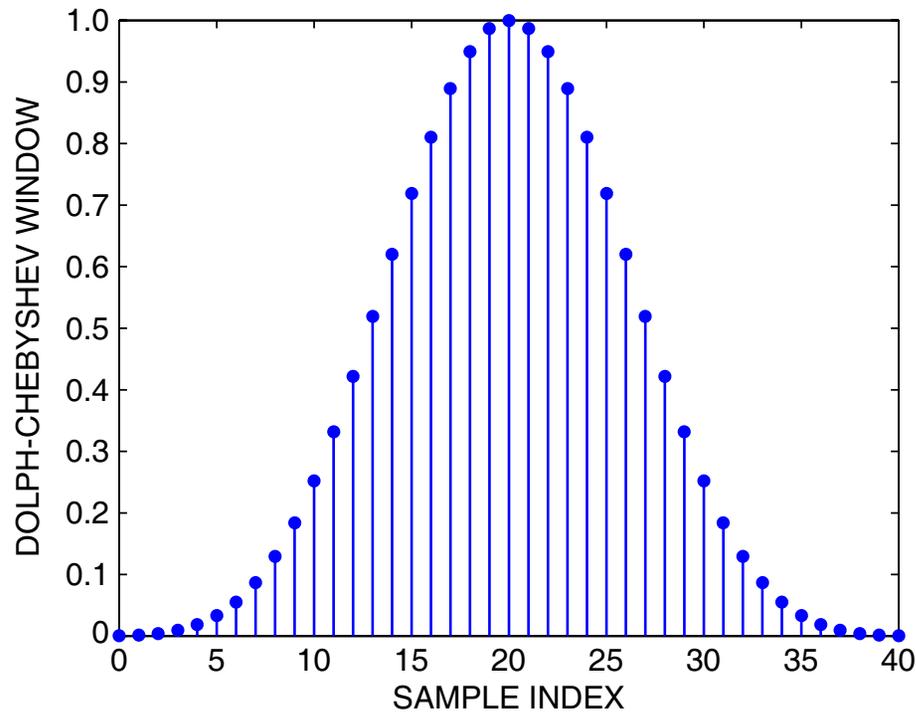
The Kaiser window has two parameters: the length  $M+1$  and a shape parameter  $\beta$ . The value of  $\beta$  required to achieve a particular maximal sidelobe level can be obtained from an empirically-derived formula—see *Oppenheim and Schaffer* pp. 474–485 for more details.

# Dolph-Chebyshev window with $-30$ dB of ripple compared to the rectangular window in the frequency domain: ( $M = 40$ )



The Dolph-Chebyshev window has two parameters: the length  $M+1$  and the desired sidelobe level. Note the equal levels of the sidelobes.

# Dolph-Chebyshev window with $-100$ dB of ripple compared to the rectangular window in the frequency domain: ( $M = 40$ )



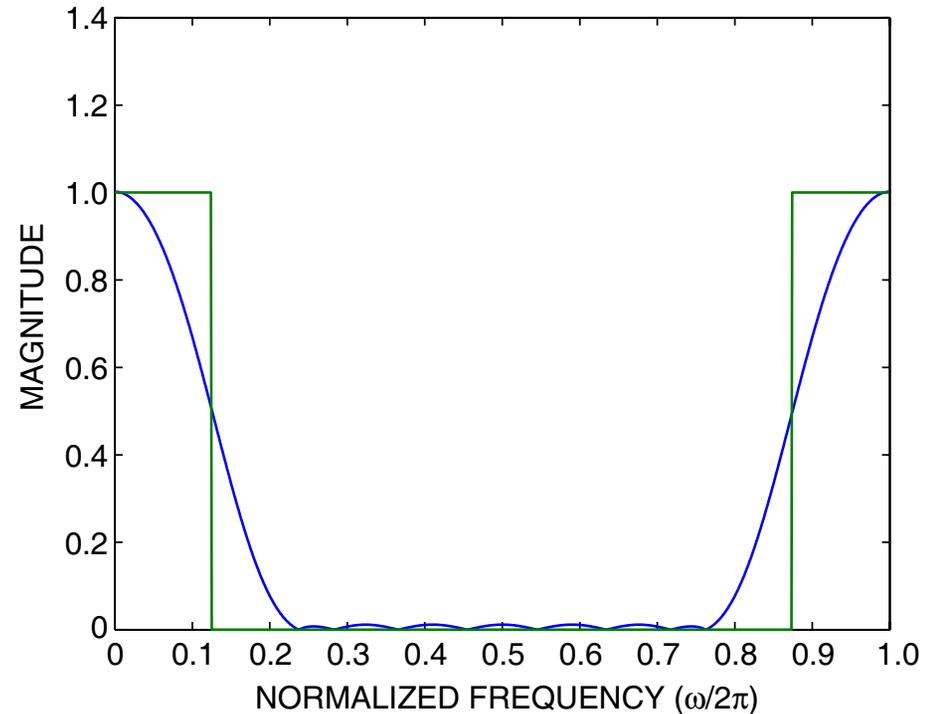
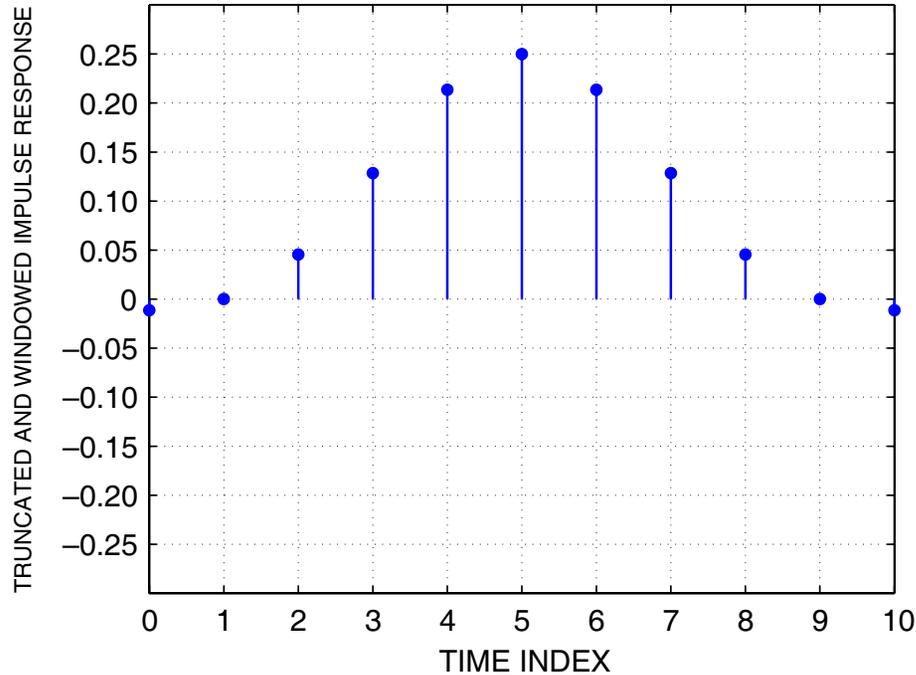
Now, let us take the Dolph-Chebyshev window and design our example lowpass filter using the window method with  $M = 10$ ,  $M = 40$ , and  $M = 160$ .

Remember that:

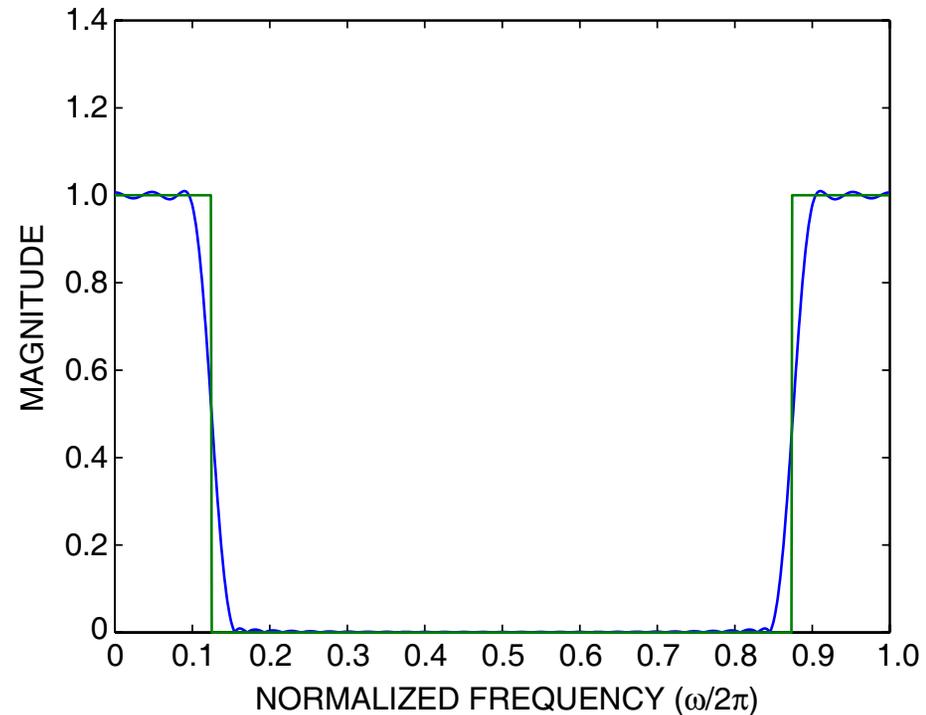
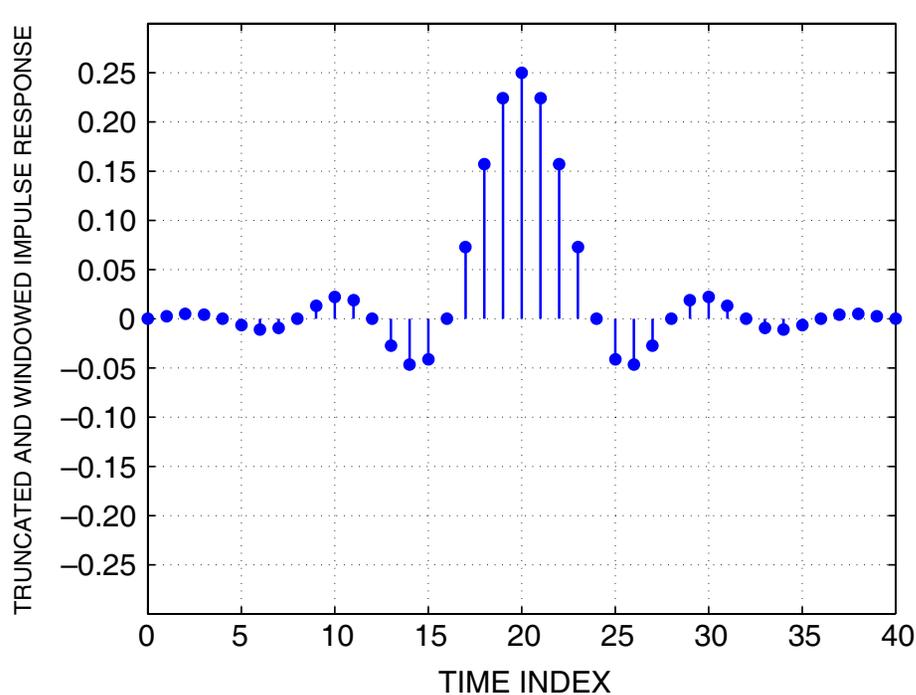
$$h_{\text{id}}[n] = \frac{\sin(\pi(n - M/2)/4)}{\pi(n - M/2)}.$$

We apply a Dolph-Chebyshev window of length  $M+1$ .

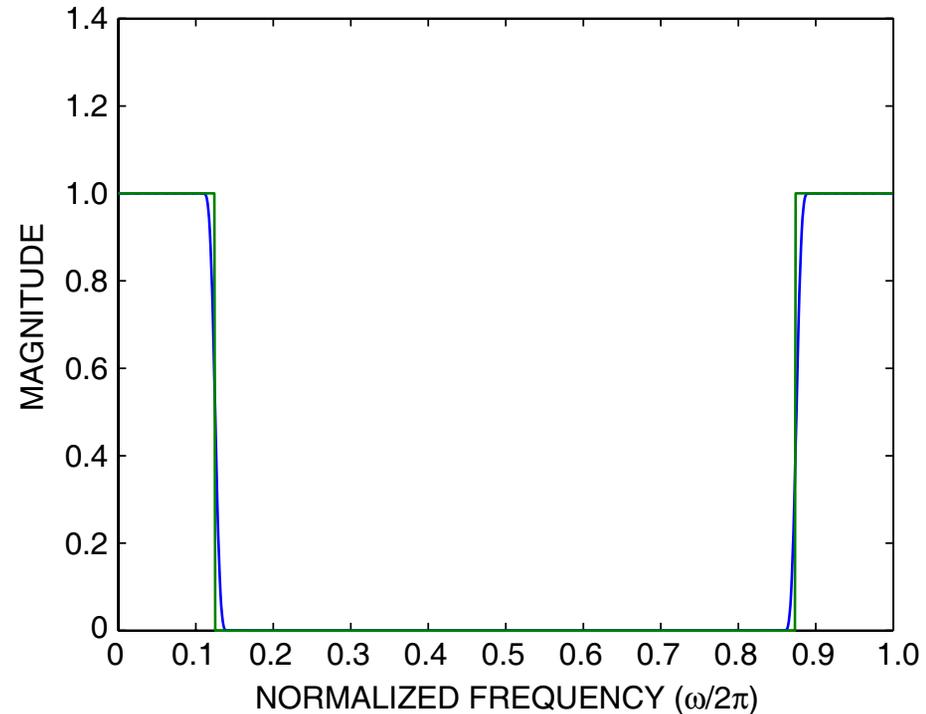
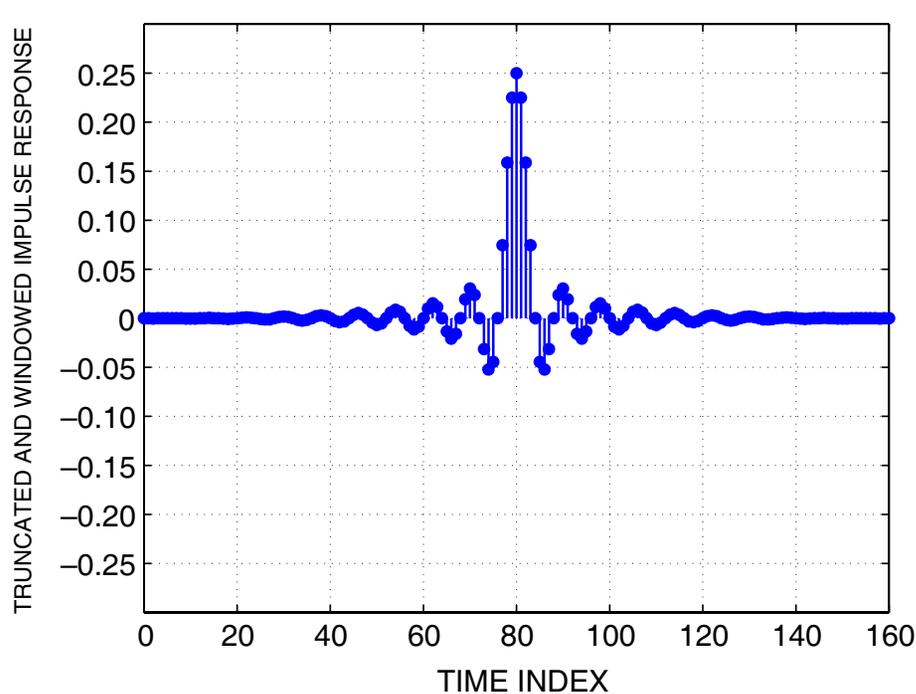
Truncated and windowed impulse response and corresponding approximation of lowpass frequency response:  
(window ripple  $-30$  dB;  $M = 10$ )



Truncated and windowed impulse response and corresponding approximation of lowpass frequency response:  
(window ripple  $-30$  dB;  $M = 40$ )



Truncated and windowed impulse response and corresponding approximation of lowpass frequency response:  
(window ripple  $-60$  dB;  $M = 160$ )



There is no Gibbs phenomenon with  
the Dolph-Chebyshev window!

Optimization-based methods: the idea is to find the best approximation to the ideal frequency response for a given fixed  $M$ .

Question: What should be our criterion for the “best” approximation?

Answer #1: How about minimizing the mean-square error?

$$\epsilon^2 = \frac{1}{2\pi} \int_{-\pi}^{\pi} |H_{\text{id}}(e^{j\omega}) - H(e^{j\omega})|^2 d\omega.$$

Unfortunately, the solution to this minimization problem is:

$$h[n] = \begin{cases} h_{\text{id}}[n] , & 0 \leq n \leq M, \\ 0 , & n > M, \end{cases}$$

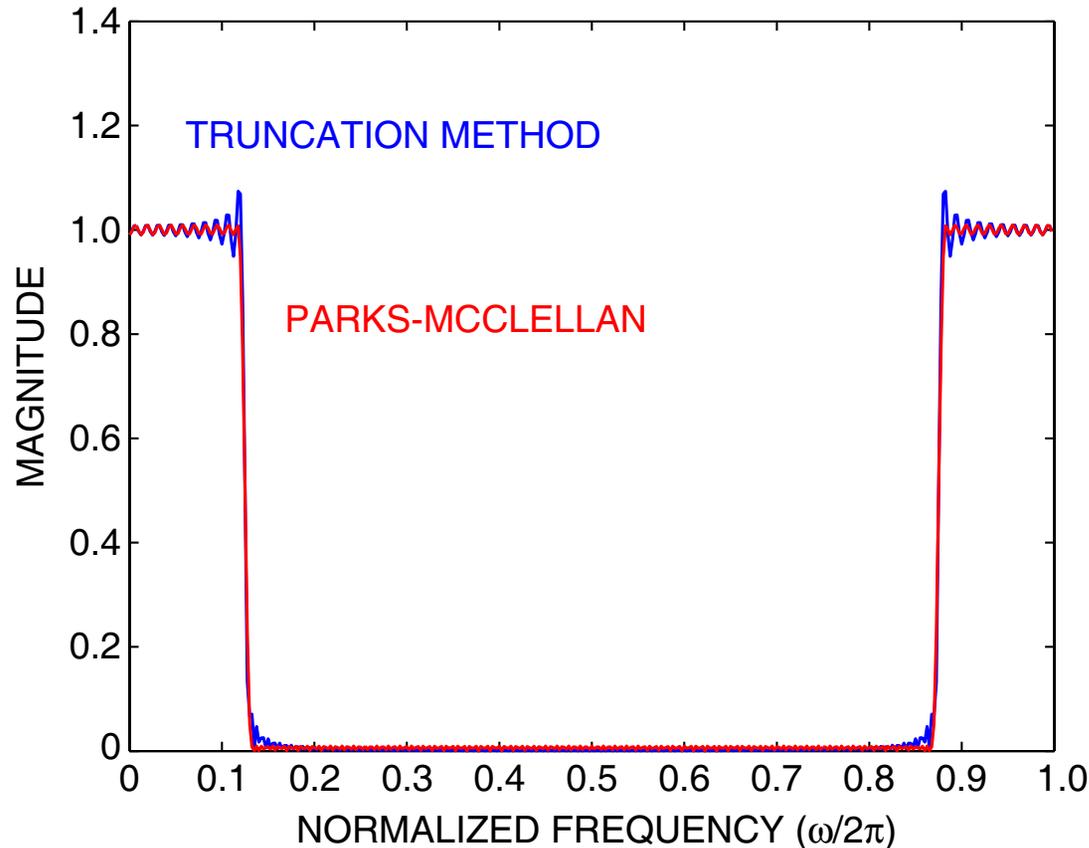
the truncation method, which we know suffers from the Gibbs phenomenon!

Answer #2: How about minimizing the maximum error?

This is referred to as the `minimax` strategy, and several algorithms have been developed to solve this problem.

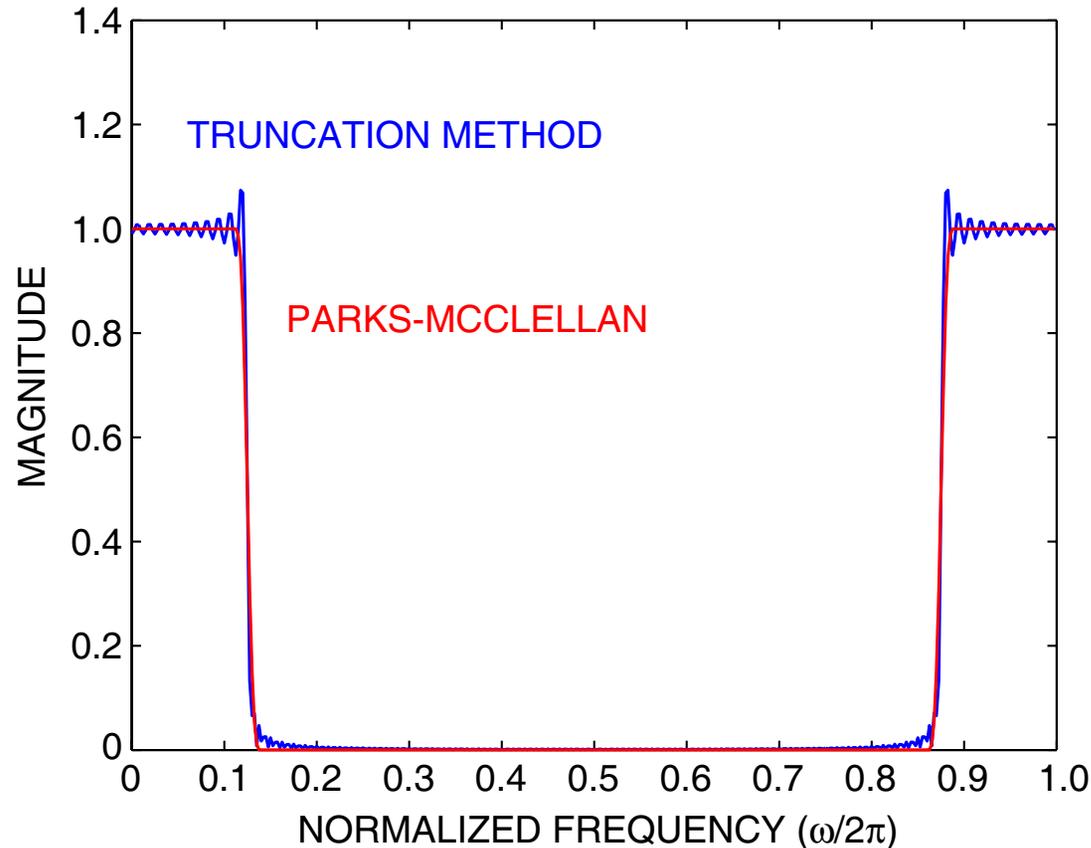
The most widely used of these is the Parks-McClellan algorithm, often referred to (mistakenly) as the `remez` algorithm, which determines the optimal (in the `minimax` sense) equiripple FIR filter for a given desired frequency response.

# Parks-McClellan optimal (minimax) equiripple FIR filter lowpass frequency response: ( $M = 160$ )



If some nonzero passband ripple and stopband attenuation is permissible, then a very sharp transition region matching that of the truncation method can be obtained.

# Parks-McClellan optimal (minimax) equiripple FIR filter lowpass frequency response: ( $M = 160$ )



With just a slight relaxation of the slope of the transition region, the passband ripple and stopband attenuation can be dramatically reduced.