

Chapter 4 (Solutions of selected problems).

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$$4.1-1 \quad F(\omega) = \int_{-\infty}^{+\infty} f(t) e^{-j\omega t} dt = \int_{-\infty}^{+\infty} f(t) \cos(\omega t) dt - j \int_{-\infty}^{+\infty} f(t) \sin(\omega t) dt$$

If $f(t)$ is an even function of t , $f(t)\sin\omega t$ is an odd function of t , and the second integral vanishes. Moreover, $f(t)\cos(\omega t)$ is an even function of t , and the first integral is twice the integral over the interval 0 to ∞ . Thus when $f(t)$ is even:

$$F(\omega) = 2 \int_0^{\infty} f(t) \cos(\omega t) dt \quad (1)$$

Similar argument shows that when $f(t)$ is odd:

$$F(\omega) = -2j \int_0^{\infty} f(t) \sin(\omega t) dt \quad (2)$$

If $f(t)$ is also real (in addition to being even), the integral (1) is real.

Moreover from (1):

$$F(-\omega) = 2 \int_0^{\infty} f(t) \cos(\omega t) dt = F(\omega)$$

Hence $F(\omega)$ is real and even function of ω . Similar arguments can be used to prove the rest of the properties.

▣

$$4.1-2 \quad f(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} F(\omega) e^{j\omega t} d\omega = \frac{1}{2\pi} \int_{-\infty}^{+\infty} |F(\omega)| e^{j\omega t + j\angle F(\omega)} d\omega$$

$$= \frac{1}{2\pi} \left[\int_{-\infty}^{+\infty} |F(\omega)| \cos[\omega t + \angle F(\omega)] d\omega + j \int_{-\infty}^{+\infty} |F(\omega)| \sin[\omega t + \angle F(\omega)] d\omega \right]$$

an even function which is twice the integral over 0 to ∞
an odd function which vanishes

$$\Rightarrow f(t) = \frac{1}{\pi} \int_0^{\infty} |F(\omega)| \cos[\omega t + \angle F(\omega)] d\omega$$

$$4.1-4 \quad a) \quad F(\omega) = \int_{-\infty}^{+\infty} f(t) e^{-j\omega t} dt = \int_0^T e^{-at} \cdot e^{-j\omega t} dt = \int_0^T e^{-(a+j\omega)t} dt$$

$$= -\frac{1}{(a+j\omega)} e^{-(a+j\omega)t} \Big|_{t=0}^{t=T} = \frac{1 - e^{-(a+j\omega)T}}{a+j\omega}$$

▣

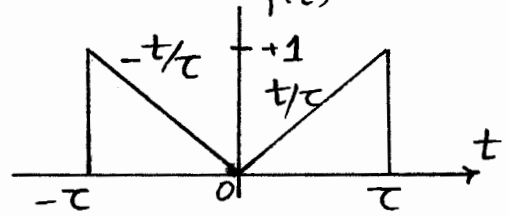
$$4.1-5 \quad (b) \quad F(\omega) = \int_{-\infty}^{+\infty} f(t) e^{-j\omega t} dt = \int_{-\tau}^0 -\frac{t}{\tau} e^{-j\omega t} dt + \int_0^{\tau} \frac{t}{\tau} e^{-j\omega t} dt$$

$$\int_{-\tau}^0 -\frac{t}{\tau} e^{-j\omega t} dt$$

using by-part
integration as:

$$t = u \rightarrow dt = du$$

$$e^{-j\omega t} dt = dv \rightarrow v = -\frac{1}{j\omega} e^{-j\omega t}$$



$$\rightarrow \int_{-\tau}^0 -\frac{t}{\tau} e^{-j\omega t} dt = -\frac{1}{\tau} \left[-\frac{t}{j\omega} e^{-j\omega t} \Big|_{-\tau}^0 - \int_{-\tau}^0 \left(-\frac{1}{j\omega}\right) e^{-j\omega t} dt \right]$$

$$= -\frac{1}{\tau} \left[-\frac{\tau}{j\omega} e^{j\omega\tau} + \frac{1}{\omega^2} e^{-j\omega t} \Big|_{-\tau}^0 \right]$$

$$= -\frac{1}{\tau} \left[-\frac{\tau}{j\omega} e^{j\omega\tau} + \frac{1}{\omega^2} (1 - e^{j\omega\tau}) \right] = \frac{1}{j\omega} e^{j\omega\tau} - \frac{1}{\omega^2\tau} (1 - e^{j\omega\tau})$$

$$\text{Similarly: } \int_0^{\tau} \frac{t}{\tau} e^{-j\omega t} dt = -\frac{1}{j\omega} e^{-j\omega\tau} + \frac{1}{\omega^2\tau} (e^{-j\omega\tau} - 1)$$

$$\Rightarrow F(\omega) = \frac{1}{j\omega} e^{+j\omega\tau} - \frac{1}{\omega^2\tau} (1 - e^{j\omega\tau}) - \frac{1}{j\omega} e^{-j\omega\tau} + \frac{1}{\omega^2\tau} (e^{-j\omega\tau} - 1)$$

$$= \frac{1}{j\omega} (e^{j\omega\tau} - e^{-j\omega\tau}) + \frac{1}{\omega^2\tau} (e^{j\omega\tau} + e^{-j\omega\tau} - 2)$$

$$= \frac{1}{j\omega} (j2 \sin \omega\tau) + \frac{1}{\omega^2\tau} (2 \cos \omega\tau - 2)$$

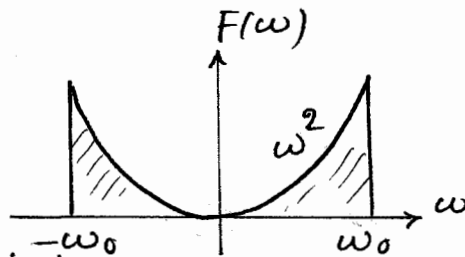
$$= \frac{2}{\omega} \sin(\omega\tau) + \frac{1}{\omega^2\tau} (2\cos\omega\tau - 2)$$

$$\rightarrow F(\omega) = \frac{2}{\omega^2\tau} (\cos\omega\tau + \omega\tau \sin\omega\tau - 1)$$

▣

4.1-6.

(a)



$$f(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} F(\omega) e^{j\omega t} d\omega = \frac{1}{2\pi} \int_{-\omega_0}^{+\omega_0} \omega^2 e^{j\omega t} d\omega$$

Using part-by-part integration to solve $I = \int_{-\omega_0}^{+\omega_0} \omega^2 e^{j\omega t} d\omega$:

$$u = \omega^2 \rightarrow du = 2\omega d\omega$$

$$e^{j\omega t} d\omega = dv \rightarrow v = \frac{1}{jt} e^{j\omega t}$$

$$\rightarrow I = \int_{-\omega_0}^{+\omega_0} \omega^2 e^{j\omega t} d\omega = \left[\frac{\omega^2}{jt} e^{j\omega t} \Big|_{-\omega_0}^{+\omega_0} - \int_{-\omega_0}^{+\omega_0} \frac{1}{jt} e^{j\omega t} 2\omega d\omega \right]$$

$$= \left[\frac{1}{jt} \left(\omega_0^2 e^{j\omega_0 t} - \omega_0^2 e^{-j\omega_0 t} \right) - \frac{2}{jt} \int_{-\omega_0}^{+\omega_0} \omega e^{j\omega t} d\omega \right]$$

$$= \frac{2}{jt} \left(e^{j\omega_0 t} - e^{-j\omega_0 t} \right)$$

$$= \omega_0^2 (j2 \sin\omega_0 t)$$

The second term, $I_1 = \int_{-\omega_0}^{+\omega_0} \omega e^{j\omega t} d\omega$, can be solve using by-part integration:

$$I_1 = \int_{-\omega_0}^{+\omega_0} \omega e^{j\omega t} d\omega: \quad u = \omega \rightarrow du = d\omega$$

$$e^{j\omega t} d\omega = dv \rightarrow v = \frac{1}{jt} e^{j\omega t}$$

(3)

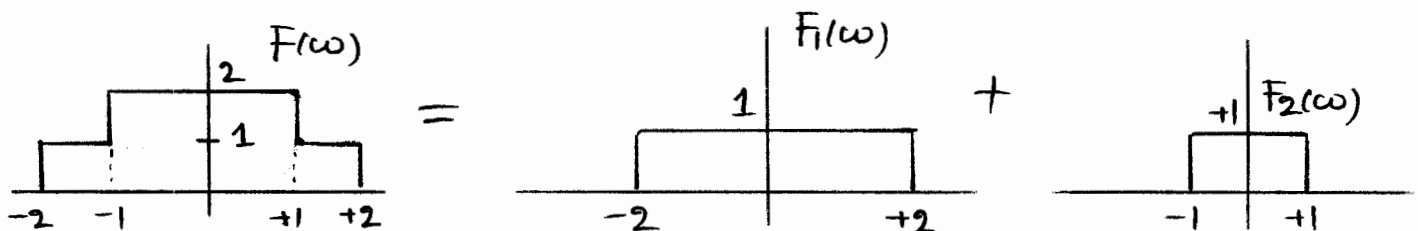
$$\begin{aligned}
 \Rightarrow I_1 &= \int_{-\omega_0}^{+\omega_0} \omega e^{j\omega t} d\omega = \frac{\omega}{jt} e^{j\omega t} \Big|_{-\omega_0}^{+\omega_0} - \int_{-\omega_0}^{+\omega_0} \frac{1}{jt} e^{j\omega t} d\omega \\
 &= \frac{1}{jt} (\omega_0 e^{j\omega_0 t} + \omega_0 e^{-j\omega_0 t}) - \frac{1}{jt} \left(\frac{1}{jt} \right) e^{j\omega t} \Big|_{-\omega_0}^{+\omega_0} \\
 &= \frac{\omega_0}{jt} (2 \cos \omega_0 t) + \frac{1}{t^2} (e^{j\omega_0 t} - e^{-j\omega_0 t}) \\
 &= \frac{\omega_0}{jt} (2 \cos \omega_0 t) + \frac{1}{t^2} (j2 \sin \omega_0 t)
 \end{aligned}$$

Putting I_1 in the expression of I ,

$$\begin{aligned}
 \Rightarrow I &= \frac{1}{jt} \omega_0^2 (j2 \sin \omega_0 t) - \frac{2}{jt} \left[\frac{\omega_0}{jt} 2 \cos \omega_0 t + \frac{1}{t^2} (j2 \sin \omega_0 t) \right] \\
 &= \frac{2\omega_0^2}{t} \sin(\omega_0 t) + \frac{4\omega_0}{t^2} \cos(\omega_0 t) - \frac{4}{t^3} \sin \omega_0 t \\
 &= \frac{2(\omega_0^2 t^2 - 2) \sin(\omega_0 t) + 4\omega_0 t \cos \omega_0 t}{t^3}
 \end{aligned}$$

$$f(t) = \frac{1}{2\pi} I \rightarrow f(t) = \frac{(\omega_0^2 t^2 - 2) \sin(\omega_0 t) + 2\omega_0 t \cos \omega_0 t}{\pi t^3}$$

(b) The derivation can be simplified by observing that $F(\omega)$ can be expressed as a sum of two gate functions $F_1(\omega)$ and $F_2(\omega)$ as shown below:

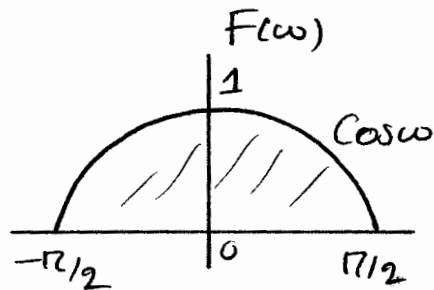


Therefore:

$$\begin{aligned}
 f(t) &= \frac{1}{2\pi} \int_{-2}^{+2} [F_1(\omega) + F_2(\omega)] e^{j\omega t} d\omega = \frac{1}{2\pi} \left\{ \int_{-2}^{+2} e^{j\omega t} d\omega + \int_{-1}^{+1} e^{j\omega t} d\omega \right\} \\
 &= \frac{1}{2\pi} \left\{ \frac{1}{jt} (e^{j2t} - e^{-j2t}) + \frac{1}{jt} (e^{jt} - e^{-jt}) \right\} \\
 &= \frac{1}{2\pi} \left\{ \frac{1}{jt} (j2 \sin 2t) + \frac{1}{jt} (j2 \sin t) \right\} = \frac{\sin 2t + \sin t}{\pi t}
 \end{aligned}$$

▣

4.1-7 (a)



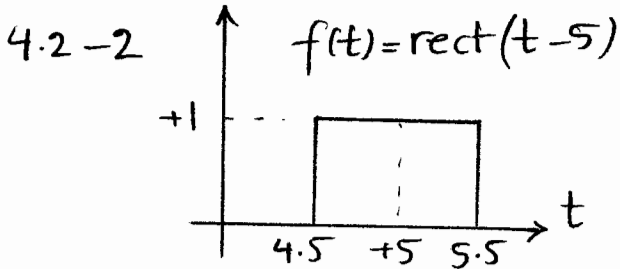
$$\begin{aligned}
 f(t) &= \frac{1}{2\pi} \int_{-\infty}^{+\infty} F(\omega) e^{j\omega t} d\omega = \frac{1}{2\pi} \int_{-\pi/2}^{+\pi/2} \cos \omega e^{j\omega t} d\omega \\
 &= \frac{1}{2\pi} \int_{-\pi/2}^{+\pi/2} \frac{1}{2} (e^{j\omega} + e^{-j\omega}) e^{j\omega t} d\omega = \frac{1}{4\pi} \left[\int_{-\pi/2}^{+\pi/2} (e^{j\omega} \cdot e^{j\omega t} + e^{-j\omega} \cdot e^{j\omega t}) d\omega \right] \\
 &= \frac{1}{4\pi} \left[\int_{-\pi/2}^{+\pi/2} e^{j(t+1)\omega} d\omega + \int_{-\pi/2}^{+\pi/2} e^{j(t-1)\omega} d\omega \right] \\
 &= \frac{1}{4\pi} \left[\frac{1}{j(t+1)} e^{j(t+1)\omega} \Big|_{\omega=-\pi/2}^{+\pi/2} + \frac{1}{j(t-1)} e^{j(t-1)\omega} \Big|_{-\pi/2}^{+\pi/2} \right] \\
 &= \frac{1}{4\pi} \left[\frac{1}{j(t+1)} \left(e^{j\frac{\pi}{2}(t+1)} - e^{-j\frac{\pi}{2}(t+1)} \right) + \frac{1}{j(t-1)} \left(e^{j\frac{\pi}{2}(t-1)} - e^{-j\frac{\pi}{2}(t-1)} \right) \right] \\
 &= \frac{1}{4\pi} \left[\frac{1}{j(t+1)} (j2 \sin(\frac{\pi}{2}(t+1))) + \frac{1}{j(t-1)} (j2 \sin(\frac{\pi}{2}(t-1))) \right]
 \end{aligned}$$

(5)

$$= \frac{1}{2\pi} \left[\frac{1}{t+1} \cos\left(\frac{\pi t}{2}\right) - \frac{1}{t-1} \cos\left(\frac{\pi t}{2}\right) \right]$$

$$= \frac{1}{2\pi} \cos\left(\frac{\pi t}{2}\right) \left(\frac{t+1-t-1}{t^2-1} \right) = \frac{1}{\pi(t^2-1)} \cos\left(\frac{\pi t}{2}\right)$$

▣

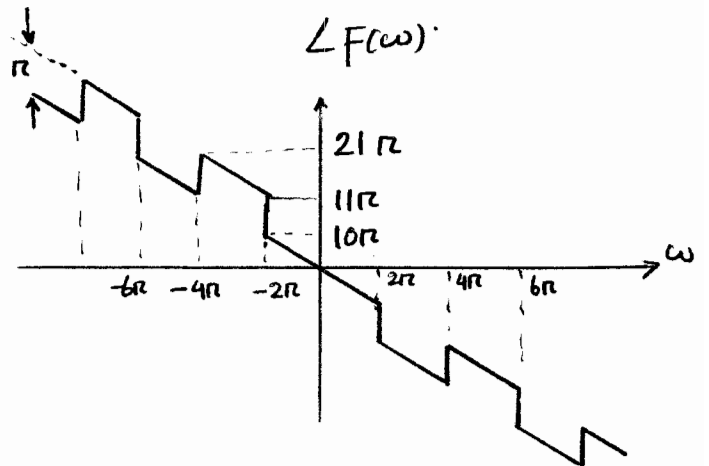
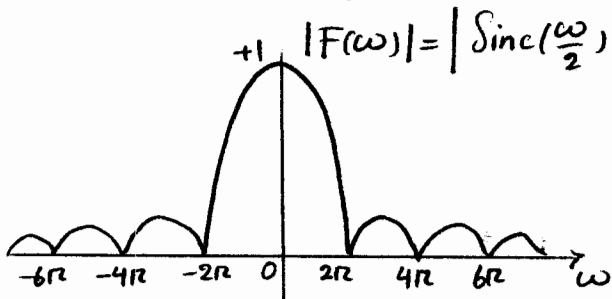


$$F(\omega) = \int_{-\infty}^{+\infty} f(t) e^{-j\omega t} dt = \int_{4.5}^{5.5} e^{-j\omega t} dt = -\frac{1}{j\omega} e^{-j\omega t} \Big|_{t=4.5}^{5.5}$$

$$= -\frac{1}{j\omega} (e^{-j5.5\omega} - e^{-j4.5\omega}) = \frac{-e^{-j5\omega}}{j\omega} (e^{-j0.5\omega} - e^{j0.5\omega})$$

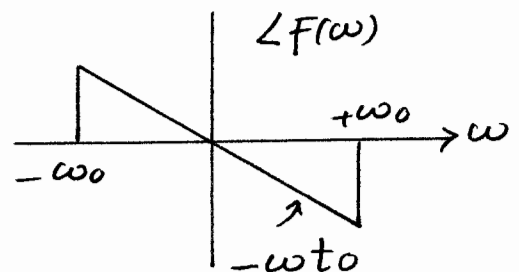
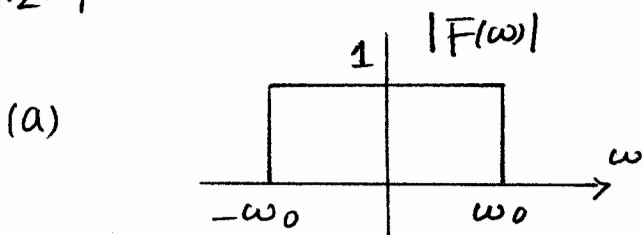
$$= \frac{-e^{-j5\omega}}{j\omega} (-j2 \sin(0.5\omega)) = \frac{2}{\omega} \sin\left(\frac{\omega}{2}\right) e^{-j5\omega} = \frac{\text{Sinc}(0.5\omega)}{0.5\omega} e^{-j5\omega}$$

$$\rightarrow F(\omega) = \text{Sinc}\left(\frac{\omega}{2}\right) e^{-j5\omega}$$



▣

4.2-4



(b)

$$F(\omega) = |F(\omega)| e^{j\angle F(\omega)} = e^{-j\omega t_0}$$

$$f(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} F(\omega) e^{j\omega t} d\omega = \frac{1}{2\pi} \int_{-\omega_0}^{+\omega_0} e^{-j\omega t_0} e^{j\omega t} d\omega = \frac{1}{2\pi} \int_{-\omega_0}^{+\omega_0} e^{-j\omega(t-t_0)} d\omega$$

$$= \frac{1}{2\pi} \left(\frac{1}{-j(t-t_0)} \right) e^{-j\omega(t-t_0)} \Big|_{\omega=-\omega_0}^{\omega_0} = \frac{1}{2\pi} \left(-\frac{1}{j(t-t_0)} \right) (e^{-j\omega_0(t-t_0)} - e^{j\omega_0(t-t_0)})$$

$$= \frac{1}{2\pi} \left(-\frac{1}{j(t-t_0)} \right) (-j2 \sin \omega_0(t-t_0))$$

$$\rightarrow f(t) = \frac{\sin \omega_0(t-t_0)}{\pi(t-t_0)} = \frac{\omega_0}{\pi} \text{sinc}(\omega_0(t-t_0))$$

▣

4.3-1

(a)

$$\underbrace{u(t)}_{f(t)} \longleftrightarrow \underbrace{\pi \delta(\omega) + \frac{1}{j\omega}}_{F(\omega)}$$

Using duality property:

$$\pi \delta(t) + \frac{1}{jt} \longleftrightarrow 2\pi u(-\omega)$$

$$\text{or } \frac{1}{2} \left[\delta(t) + \frac{1}{j\pi t} \right] \longleftrightarrow u(-\omega)$$

Application of equation (4.35) yields:

$$\frac{1}{2} \left[\delta(-t) + \frac{j}{\pi t} \right] \longleftrightarrow u(\omega)$$

but $\delta(t)$ is an even function, that is $\delta(-t) = \delta(t)$ and

$$\frac{1}{2} \left[\delta(t) + \frac{j}{\pi t} \right] \longleftrightarrow u(\omega)$$

(c).
$$\underbrace{\sin(\omega_0 t)}_{f(t)} \longleftrightarrow \underbrace{j\pi [\delta(\omega + \omega_0) - \delta(\omega - \omega_0)]}_{F(\omega)}$$

Using duality property:

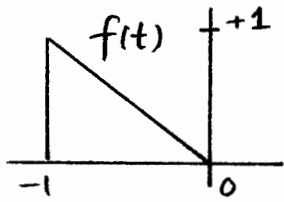
$$\underbrace{j\pi [\delta(t+\omega_0) - \delta(t-\omega_0)]}_{F(t)} \longleftrightarrow \underbrace{2\pi \sin(-\omega_0\omega)}_{2\pi f(-\omega)} = -2\pi \sin(\omega_0\omega)$$

Setting $\omega_0 = T$ yields:

$$\delta(t+T) - \delta(t-T) \longleftrightarrow 2j \sin T\omega.$$

▣

4.3-2



$$\leftrightarrow F(\omega) = \frac{1}{\omega^2} (e^{j\omega} - j\omega e^{-1})$$

$$(b) f_1(t) = f(-t) \rightarrow F_1(\omega) = F(-\omega) = \frac{1}{\omega^2} [e^{-j\omega} + j\omega e^{-1}]$$

(c) $f_2(t) = f(t-1) + f_1(t-1)$ therefore

$$\begin{aligned} F_2(\omega) &= F(\omega) e^{-j\omega} + F_1(\omega) e^{-j\omega} = [F(\omega) + F(-\omega)] e^{-j\omega} \\ &= \left(\frac{1}{\omega^2} e^{j\omega} - \frac{j}{\omega} e^{j\omega} - \frac{1}{\omega^2} + \frac{1}{\omega^2} e^{-j\omega} + \frac{j}{\omega} e^{-j\omega} - \frac{1}{\omega^2} \right) e^{-j\omega} \\ &= \left[\frac{1}{\omega^2} (e^{j\omega} + e^{-j\omega}) - \frac{j}{\omega} (e^{j\omega} - e^{-j\omega}) - \frac{2}{\omega^2} \right] e^{-j\omega} \\ &= \left[\frac{1}{\omega^2} (2 \cos \omega) - \frac{j}{\omega} (2j \sin \omega) - \frac{2}{\omega^2} \right] e^{-j\omega} \\ &= \frac{2e^{-j\omega}}{\omega^2} (\cos \omega + \omega \sin \omega - 1) \end{aligned}$$

(f) $f_5(t)$ can be obtained in three steps: (i) time expanding $f(t)$ by a factor 2. (ii) then delaying it by 2 seconds, (iii) and multiplying it by 1.5 [we may interchange the sequence for steps (i) and (ii)]

The first step yields:

$$f\left(\frac{t}{2}\right) \longleftrightarrow 2F(2\omega) = \frac{1}{2\omega^2} (e^{j2\omega} - j2\omega e^{-1} - 1)$$

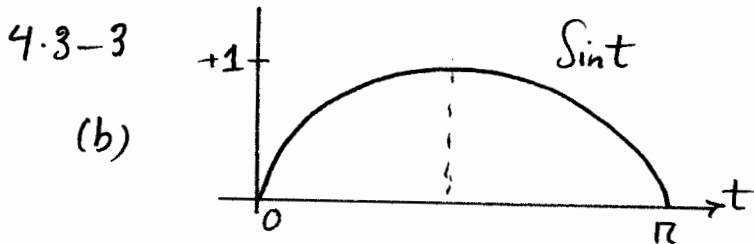
Second step of time delay of 2 seconds yields:

$$\begin{aligned} f\left(\frac{t-2}{2}\right) &\longleftrightarrow \frac{1}{2\omega^2} (e^{j2\omega} - j2\omega e^{-1} - 1) e^{-j2\omega} \\ &= \frac{1}{2\omega^2} (1 - j2\omega - e^{-j2\omega}) \end{aligned}$$

The third step of multiplying the resulting signal by 1.5 yields:

$$f_5(t) = 1.5f\left(\frac{t-2}{2}\right) \longleftrightarrow \frac{3}{4\omega^2} (1 - j2\omega - e^{-j2\omega})$$

▣



$$f(t) = \text{Sint} u(t) + \text{Sin}(t-\pi) u(t-\pi)$$

Note that $\text{Sin}(t-\pi)u(t-\pi)$ is $\text{Sint}u(t)$ delayed by π .

$$\text{Sint}u(t) \longleftrightarrow \frac{\pi}{j2} [\delta(\omega-1) - \delta(\omega+1)] + \frac{1}{1-\omega^2} \text{ and:}$$

$$\text{Sin}(t-\pi)u(t-\pi) \longleftrightarrow \left\{ \frac{\pi}{j2} [\delta(\omega-1) - \delta(\omega+1)] + \frac{1}{1-\omega^2} \right\} e^{-j\pi\omega}$$

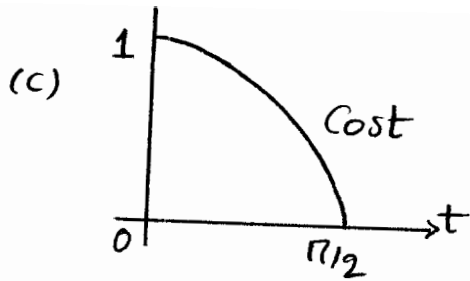
$$\rightarrow F(\omega) = \left\{ \frac{\pi}{j2} [\delta(\omega-1) - \delta(\omega+1)] + \frac{1}{1-\omega^2} \right\} (1 + e^{-j\pi\omega})$$

Recall that $f(x)\delta(x-x_0) = f(x_0)\delta(x-x_0)$. Therefore:

$$\delta(\omega \pm 1)(1 + e^{-j\pi\omega}) = 0.$$

$$\rightarrow F(\omega) = \frac{1}{1-\omega^2} (1 + e^{-j\pi\omega})$$

(9)



$$f(t) = \text{Cost} \left[u(t) - u\left(t - \frac{\pi}{2}\right) \right] = \text{Cost} u(t) - \text{Cost} u\left(t - \frac{\pi}{2}\right)$$

but $\text{Sin}\left(t - \frac{\pi}{2}\right) = -\text{Cost}$ and the second term can be written as:
 $\text{Sin}\left(t - \frac{\pi}{2}\right) u\left(t - \frac{\pi}{2}\right)$. Hence:

$$f(t) = \text{Cost} u(t) + \text{Sin}\left(t - \frac{\pi}{2}\right) u\left(t - \frac{\pi}{2}\right)$$

$$\rightarrow F(\omega) = \frac{\pi}{2} \left[\delta(\omega - 1) + \delta(\omega + 1) \right] + \frac{j\omega}{1 - \omega^2} + \left\{ \frac{\pi}{j^2} \left[\delta(\omega - 1) - \delta(\omega + 1) \right] + \frac{1}{1 - \omega^2} \right\} e^{-j\pi\omega/2}$$

Also because $f(x) \delta(x - x_0) = f(x_0) \delta(x - x_0)$ we have:

$$\delta(\omega \pm 1) e^{-j\frac{\pi\omega}{2}} = \delta(\omega \pm 1) e^{\pm j\frac{\pi}{2}} = \pm j \delta(\omega \pm 1)$$

Therefore:

$$F(\omega) = \frac{j\omega}{1 - \omega^2} + \frac{e^{-j\pi\omega/2}}{1 - \omega^2} = \frac{1}{1 - \omega^2} \left(j\omega + e^{-j\pi\omega/2} \right)$$



4.3-5

$$\begin{aligned} f(t) \text{Sin}(\omega_0 t) &= f(t) \left(\frac{1}{j^2} e^{j\omega_0 t} - \frac{1}{j^2} e^{-j\omega_0 t} \right) \\ &= \frac{1}{2j} \left(f(t) e^{j\omega_0 t} - f(t) e^{-j\omega_0 t} \right) \end{aligned}$$

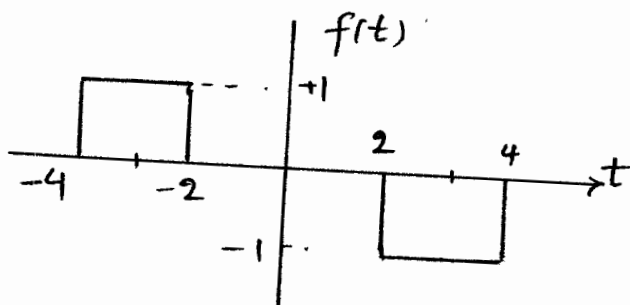
$$\leftrightarrow \frac{1}{j^2} [F(\omega - \omega_0) - F(\omega + \omega_0)]$$

$$\begin{aligned} \frac{1}{2j} [f(t+T) - f(t-T)] &\leftrightarrow \frac{1}{2j} \left[F(\omega) e^{+j\omega T} - F(\omega) e^{-j\omega T} \right] \\ &= \frac{1}{2j} \left[F(\omega) (e^{j\omega T} - e^{-j\omega T}) \right] \end{aligned}$$

(10)

$$= \frac{1}{2j} [F(\omega) (2j \sin \omega T)] = F(\omega) \sin \omega T.$$

For the following signal:



$$f(t) = \text{rect}\left(\frac{t+3}{2}\right) - \text{rect}\left(\frac{t-3}{2}\right)$$

$$\text{rect}(t) \longleftrightarrow \text{sinc}(\omega/2)$$

$$\rightarrow \text{rect}\left(\frac{t}{2}\right) \longleftrightarrow 2 \text{sinc}(\omega)$$

$$\rightarrow \text{rect}\left(\frac{t+3}{2}\right) - \text{rect}\left(\frac{t-3}{2}\right) \longleftrightarrow 2j [2 \text{sinc}(\omega) \sin(3\omega)]$$

$$= j4 \text{sinc}(\omega) \sin(3\omega)$$

▣

4.3-6.

(a) The signal $f(t)$ in this case is a triangle pulse $\Delta\left(\frac{t}{2R}\right)$ multiplied by $\cos(10t)$:

$$f(t) = \Delta\left(\frac{t}{2R}\right) \cos(10t)$$

From table 4.1 (pair 19): $\Delta\left(\frac{t}{2R}\right) \longleftrightarrow R \text{sinc}^2\left(\frac{\pi\omega}{2}\right)$. Using modulation property (4.41) it follows that:

$$f(t) = \Delta\left(\frac{t}{2R}\right) \cos(10t) \longleftrightarrow \frac{R}{2} \left\{ \text{sinc}^2\left[\frac{\pi(\omega-10)}{2}\right] + \text{sinc}^2\left[\frac{\pi(\omega+10)}{2}\right] \right\}$$

(b) The signal $f(t)$ here is the same as the signal in Fig. (a) delayed by $2R$. From time shifting property, its Fourier transform can be written as:

(11)

$$F(\omega) = \frac{\tau}{2} \left\{ \text{sinc}^2 \left[\frac{\tau(\omega - \omega_0)}{2} \right] + \text{sinc}^2 \left[\frac{\tau(\omega + \omega_0)}{2} \right] \right\} e^{-j2\tau\omega}$$

The Fourier transform in this case is the same as that in part (a) multiplied by $e^{-j2\tau\omega}$. This multiplying factor represents a linear phase spectrum $-2\tau\omega$. Thus we have an "amplitude spectrum" as well as a linear "phase spectrum" $\angle F(\omega) = -2\tau\omega$.

(c) In this case the signal is identical to that in fig. b except that the basic pulse is $\text{rect}(\frac{t}{2\tau})$ instead of a triangle pulse $\Delta(\frac{t}{2\tau})$.
Now:

$$\text{rect}(\frac{t}{2\tau}) \longleftrightarrow 2\tau \text{sinc}(\tau\omega)$$

Using the same argument as for part (b), we obtain:

$$F(\omega) = \tau \left\{ \text{sinc}[\tau(\omega + \omega_0)] + \text{sinc}[\tau(\omega - \omega_0)] \right\} e^{-j2\tau\omega}$$

▣

4.3-8

$$(a) \quad u(t) \longleftrightarrow \tau \delta(\omega) + \frac{1}{j\omega}$$

$$e^{\lambda t} u(t) \longleftrightarrow \frac{1}{j\omega - \lambda}$$

If $f(t) = e^{\lambda t} u(t) * u(t)$ then

$$F(\omega) = \left(\frac{1}{j\omega - \lambda} \right) \left(\tau \delta(\omega) + \frac{1}{j\omega} \right)$$

$$= \frac{\tau \delta(\omega)}{j\omega - \lambda} + \left[\frac{1}{j\omega(j\omega - \lambda)} \right] = -\frac{\tau}{\lambda} \delta(\omega) + \left[\frac{-\frac{1}{\lambda}}{j\omega} + \frac{\frac{1}{\lambda}}{j\omega - \lambda} \right]$$

$$= \frac{1}{\lambda} \left[\frac{1}{j\omega - \lambda} - \left(\tau \delta(\omega) + \frac{1}{j\omega} \right) \right]$$

Taking the inverse Fourier transform of this equations yields:

(12)

$$f(t) = \frac{1}{\lambda} (e^{\lambda t} - 1) u(t).$$

$$(c) \quad e^{\lambda_1 t} u(t) \longleftrightarrow \frac{1}{j\omega - \lambda_1}; \quad e^{\lambda_2 t} u(-t) \longleftrightarrow -\frac{1}{j\omega - \lambda_2}$$

If $f(t) = e^{\lambda_1 t} u(t) * e^{\lambda_2 t} u(-t)$ then:

$$F(\omega) = \frac{-1}{(j\omega - \lambda_1)(j\omega - \lambda_2)} = \frac{\frac{1}{\lambda_2 - \lambda_1}}{j\omega - \lambda_1} - \frac{\frac{1}{\lambda_2 - \lambda_1}}{j\omega - \lambda_2}$$

$$\text{therefore: } f(t) = \frac{1}{\lambda_2 - \lambda_1} [e^{\lambda_1 t} u(t) + e^{\lambda_2 t} u(-t)].$$

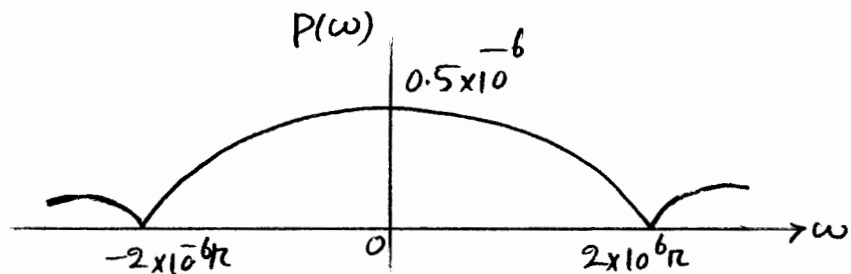
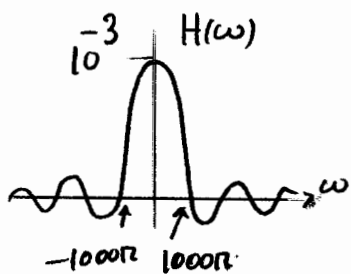
Note that because $\lambda_2 > 0$ the inverse transform of $-\frac{1}{j\omega - \lambda_2}$ is $e^{\lambda_2 t} u(-t)$ and not $-e^{\lambda_2 t} u(t)$. The Fourier transform of the latter does not exist because $\lambda_2 > 0$.



4.4-4

$$h(t) = \text{rect}\left(\frac{t}{10^{-3}}\right) \longleftrightarrow H(\omega) = 10^{-3} \text{Sinc}\left(\frac{\omega}{2000}\right)$$

$$p(t) = \Delta\left(\frac{t}{10^{-6}}\right) \longleftrightarrow P(\omega) = 0.5 \times 10^{-6} \text{Sinc}^2\left(\frac{\omega}{4 \times 10^6}\right)$$



From the figures, it is clear that $H(\omega)$ is much narrower than $P(\omega)$ and we may consider $P(\omega)$ to be nearly constant of value $P(0) = 10^{-6}/2$ over the entire band of $H(\omega)$. Hence:

$$Y(\omega) = P(\omega)H(\omega) \approx P(0)H(\omega) = 0.5 \times 10^{-6} H(\omega) \rightarrow y(t) = 0.5 \times 10^{-6} h(t)$$

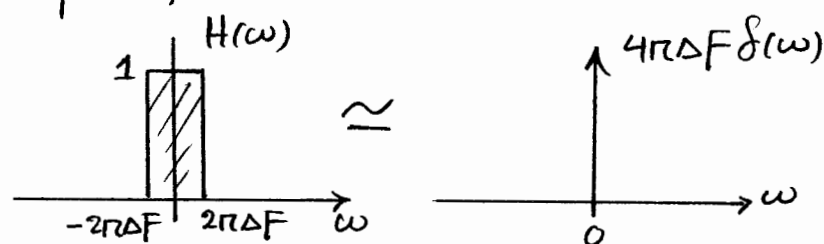
Recall that $h(t)$ is the unit impulse response of the system.

Therefore the output $y(t)$ is equal to the system response to an input $0.5 \times 10^{-6} \delta(\omega) = A \delta(\omega)$.

□

4.6-3

if $f^2(t) \leftrightarrow A(\omega)$ then the output $Y(\omega) = A(\omega) H(\omega)$ where $H(\omega)$ is the lowpass filter transfer function. As $\Delta F \rightarrow 0$, we may express $H(\omega)$ as an impulse function of area $4\pi\Delta f$:



$$\rightarrow H(\omega) \approx [4\pi\Delta F] \delta(\omega) \text{ and } Y(\omega) \approx [4\pi A(\omega)\Delta F] \delta(\omega) \\ = [4\pi A(0)\Delta F] \delta(\omega)$$

Here we used the property $f(x) \delta(x) = f(0) \delta(x)$. This yields:

$$y(t) = 2A(0)\Delta F.$$

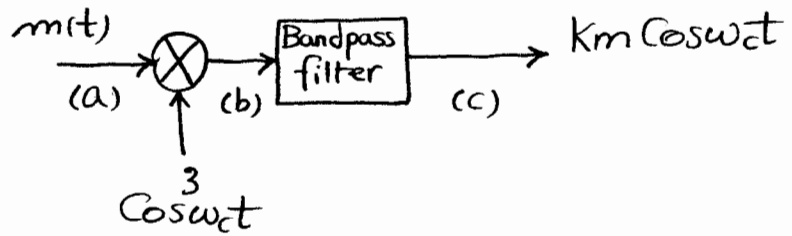
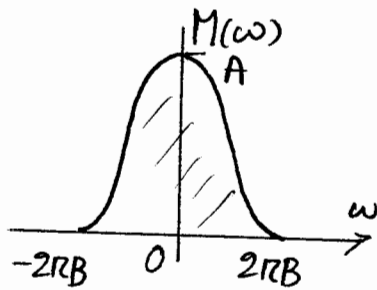
Now, because $f^2(t) \leftrightarrow A(\omega)$, we have:

$$A(\omega) = \int_{-\infty}^{+\infty} f^2(t) e^{-j\omega t} dt \rightarrow A(0) = \int_{-\infty}^{+\infty} f^2(t) e^0 dt = \int_{-\infty}^{+\infty} f^2(t) dt = Ef.$$

$$\Rightarrow y(t) = 2Ef\Delta f.$$

□

4.7-2



ca) The signal at point b is:

$$f(t) = m(t) \cos^3 \omega_c t$$

Using the following trigonometric equality:

$$\cos 3\omega_c t = 4 \cos^3 \omega_c t - 3 \cos \omega_c t \rightarrow \cos^3 \omega_c t = \frac{3}{4} \cos \omega_c t + \frac{1}{4} \cos 3\omega_c t$$

$$\rightarrow f(t) = m(t) \left(\frac{3}{4} \cos \omega_c t + \frac{1}{4} \cos 3\omega_c t \right)$$

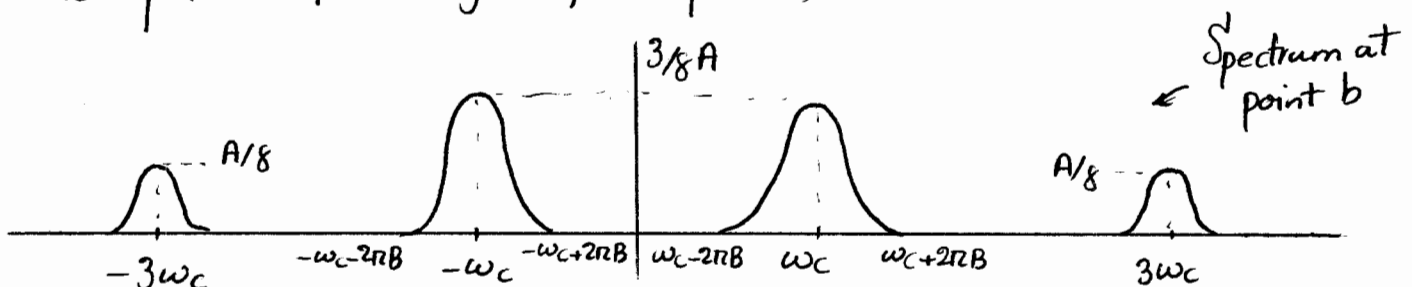
$$\frac{3}{4} m(t) \cos \omega_c t + \frac{1}{4} m(t) \cos (3\omega_c t)$$

Using the "modulation" property of Fourier transform, the frequency spectrum of $f(t)$ can be obtained as:

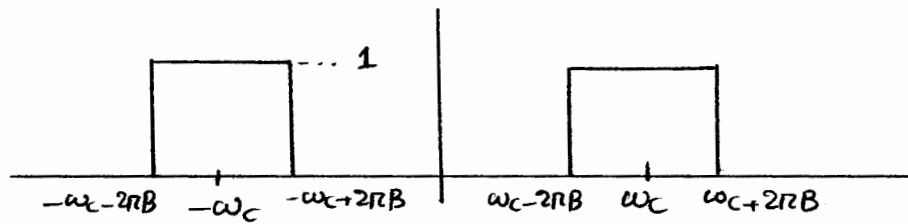
$$F(\omega) = \frac{3}{8} [M(\omega - \omega_c) + M(\omega + \omega_c)] + \frac{1}{8} [M(\omega - 3\omega_c) + M(\omega + 3\omega_c)]$$

desired modulated signal spectrum centered at $\omega = \pm \omega_c$ undesired signal spectrum centered at $\omega = \pm 3\omega_c$

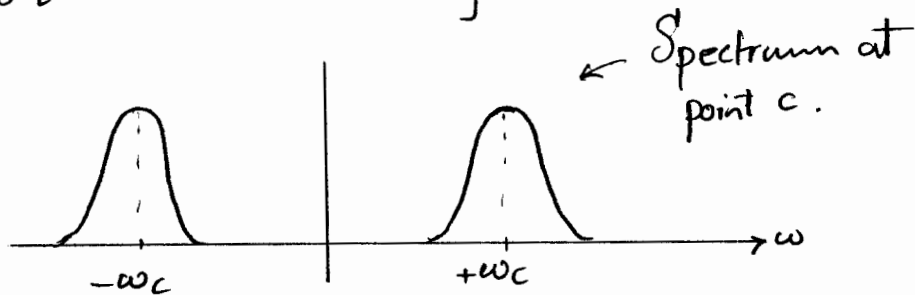
The spectrum of the signal $f(t)$ (point b) is shown below:



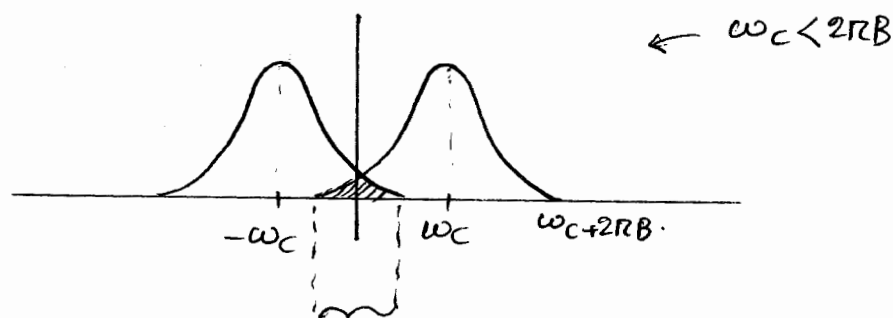
For the bandpass filter centered at $\pm\omega_c$ with bandwidth $4RB$, the Spectrum is:



Passing $F(\omega)$ through the shown bandpass filter allows the desired modulated signal to pass, but the undesired signal centered at $\omega = \pm 3\omega_c$ will be suppressed. Hence the output of the bandpass filter is a signal with spectrum $\frac{3}{8} [M(\omega - \omega_c) + M(\omega + \omega_c)]$.



c) The minimum usable value of ω_c is $2RB$ in order to avoid spectral folding at dc. The following figure illustrates a value of ω_c less than $2RB$. Notice to the region where the modulated signal overlaps. Overlapping the spectrum in this case is equal to losing information which is undesirable.



Overlapping of the spectrum of the modulated signal (losing information!)

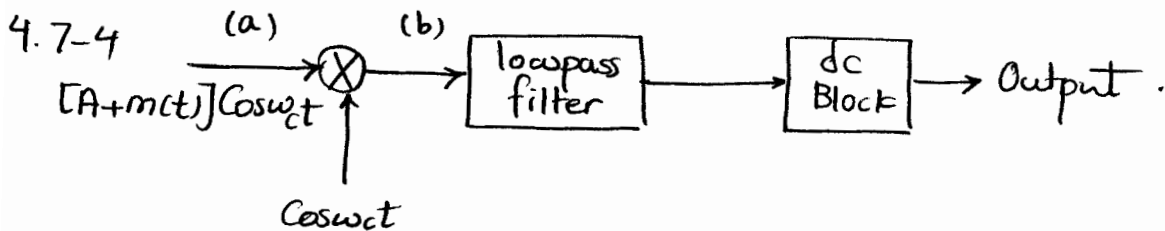
$$d) m(t) \cos^2 \omega_c t = \frac{m(t)}{2} [1 + \cos(2\omega_c t)].$$

$$= \frac{1}{2} m(t) + \frac{1}{2} m(t) \cos(2\omega_c t)$$

This signal at point (b) consists of a baseband signal $\frac{1}{2} m(t)$ and a modulated signal $\frac{1}{2} m(t) \cos(2\omega_c t)$ which has a carrier frequency $2\omega_c$ not the desired value ω_c . Both of these two components will be suppressed by the filter, whose center frequency is ω_c . Hence this system will not do the desired job.

- (e) You may verify that the identity for $\cos^n \omega_c t$ contains a term $\cos(\omega_c t)$ when n is odd. This is not true when n is even. Hence the system works for a carrier $\cos^n \omega_c t$ only when n is odd.

▣



$$f_a(t) = [A + m(t)] \cos \omega_c t. \text{ Hence}$$

$$f_b(t) = [A + m(t)] \cos^2 \omega_c t = \frac{1}{2} [A + m(t)] + \frac{1}{2} [A + m(t)] \cos(2\omega_c t)$$

The first term is a lowpass signal centered at $\omega = 0$. The lowpass filter allows this term to pass, but suppresses the second term whose spectrum is centered at $\pm 2\omega_c$. Hence the output of the lowpass filter is:

$$y(t) = A + m(t)$$

The dc block simply suppresses the dc component, A . This shows that the system can demodulate AM signal regardless of the value of A . This is synchronous or coherent demodulation.