

①

Q₁: This is an odd signal

$$a_0 = 0, a_n = 0$$

$$b_n = \frac{4}{T_0} \int_0^{\frac{T_0}{2}} f(t) \sin n\omega_0 t \, dt$$

$$f(t) = \frac{2t}{T_0} \quad -\frac{T_0}{2} \leq t \leq \frac{T_0}{2}, \quad T_0 = 2 \text{ msec}$$

$$\Rightarrow b_n = \frac{4}{T_0} \int_0^{\frac{T_0}{2}} \frac{2t}{T_0} \sin n\omega_0 t \, dt$$

$$b_n = \frac{8}{T_0^2} \int_0^{\frac{T_0}{2}} \frac{2t}{T_0} \sin n\omega_0 t \, dt$$

Integrating by parts, we get

$$b_n = \frac{8}{T_0^2} \left[-\frac{1}{n\omega_0} t \cos n\omega_0 t \Big|_0^{\frac{T_0}{2}} + \frac{1}{n\omega_0} \int_0^{\frac{T_0}{2}} \cos n\omega_0 t \, dt \right]$$

$$b_n = -\frac{8}{nT_0^2 \omega_0} \times \frac{T_0}{2} \cos n\omega_0 \frac{T_0}{2} + 0$$

Noticing that $\omega_0 T_0 = 2\pi$, we have

$$b_n = \frac{-4}{n \cdot 2\pi} \cos n\pi$$

$$b_n = \frac{2}{n\pi} (-1)^{n+1} \leftarrow \textcircled{a}$$

$$D_n = \frac{a_n - j b_n}{2} = \frac{j(-1)^n}{n\pi}, \quad D_0 = a_0 = 0$$

To get signal power, we have

$$P = \frac{1}{T_0} \int_{-\frac{T_0}{2}}^{\frac{T_0}{2}} |f(t)|^2 dt$$

$$P = \frac{1}{T_0} \int_{-\frac{T_0}{2}}^{\frac{T_0}{2}} \frac{4t^2}{T_0^2} dt$$

$$P = \frac{1}{T_0^3} \left. \frac{4t^3}{3} \right|_{-\frac{T_0}{2}}^{\frac{T_0}{2}}$$

$$P = \left(\frac{4}{24} + \frac{4}{24} \right) = \frac{1}{3} \text{ W/Hr}$$

$$\text{Also, } 2 \sum_{n=1}^{\infty} |D_n|^2 = 2 \sum_{n=1}^{\infty} \frac{1}{n^2 \pi^2} \quad (2)$$

$$2 \sum_{n=1}^{\infty} |D_n|^2 = \frac{2}{\pi^2} \sum_{n=1}^{\infty} \frac{1}{n^2} = \frac{2}{\pi^2} \cdot \frac{\pi^2}{6}$$

$$\Rightarrow 2 \sum_{n=1}^{\infty} |D_n|^2 = \frac{1}{3} = \rho$$

Q1.c) This circuit is a parallel resonance circuit

$$\omega_c = \frac{1}{\sqrt{LC}} = 3.135 \times 10^3 \\ = 1000\pi = \omega_0$$

only first harmonic is allowed to pass unattenuated

$$y_1(t) \approx D_1 e^{j\omega_0 t} + D_{-1} e^{-j\omega_0 t}$$

$$y_1(t) \approx 2|D_1| \cos(\omega_0 t + \angle D_1)$$

$$\text{but } |D_1| = \frac{1}{\pi}, \angle D_1 = -\frac{\pi}{2}$$

$$\Rightarrow y_1(t) \approx \frac{2}{\pi} \cos(\omega_0 t - \frac{\pi}{2})$$

$$y_1(t) \approx \frac{2}{\pi} \sin \omega_0 t$$

$$y_1(t) = 6_1 \sin(1000\pi t)$$

$$Q2: f = \frac{1}{10^{-3}} = 1 \text{ kHz}$$

Nyquist frequency is $2f = 2 \text{ kHz}$

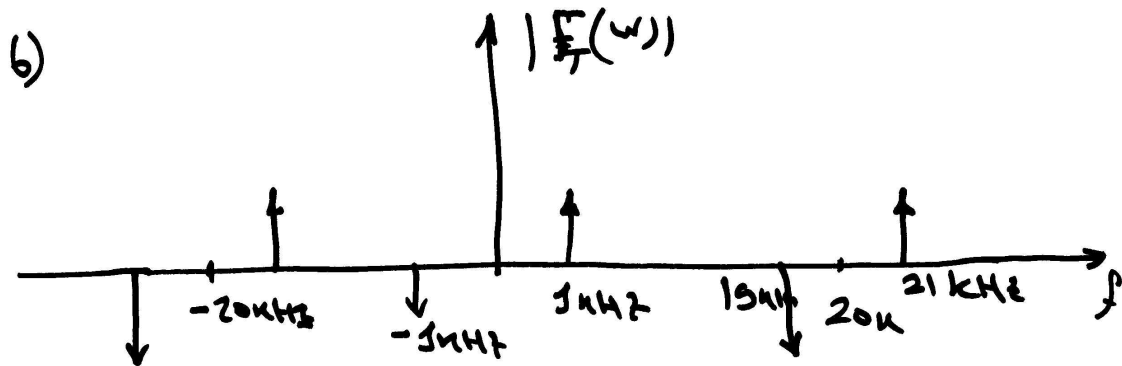
⇒ Sampling frequency = 20 kHz

⇒ 20 Samplers per Cycle

$$T = \frac{1}{20 \times 10^3} = 5 \times 10^{-5} \text{ sec}$$

$$f_T(t) = \sum_{k=-\infty}^{\infty} \sin(2\pi \times 10^3 \times k \times T) \delta(t - kT)$$

$$f_T(t) = \sum_{k=-\infty}^{k=\infty} \sin(0.1\pi k) \delta(t - k * 5 * 10^5) \quad (4)$$

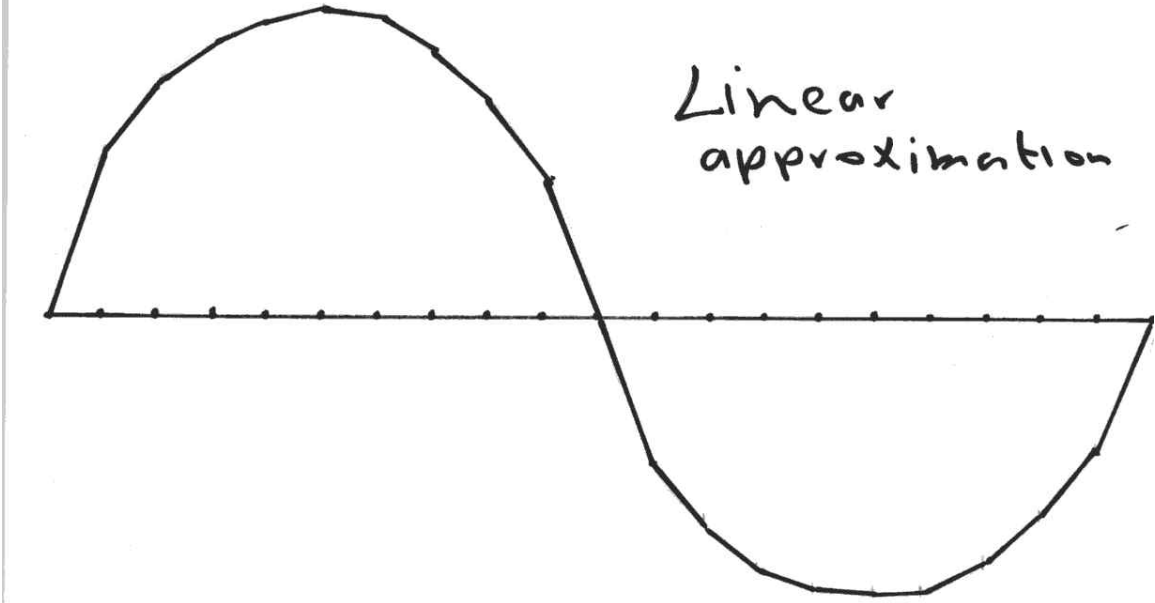


c) The reconstructed signal is

$$f_c(t) = \sum_{k=-\infty}^{\infty} \sin(0.1\pi k) h(t - k * 5 * 10^5)$$

d) This filter obtains a linear approximation of the signal,

Linear
approximation



Q3)

$$x(t) = 2e^{-t} \Rightarrow X(\omega) = \frac{2}{j\omega + 1}$$

$$y(t) = x(2t) = 2e^{-2t}u(t)$$

$$Y(\omega) = \frac{1}{2} F\left(\frac{\omega}{2}\right) = \frac{1}{j\frac{\omega}{2} + 1} = \frac{2}{j\omega + 2}$$

$$G(\omega) = \frac{\frac{1}{j\omega} Y(\omega)}{1 + \frac{1}{j\omega}} = \frac{1}{j\omega + 1} \cdot \frac{2}{j\omega + 2}$$

$$G(\omega) = \frac{2}{j\omega + 1} - \frac{2}{j\omega + 2}$$

$$g(t) = 2e^{-t}u(t) - 2e^{-2t}u(t)$$

$$V(\omega) = \frac{2}{j\omega + 1} \cdot \frac{1}{1 + j\omega}$$

$$v(t) = 2te^{-t}u(t)$$

$$W(t) = v(2t) = 2(2t)e^{-2t}u(2t)$$

$$W(t) = 4te^{-2t}u(t)$$

$W(t) \neq g(t)$ as order makes a difference

$$Q4) \cdot T_0 = 5T = 0.5 \text{ sec}$$

$$\omega_0 = \frac{2\pi}{T_0} = 4\pi \text{ rad/sec}$$

$$\Omega_0 = 4\pi \times 0.1 = 0.4\pi \text{ rad/sample}$$

$$F_r = \sum_{k=0}^4 f_k e^{-jk\Omega_0 r}$$

$$F_r = T \sum_{k=0}^4 f(kT) e^{-jk\Omega_0 r}$$

$$F_0 = 0.1 (1 + 2 - 2 - 1 + 0) = 0$$

$$F_1 = T \sum_{k=0}^4 f(kT) e^{-jk\Omega_0}$$

$$F_1 = 0.1 (1 \cdot e^{-j0} + 2 e^{-j0.4\pi} - 2 e^{-j0.8\pi} - 1 e^{-j1.2\pi} + 0)$$

$$F_2 = 0.1 \sum_{k=0}^4 f(kT) e^{-jk(0.8\pi)}$$

$$F_3 = 0.1 \sum_{k=0}^4 f(kT) e^{-jk(1.2\pi)}$$

$$F_4 = 0.1 \sum_{k=0}^4 f(kT) e^{-jk(1.6\pi)}$$

$$a) \quad y'' + 2y' + 5y = \delta(t)$$

$$\Rightarrow s^2 Y(s) - sy(0) - y'(0) + 2(sY(s) - y(0)) + 5Y(s) = 1$$

For ZIR we have

$$Y(s)(s^2 + 2s + 5) - 2s - 0 - 2 = 0$$

$$Y(s) = \frac{2s + 2}{(s^2 + 2s + 5)} = \frac{2(s+1)}{(s+1)^2 + 4}$$

$$\underset{\text{ZIR}}{y}(t) = 2e^{-t} \cos 2t$$

For ZSR, we have

$$Y(s)(s^2 + 2s + 5) = 1$$

$$Y(s) = \frac{1}{(s+1)^2 + 4}$$

$$\underset{\text{ZSR}}{y}(t) = \frac{1}{2} e^{-t} \sin 2t$$

$$Q6) \quad T(s) = \frac{G(s)}{1 + 3G(s)}$$

$$T(s) = \frac{\frac{s}{s+1000}}{1 + \frac{3s}{s+1000}} = \frac{s}{4s+1000}$$

$$T(s) = \frac{s}{4(s+250)}$$

* Original filter $\frac{s}{s+1000}$ is a high pass filter with gain = 1 & cut-off frequency 1000 rad/sec

* New filter $\frac{s}{4(s+250)}$ is also a high pass filter with gain = $\frac{1}{4}$ & cut-off frequency of 250 rad/sec
