

## Lecture #23

from Chapter 7 in Jaeger, Chapter 15 in Spencer

## CMOS Logic Design

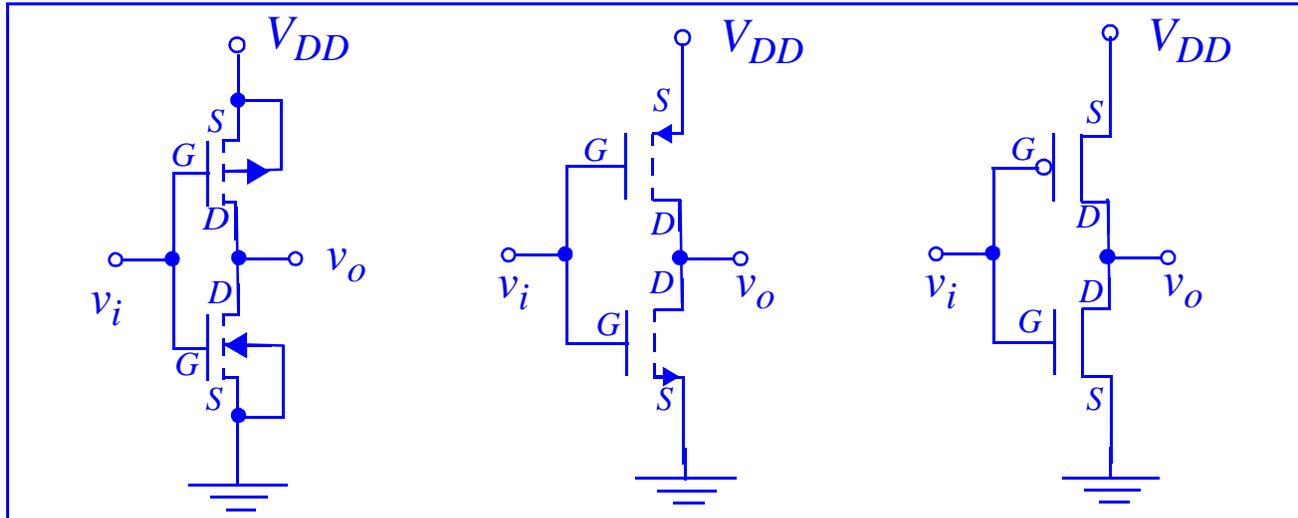
### **Outline/Learning Objectives:**

- Explain the basic function of the primary digital circuit building block -the inverter.
- Describe (static) CMOS logic inverter technology.
- Analyze the (static) CMOS inverter circuit, including its voltage transfer characteristic and static and dynamic power dissipation.

### **Selected problems:**

- 7.8, 7.23, 7.58, 7.60, 7.63

## CMOS Inverter



B tied to S in both NMOS and PMOS devices.

$$V_{TH \text{ NMOS}} - V_{TN} = V_{T0N} + \gamma_N (\sqrt{v_{SBN} + 2\phi_{FN}} - \sqrt{2\phi_{FN}}) \quad (1)$$

$$V_{TH \text{ PMOS}} - V_{TP} = V_{T0P} - \gamma_P (\sqrt{v_{BSP} + 2\phi_{FP}} - \sqrt{2\phi_{FP}}) \quad (2)$$

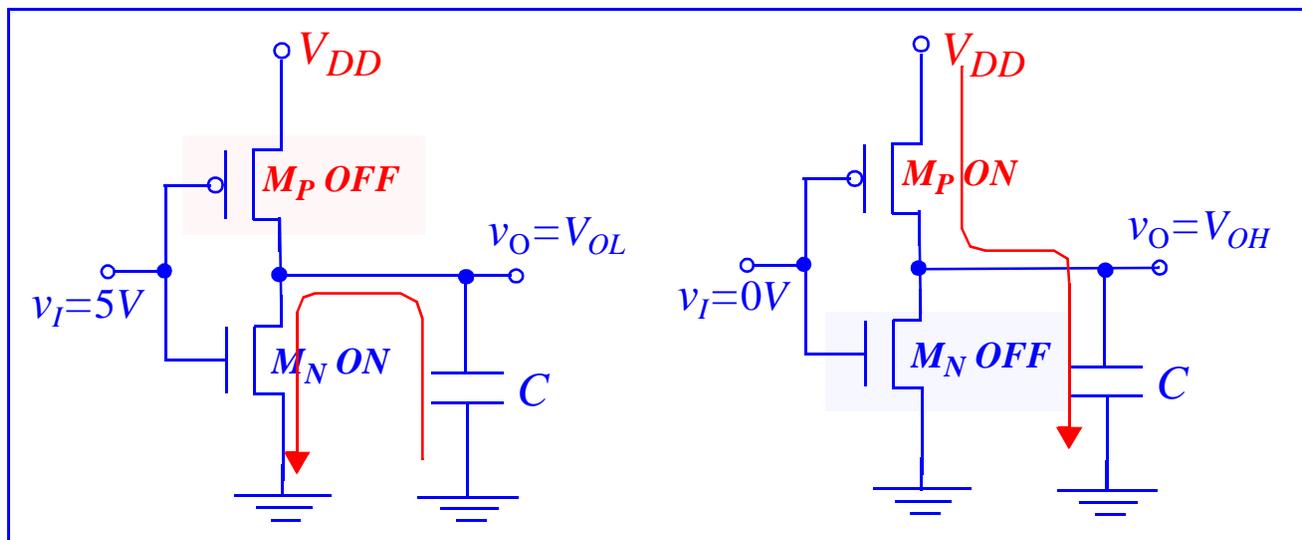
$$\gamma = \frac{\sqrt{2qN_{sub}\epsilon_s}}{C_{ox}} = \text{body effect parameter (V}^{0.5}\text{)} \quad (3)$$

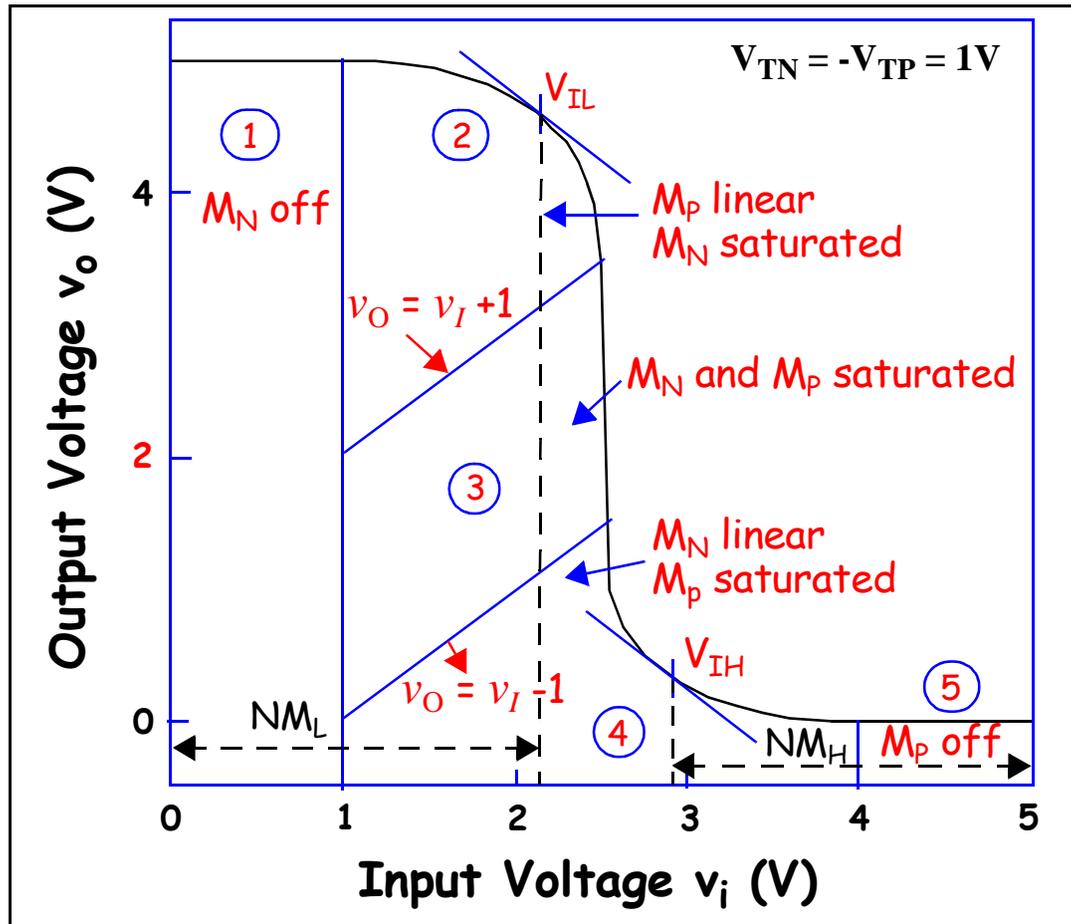
$$2\phi_F = \text{surface potential parameter} \quad \phi_F = \frac{kT}{q} \ln\left(\frac{N_{sub}}{n_i}\right) \quad (4)$$

## Sample NMOS and PMOS Parameters

Parameter	NMOS	PMOS
$V_{T0}$	1 V	-1 V
$\gamma$	$0.5 \text{ V}^{1/2}$	$0.75 \text{ V}^{1/2}$
$\phi_F$	0.60 V	0.70 V
$K'$	$25 \mu\text{A}/\text{V}^2$	$10 \mu\text{A}/\text{V}^2$
$\lambda$	$0.02 \text{ V}^{-1}$	$0.02 \text{ V}^{-1}$

## Static Characteristics of CMOS Inverter





Sat. of PMOS requires  $v_{SD} \geq v_{SG} + V_{TP}$  or  $(5 - v_o) \geq (5 - v_i) - 1$  or  $v_o \leq v_i + 1$  (5)

Sat. of NMOS requires  $v_{DS} \geq v_{GS} - V_{TN}$  or  $v_o \geq v_i - 1$  (6)

## Regions of Operation of MOSTs in Symmetrical CMOS Inv.

Region	$v_I$	$v_O$	NMOS	PMOS
1	$v_I \leq V_{TN}$	$V_{OH} = V_{DD}$	Off	Lin
2	$V_{TN} < v_I \leq v_O + V_{TP}$	High	Sat	Lin
3	$v_I \approx V_{DD}/2$	$V_{DD}/2$	Sat	Sat
4	$v_O + V_{TN} < v_I \leq V_{DD} -  V_{TP} $	Low	Lin	Sat
5	$v_I \geq V_{DD} -  V_{TP} $	$V_{OL} = 0$	Lin	Off

### Noise Margins for the CMOS Inverter

High Input - PMOS sat., NMOS lin., so  $i_{DSN} = i_{DSP}$

$$K_n \left( v_I - V_{TN} - \frac{v_O}{2} \right) v_O = \frac{K_p}{2} (V_{DD} - v_I + V_{TP})^2 \quad (7)$$

$$K_R (2v_I - 2V_{TN} - v_O) v_O = (V_{DD} - v_I + V_{TP})^2 \quad \text{Note } \boxed{K_R = K_n / K_p} \quad (8)$$

$$v_O = (v_I - V_{TN}) \pm \sqrt{(v_I - V_{TN})^2 - \frac{(V_{DD} - v_I + V_{TP})^2}{K_R}} \quad (9)$$

$$V_{IH} = \frac{2K_R(V_{DD} - V_{TN} + V_{TP})}{(K_R - 1)\sqrt{1 + 3K_R}} - \frac{V_{DD} - K_R V_{TN} + V_{TP}}{K_R - 1} \quad (10)$$

$$NM_H = V_{OH} - V_{IH} = V_{DD} - V_{IH} \quad (11)$$

For  $K_R = 1$ ,

$$V_{IH} = \frac{5V_{DD} + 3V_{TN} + 5V_{TP}}{8} \quad (12)$$

and

$$NM_H = \frac{3V_{DD} - 3V_{TN} - 5V_{TP}}{8} \quad (13)$$

**Low Input** - PMOS lin., NMOS sat.,  $i_{DSN} = i_{DSP}$

$$K_p \left( V_{DD} - v_I + V_{TP} - \frac{V_{DD} - v_O}{2} \right) (V_{DD} - v_O) = K_n (v_I - V_{TN})^2 \quad (14)$$

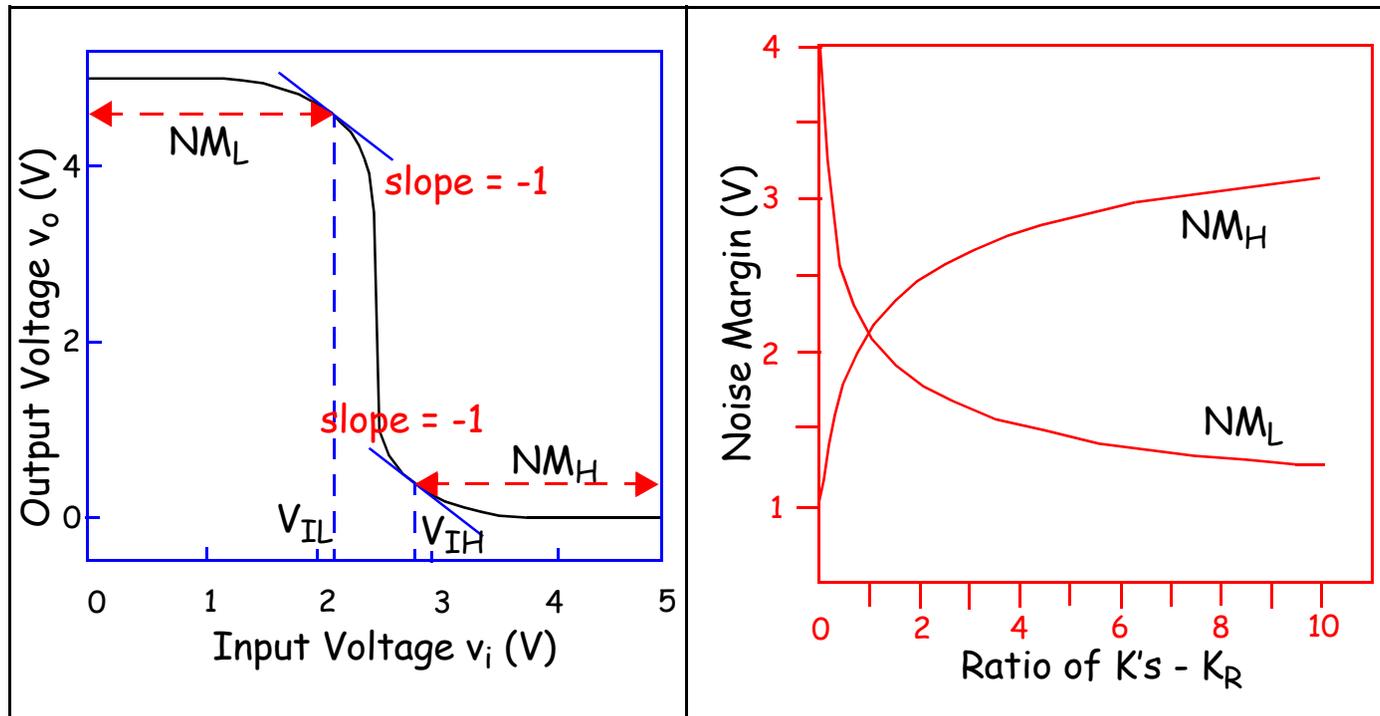
$$(V_{DD} - 2v_I + 2V_{TP} - v_O)(V_{DD} - v_O) = K_R (v_I - V_{TN})^2 \quad \text{Note } K_R = K_n / K_p \quad (15)$$

$$V_{IL} = \frac{2\sqrt{K_R}(V_{DD} - V_{TN} + V_{TP})}{(K_R - 1)\sqrt{K_R + 3}} - \frac{V_{DD} - K_R V_{TN} + V_{TP}}{K_R - 1} \quad (16)$$

$$NM_L = V_{IL} - V_{OL} = V_{IL} - 0 \quad (17)$$

For  $K_R = 1$ , 
$$V_{IL} = \frac{3V_{DD} + 5V_{TN} + 3V_{TP}}{8} \quad (18)$$

and 
$$NM_L = V_{IL} \quad (19)$$



Example: What are the noise margins of a minimum size CMOS inverter in which both W/L ratios are 2/1 with  $K_n' = 25 \mu\text{A}/\text{V}^2$  and  $K_p' = 10 \mu\text{A}/\text{V}^2$ ,  $V_{DD} = 5.2 \text{ V}$  and  $V_{TN} = -V_{TP} = 1 \text{ V}$ ?

$$K_R = \frac{K_n}{K_p} = 2.5$$

$$V_{IH} = \frac{2K_R(V_{DD} - V_{TN} + V_{TP})}{(K_R - 1)\sqrt{1 + 3K_R}} - \frac{V_{DD} - K_R V_{TN} + V_{TP}}{K_R - 1}$$

$$V_{IH} = \frac{2(2.5)(5.2 - 1 - 1)}{(2.5 - 1)(\sqrt{1 + 3(2.5)})} - \frac{5.2 - 2.5(1) - 1}{2.5 - 1} = 2.53 \text{ V}$$

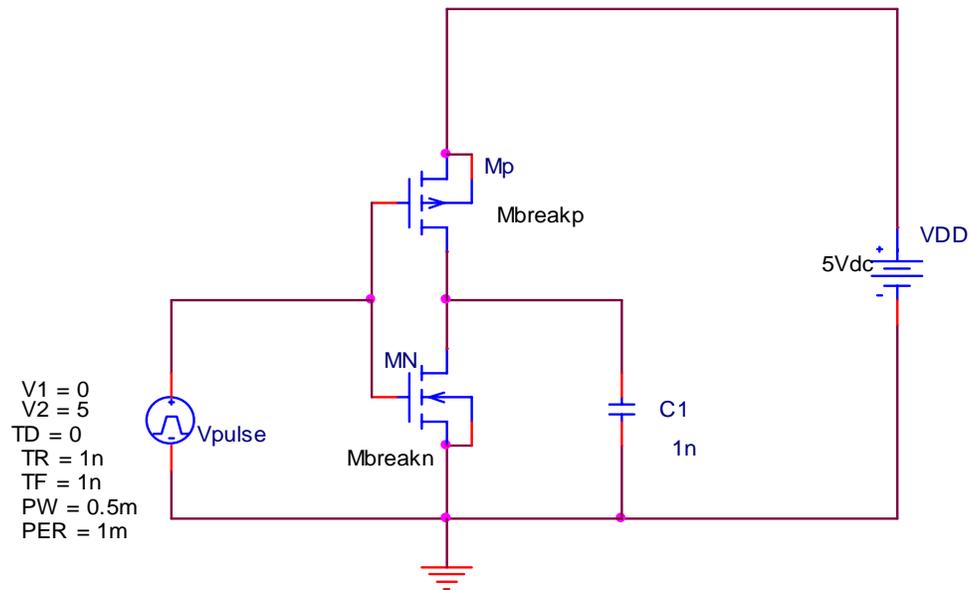
$$\therefore NM_H = V_{DD} - V_{IH} = 5.2 - 2.53 = 2.67 \text{ V}$$

$$V_{IL} = \frac{2\sqrt{K_R}(V_{DD} - V_{TN} + V_{TP})}{(K_R - 1)\sqrt{K_R + 3}} - \frac{V_{DD} - K_R V_{TN} + V_{TP}}{K_R - 1}$$

$$V_{IL} = \frac{2(\sqrt{2.5})(5.2 - 1 - 1)}{(2.5 - 1)(\sqrt{2.5 + 3})} - \frac{5.2 - 2.5(1) - 1}{2.5 - 1} = 1.83 \text{ V}$$

$$\therefore NM_L = V_{IL} = 1.83 \text{ V}$$

# PSPICE EXAMPLE



**\*Libraries:**

**\* Local Libraries :**

**.LIB ".\example9.lib"**

**\* From [PSPICE NETLIST] section of C:\Program Files\OrcadLite\PSpice\PSpice.ini file:**

**.lib "nom.lib"**

**\*Analysis directives:**

**.TRAN 0 20ms 0 1u**

## PSPICE EXAMPLE (Cont'd)

```
.PROBE V(*) I(*) W(*) D(*) NOISE(*)  
.INC ".\example9-SCHEMATIC1.net"  
**** INCLUDING example9-SCHEMATIC1.net ****  
* source EXAMPLE9  
M_MN      N00163 N00248 0 0 Mbreakn  
V_Vpulse  N00248 0  
+PULSE 0 5 0 1n 1n 0.5m 1m  
V_VDD     N00144 0 5Vdc  
M_Mp      N00163 N00248 N00144 N00144 Mbreakp  
C_C1      0 N00163 1n  
**** RESUMING example9-SCHEMATIC1-Example9.sim.cir ****  
.END  
  
**** MOSFET MODEL PARAMETERS
```

```
*****
```

## PSPICE EXAMPLE (Cont'd)

```
      Mbreakn      Mbreakp
      NMOS        PMOS
      LEVEL      1      1
L 100.000000E-06 100.000000E-06
W 100.000000E-06 100.000000E-06
VTO 1 -1
KP 500.000000E-06 500.000000E-06
GAMMA 0 0
PHI .6 .6
LAMBDA 0 0
```

```
**** INITIAL TRANSIENT SOLUTION TEMPERATURE = 27.000 DEG C
```

```
*****
```

```
NODE VOLTAGE NODE VOLTAGE NODE VOLTAGE NODE VOLTAGE
(N00144) 5.0000 (N00163) 5.0000 (N00248) 0.0000
```

## PSPICE EXAMPLE (Cont'd)

### VOLTAGE SOURCE CURRENTS

NAME	CURRENT
V_Vpulse	0.000E+00
V_VDD	-5.010E-12

TOTAL POWER DISSIPATION 2.51E-11 WATTS

JOB CONCLUDED

TOTAL JOB TIME 2.79

# PSPICE EXAMPLE (Cont'd)

